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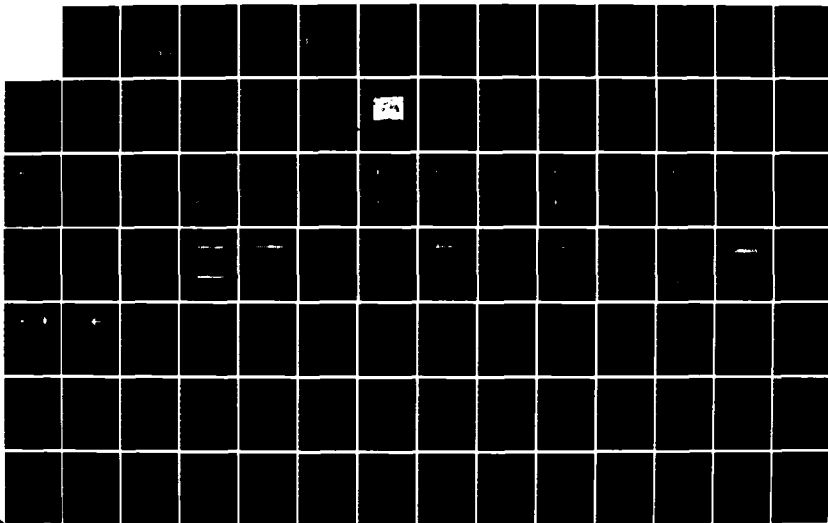
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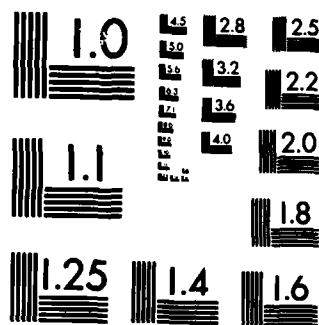
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# EMPLOYMENT OF ADAPTIVE LEARNING TECHNIQUES FOR THE DISCRIMINATION OF ACOUSTIC EMISSIONS

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PHASE II FINAL REPORT ✓  
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19. ABSTRACT (Continue on reverse if necessary and identify by block number)  Under Phase I of this contract, several new acoustic emission techniques were developed. These techniques showed promise as a means of adaptively learning and removing multimode and multipath effects from acoustic emission signals in a preprocessing step prior to source characterization. In this Phase II effort, software was developed to implement the techniques, and a series of experiments was carried out to assess the effectiveness and practicality of these techniques on real signals. Analysis of the data revealed that reasonably accurate transfer functions could be determined adaptively, when a flat response wide-band transducer was employed, and when relatively short record lengths were used. Available commercial transducers with the requisite frequency response are quite fragile, however, and are far too insensitive for practical use. Pattern recognition techniques were also explored as a means of characterizing acoustic emission sources; specifically, details of the pulse microstructure were examined for evidence of characterizable features. No such evidence was found in either "raw" (reverberation dominated) or processed data. <i>Originals supplied by NRC include:</i>			
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## Section 1

### INTRODUCTION AND SUMMARY

#### 1.1 Introduction

Until recently, workers in the acoustic emission (AE) research community appear to have been unaware of a number of techniques applied successfully in other research areas (including sonar, seismology, and speech processing) where certain advanced signal processing techniques are frequently used to deal with distortion effects imposed on the signals of interest. For example, volume reverberation and multimode propagation effects are common problems in these applications, and much effort has been expended in the development of advanced techniques to compensate for them. Since these effects also severely distort real AE signals, those advanced techniques are expected to be useful in improving the performance and reliability of AE-based nondestructive evaluation (NDE) systems. This report describes the results of a research program that seeks to capitalize on those successful research efforts, extending them and applying them to the problems of the detection, location, and characterization of AE events.

#### 1.2 Signal Processing

Current AE analysis techniques make limited use of the phase information in the received signal and rely on a variety of incoherent techniques for source identification and location; pulse energies, rise times, ringdown times, and amplitude distributions are commonly used signal parameters. Although these parameters often yield useful information about the nature and location of the AE source, they do have serious limitations. In particular, they tend to be less useful when the signal-to-noise (S/N) ratio is poor, when complex structural features produce confusing reflections, or when multiple sources are present. Unfortunately, the most important application areas for acoustic emissions often suffer badly from just these problems. Specifically, on-line AE-based NDE systems often must deal simultaneously with adverse S/N ratios, complex structures producing serious multipath/multimode interference, as well as unknown (often multiple) source locations.

Coherent methods (which make use of the phase information) provide some basis for hope in dealing with these complex problems. Unfortunately, linear methods to compensate for these effects have not proven to be particularly useful. In light of the role of mode conversion and multimode propagation in signal distortion, this is perhaps not surprising. Nonlinear methods offer considerably more promise and, in fact, have been used with much success in dealing with similar problems in other fields. Homomorphic deconvolution especially has been used with considerable success to eliminate multipath and multimode distortion in seismic, speech, and audio processing applications where S/N ratios are relatively good. A series of potentially useful homomorphic signal processing approaches were developed under Phase I of this contract and are described in detail in the Phase I final report. This report will focus on the experimental evaluation of the potential use these techniques may ultimately find in practical acoustic emission problems and, in particular, evaluate their

potential for compensating for transducer and structural resonances. Reliable multimode and multipath compensation, if achievable, can provide the technical basis for automated on-line AE monitoring of complex structures for cracks; if the structural complexities can be reduced or eliminated, it may be practical to monitor crack growth rates, to carry out crack severity assessments, and to locate cracks.

In other situations, particularly when the S/N ratio is poor, homomorphic methods, especially those based on cepstral techniques, are less useful and suffer from problems with accurate phase unwrapping. Under these conditions, adaptive nonlinear methods may be very applicable. The experimental work reported here focused on the deconvolution problem for situations where a relatively good S/N ratio was available.

### **1.3 Sensor Evaluation**

The most commonly used AE transducers—resonant piezoelectric transducers—seriously distort the microstructure of AE signals through transducer ringdown. Although the nonlinear signal processing approaches outlined above will to a large extent compensate for those problems, a better solution would be to use a transducer without massive intrinsic phase distortion. In this project we studied the performances of both the resonant transducers and the conical transducers. The conical transducers used were commercially available piezoelectric sensors based on ideas developed and demonstrated by Eitzen et al.<sup>1</sup> at the National Bureau of Standards. They have relatively flat frequency response up to 1 MHz. Such improved frequency response is achieved by reducing the contact area of the active element, by creating a smooth acoustical impedance transition to the backing material, and by increasing the backing material volume. Because of the reduced contact area, however, the detected signals are much reduced in amplitude compared to the resonant transducers, and very poor S/N ratios can be a serious problem in practical experimental situations.

### **1.4 Experimental Confirmation**

Finally, as the major element of the Phase II contract effort, a series of experiments of gradually increasing complexity were planned and performed to confirm and evaluate the effectiveness of the signal processing algorithms and techniques developed. The experiments were carried out by General Electric at the Research and Development Center and at the Materials and Processes Laboratory, in Schenectady, N.Y., under the joint direction of the co-investigators.

### **1.5 Summary**

This project represents a radically new approach to the acquisition and analysis of AE signals. The approach is based on the use of new signal processing methods and sensors and focuses on the removal of multipath and multimode distortion from the signals.

## Section 2

### THE THEORY REVISITED

The analytical basis for this work is described in detail in the Phase I final report, and will not be repeated here. It should be noted, however, that in the course of carrying out the experimental portion of Phase II, it was discovered that some of the theoretical results developed in Phase I of this project needed to be extended, and, in some instances, modified. Some of these changes came about in the course of analyzing the experimental data and in learning more about actual experimental uncertainties. Other changes arose out of a more careful examination of the physical assumptions behind the application of these techniques to the AE problem. The topics modified or added include exponential weighting, bandpass mapping, and the proper choice of the time distribution function in analyzing the multiple-event waveforms. Detailed descriptions of these changes, and the reasons behind them, are outlined below.

#### 2.1 Generalized Time Distribution Function for Multiple-Event Waveforms

As was shown in the Phase I report, the measured signals  $y(t)$  in acoustic emission experiments can be written in the form

$$y(t) = \left\{ \left[ \sum_i a_i h_{pi}(t-\tau_i) \right] * h_r(t) \right\} \cdot g(t) + n(t) \quad (1)$$

where

\* = the operation of convolution

$h_r(t)$  = the ringdown impulse response due to the boundary reverberation and the transducer response

$h_{pi}(t-\tau_i)$  = the AE impulse pulse emitted at the source at time  $\tau_i$

$g(t)$  = the measurement time gate

$n(t)$  = the background

The duration of the measurement time gate  $g(t)$  is assumed to be much longer than those of  $h_{pi}$  and  $h_r$  and therefore will be omitted from the rest of this report. This point will be treated in more detail in Section 4.2.

#### 2.2 Single-Event Approximation

Single events represent a much simpler case, and in the case of a solitary event, Equation 1 reduces to

$$y(t) = h_{pi}(t) * h_r(t) + n(t) \quad (2)$$

If the noise component  $n(t)$  is negligible, Equation 2 can be treated by the conventional homomorphic deconvolution technique to recover the signal  $h_{pi}(t)$ . Under this approach, the measurement  $y(t)$  is transformed to cepstral domain by the homomorphic operation  $IFT(\text{LOG}(\text{FT}(y(t))))$ . This composite operation converts the convolution operation into the addition operation. The cepstrum of  $y(t)$  is the sum of

those of  $h_{pi}(t)$  and  $h_r(t)$ , and hence the two cepstra can be separated by linear filtering.

This technique will work especially well when the cepstra of  $h_{pi}(t)$  and  $h_r(t)$  are well separated in cepstral period. In AE measurements, these conditions typically are met; typical AE  $h_{pi}$ s are narrow impulses with durations on the order of a few microseconds, while the  $h_r$ s have substantial low-frequency energy components and are much longer in duration. Since the cepstrum of a delta-like function is also a delta-like function, the cepstra of the  $h_{pi}$ s are concentrated near the origin in the cepstral domain.<sup>2,3</sup> In general the cepstrum of  $h_r$  is more spread out; so the criterion of separability of cepstra is approximately satisfied. One way to further reduce the contribution of the cepstrum of  $h_r$  near the origin is to convert  $h_r$  into a minimum-phase sequence by using the technique of exponential weighting, which will be described in detail in Section 2.4.

Once the cepstra of  $h_{pi}(t)$  and  $h_r(t)$  are separated, they can be operated on by the inverse homomorphic transform to yield either  $h_{pi}(t)$  and  $h_r(t)$ , respectively. The recovered pulse shape  $h_{pi}(t)$  can be analyzed for features associated with the particular class of events characterizing the source.

## 2.3 Multiple Events

In general, AE events do not occur singly, but rather in associated bursts; in some structures, the events may be easily visible as single pulses, but in large reverberant structures, the individual pulses, extended by reverberation, will overlap and be difficult to distinguish from one another.

### 2.3.1 Multiple Events with Occasional Single Events

On the other hand, provided that the AE events are associated with a localized crack formation and are statistically stationary, and further provided that a single reverberated AE waveform is available for analysis, the derived ringdown function  $h_r(t)$  can be used to dereverberate the multiple-event waveforms (if they occur) from the same source, through simple Fourier inversion. These dereverberated waveforms would reveal the original impulse time sequence of the overlapping individual events, and hence similar data from another transducer located at a different spatial location could be analyzed jointly via cross-correlation techniques to yield information on the time of arrival and hence location of the events. Similarly, the rate of occurrence of these waveforms could be easily determined and could be used in estimating the activity strength and otherwise characterizing the source.

### 2.3.2 Overlapping Multiple Events

More generally, in dealing with multiple-event waveforms, one has to resort to Equation 1. Compared to Equation 2, Equation 1 has two complications: the constancy of the impulse shape  $h_{pi}$  and the time distribution of the events. Assuming the different events all have the common pulse shape  $h_p(t)$  and only differ in their amplitudes  $a_i$ , then we have

$$y(t) = h_p(t) * p(t) * h_r(t) + n(t) \quad (3)$$

where  $h_p(t)$  is the AE pulse emitted at time  $t$ , and  $p(t) = \sum_i a(i)\delta(t - \tau_i)$  is the time

distribution of the events weighted with amplitudes corresponding to those of the different events.

The cepstrum of the measured signal  $y(t)$  is then the sum of the cepstra of  $h_p(t)$ ,  $p(t)$ , and  $h_r(t)$ . If the pulse train  $p(t)$  has been converted to minimum phase, it can be shown that its cepstrum is nonzero only for positive times greater than or equal to the interval between the first two arrivals,<sup>2</sup> and thus is well separated from that of  $h_p(t)$  if the intervals between the impulses are not too close to each other. In this case,  $h_p(t)$  can again be recovered by homomorphic deconvolution.

(If the different events do not have the same pulse shape, the signals  $\sum a_i h_{p_i}(t - \tau_i)$  contained in the measurement  $y(t)$  cannot be simplified to a convolution of the pulse train  $p(t)$  with a common pulse shape  $h_p(t)$ , as in Equation 3. Since it is difficult to estimate the cepstra of such complex signals, cepstral analysis is not practical.)

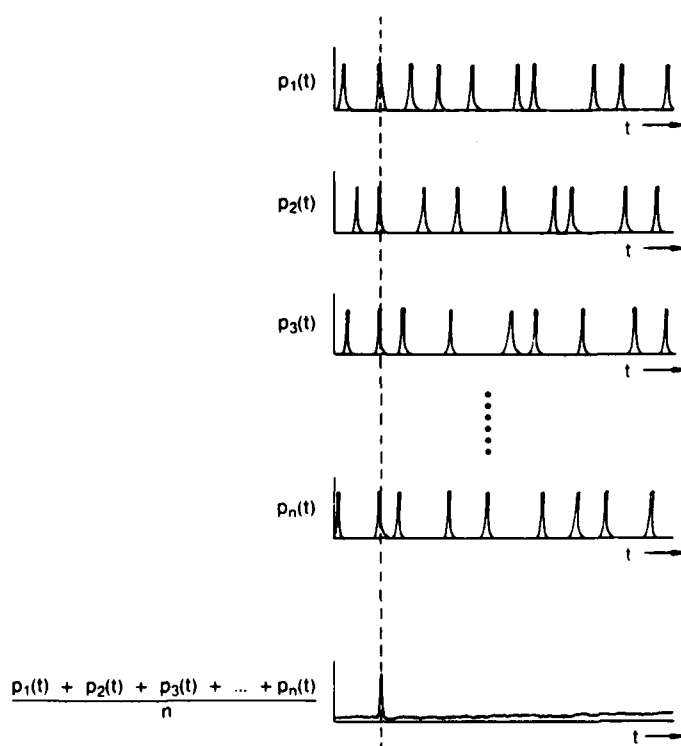
On the other hand, since the cepstra of  $h_r(t)$  and  $p(t)$  usually overlap,  $h_r(t)$  cannot be separated out. (Note that this result is distinctly different from the result reported in the Phase I study, where the assumption that an exponential distribution function holds for an ensemble of overlapping AE pulses lead to a contrary conclusion. A more careful analysis of the physical situation suggests that an exponential distribution is not appropriate for an overlapping ensemble of AE pulses.) Fortunately there is another way out of this dilemma. The method works on the fact that if  $p(t)$  is in fact a delta-like function, then the cepstrum of  $p(t)$  is also a delta-like function and thus is well separated from that of  $h_r(t)$ . One way to simulate this situation is by averaging a large number of waveforms with one of the events in each waveform aligned at about the same location. A simple and practical way of accomplishing this, for example, is to use a transient waveform capturing device set to trigger on a strong impulse. This will result in a series of captured waveforms where the first strong event in each waveform which exceeds the triggering level always occurs at about the same location  $t_0$  in the waveform records. Under these circumstances, the time distribution function can be approximated by

$$p(t) = \delta(t - t_0) + b \quad (4)$$

where  $b$  represents the averaged time distribution of the other events in the waveforms.

The procedure is illustrated in Figure 1. If the number of waveforms  $N$  used in the averaging is large enough,  $b$  will approach a uniformly flat distribution. The magnitude of  $b$  decreases with the number  $N$ , approaching zero as  $N$  tends to infinity. In this case  $p(t)$  is basically a delta function with a cepstrum concentrated at the origin, and thus will not interfere with the cepstrum of  $h_r(t)$ .

Note that if there were no aligning in the averaging, all the impulses would be distributed randomly in location, and  $p(t)$  would approach a uniformly flat distribution. Since the convolution of a flat distribution with any function is also a flat distribution, the averaged waveform  $y(t)$  is also a flat distribution and therefore contains no information at all.



**Figure 1. The delta-shaped time distribution function obtained by averaging the time distribution function of a large number of multiple-event waveforms with one of the events in each waveform aligned**

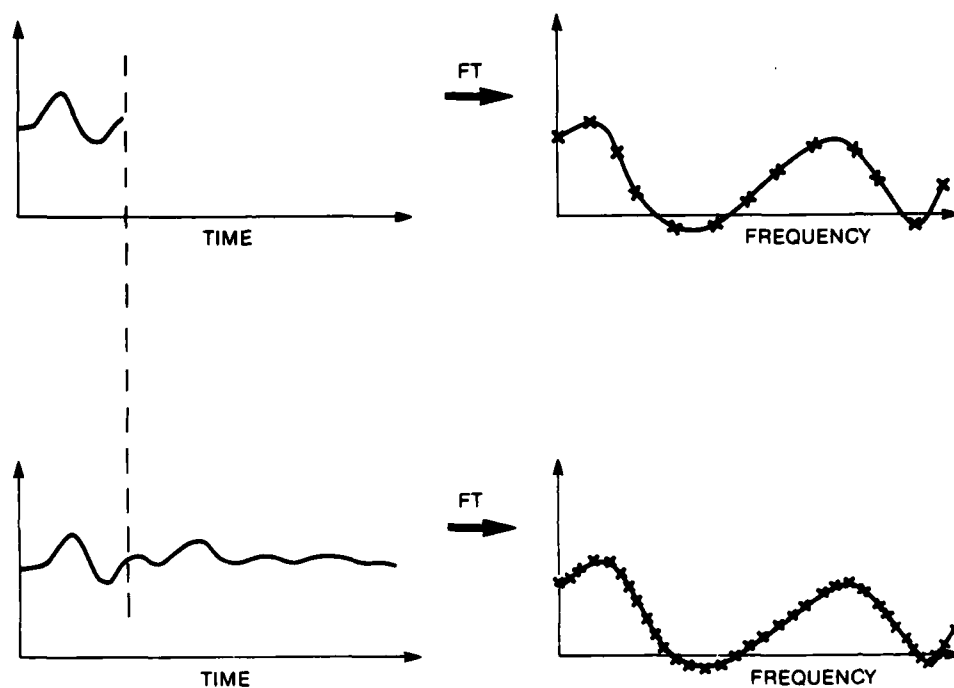
## 2.4 Exponential Weighting

One of the steps in cepstral analysis is phase unwrapping, which is the process of making the phase  $\theta$  of the Fourier transform,  $r \exp(i\theta)$ , of the time series waveform  $\{a(n)\}$  into a continuous function.<sup>2</sup> However, if some Fourier components are zero, their phase  $\theta$  would be undefined, and phase unwrapping would fail. This problem is especially serious with waveforms of long record length. The reason for this is that long record length corresponds to fine sampling in the Fourier domain,<sup>4</sup> and hence waveforms with long record length are more likely to pick up zero Fourier components than those with short record length. This phenomenon is illustrated in Figure 2.

Now the Fourier transform of a waveform corresponds to the  $z$ -transform of the function at the unit circle  $|z| = 1$ , where the  $z$ -transform  $A(z)$  of a sequence  $a(n)$  is defined as:

$$A(z) = \sum_{n=-\infty}^{\infty} a(n)z^{-n},$$

where  $z$  is a complex variable.



**Figure 2. The relation between record length and frequency sampling interval**

One way to get around the phase wrapping problem when some Fourier components are zero is by using exponential weighting.<sup>2</sup> The original waveform  $\{a(n)\}$  is weighted with the series  $\{\alpha^n\}$  before  $z$ -transformation:

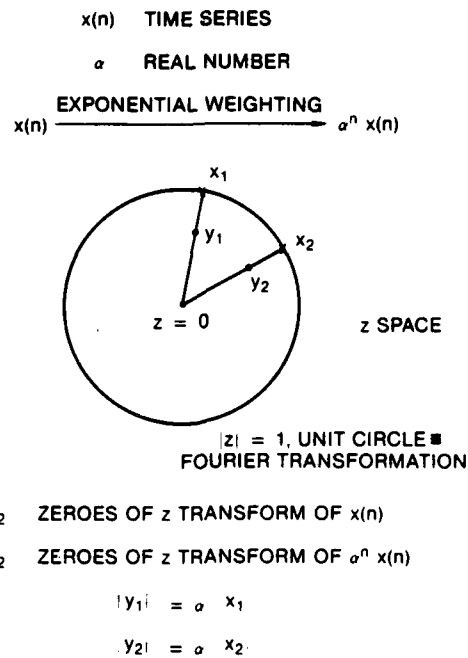
$$\{a(n)\} \rightarrow \{a(n)\alpha^n z^{-n}\}$$

where  $\alpha$  is a real number less than 1. Exponentially weighting a function by  $\{\alpha^n\}$  has the effect of scaling the magnitude of the zeroes of the  $z$ -transform of the function by  $\alpha$ , as illustrated in Figure 3. Thus the zeroes in the Fourier transform of a function can be removed by exponentially weighting the function before inputting to the homomorphic system, and phase unwrapping can be successfully carried out.

The use of exponential weighting can also improve the separability of the cepstra. If  $\alpha$  is small enough so that all the zeroes of the reverberation sequence are scaled to lie within the unit circle, the sequence is converted to a minimum-phase sequence. Under this condition the reverberation sequence will contribute to the cepstrum only for positive times greater than or equal to the interval between the first two arrivals in the sequence,<sup>2</sup> and hence the separability of the cepstra will be improved. In contrast, the structure of the cepstrum of a mixed-phase sequence is complicated, difficult to predict, and frequently unstable.

Note that exponential weighting is different from the usual windowing procedure used in Fourier transformation, which is an approximation as far as convolution is concerned. In general, convolution is not conserved under multiplication by a windowing function: the product of a windowing function,  $w(n)$ , with the convolution of





**Figure 3. Removing zeroes in  $z$ -transform by exponential weighting**

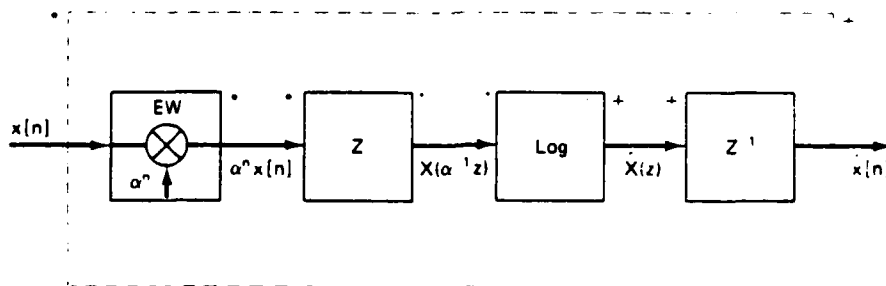
two functions,  $f(n)$  and  $g(n)$ , is different from the convolution of  $w(n)f(n)$  and  $w(n)g(n)$ :

$$w(n) \left[ f(n) * g(n) \right] \neq \left[ w(n)f(n) \right] * \left[ w(n)g(n) \right]$$

Convolution will be conserved, however, if  $w(n)$  is in the form  $\alpha^n$ :

$$\alpha^n \left[ f(n) * g(n) \right] = \left[ \alpha^n f(n) \right] * \left[ \alpha^n g(n) \right]$$

The output from the inverse homomorphic system is deweighted by the sequence  $\{1/\alpha^n\}$ . The modified homomorphic characteristic system including exponential weighting is illustrated in Figure 4.



**Figure 4. Modified homomorphic characteristic system including exponential weighting**

## 2.5 Bandpass Mapping

If the signals one wishes to analyze have intrinsic bandpass characteristics, as, for example, the data taken with resonant transducers, homomorphic analysis cannot be accomplished by means of full-band homomorphic systems. In fact, the analysis cannot be performed on the unit circle of the  $z$ -plane, since then the logarithm would become unbounded in the frequency bands with zero energy. Neither can it be performed off the unit circle, since the  $z$ -transform of such signals does not converge anywhere on the  $z$ -plane, but on the unit circle. Thus the exponential weighting procedure used to remove the zeroes of a signal off the unit circle cannot be applied to the class of bandpass signals. Since real data always contain some out-of-band noise, and can only have a countable number of zeroes on the unit circle, sometimes it appears possible to employ the exponential weighting procedure to remove such zeroes from the unit circle. Such an approach, however, is inherently ill-conditioned, and often leads to erroneous results.

One way to solve this problem involves a restriction on the domain of the logarithmic mapping to encompass only the passband of the input. This approach may be conveniently formulated in terms of a frequency scaling operation that shifts and stretches the signal's passband to occupy the entire frequency domain, as illustrated in Figure 5. The result of this operation is then a full-band sequence, which may be analyzed using full-band homomorphic systems. The analyzed output from the homomorphic system undergoes the inverse bandpass mapping. The characteristic system for bandpass homomorphic systems is illustrated in Figure 6.

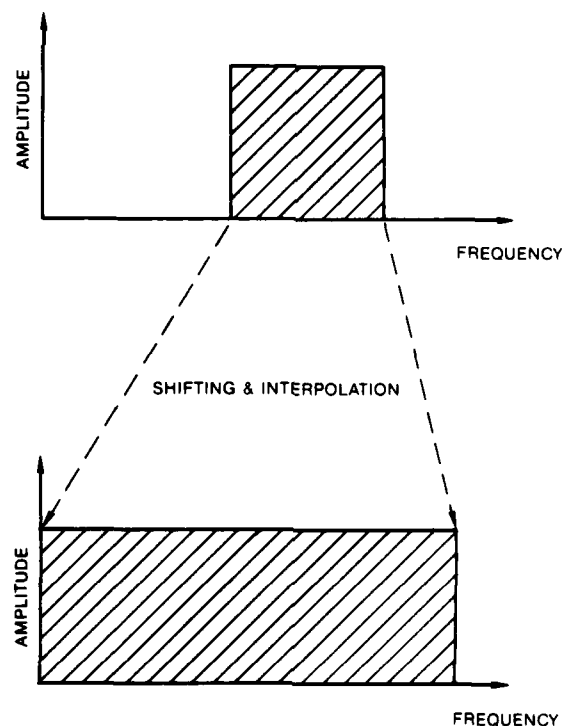
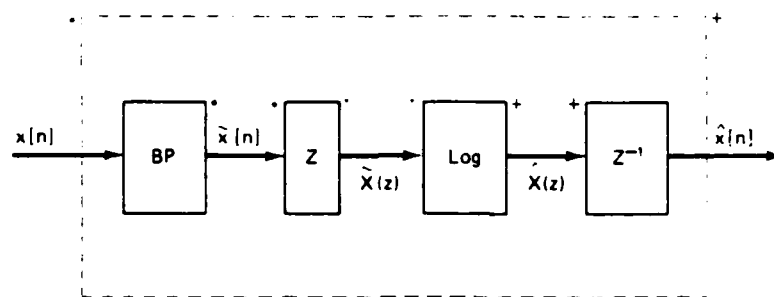


Figure 5. Bandpass mapping operation



**Figure 6. Characteristic system for bandpass homomorphic systems**

Bear in mind that bandpass mapping is not a panacea, but rather a method to remove the effects of the out-of-band noise in phase unwrapping. It does not add any information in the frequency bands with zero energy and may not produce particularly satisfying results, especially if the system under analysis is sharply resonant, with a relatively small range of useful, information-carrying frequencies.

### Section 3

## EXPERIMENTAL APPARATUS

A VAX-based experimental system was assembled to provide a suitable testbed for the acoustic emission experiments carried out under this contract. The system, shown in schematic form in Figure 7, provided a great deal of flexibility in the acquisition, validation, and analysis of the acoustic emission data. Data acquisition, for example, was carried out under VAX control; this approach gave the experimenters access to VAX-based tools to monitor the newly acquired data and verify its integrity during the course of the experiments. In addition, the VAX system incorporated an interactive digital signal processing software system, which was used extensively in the analysis of the data; the system was useful not only in the exploratory stages while appropriate algorithms were being developed, but also during the production analysis phase when validated experimental data were being batch processed. Subsequent sections of this report will provide details on the hardware and software used in these experiments.

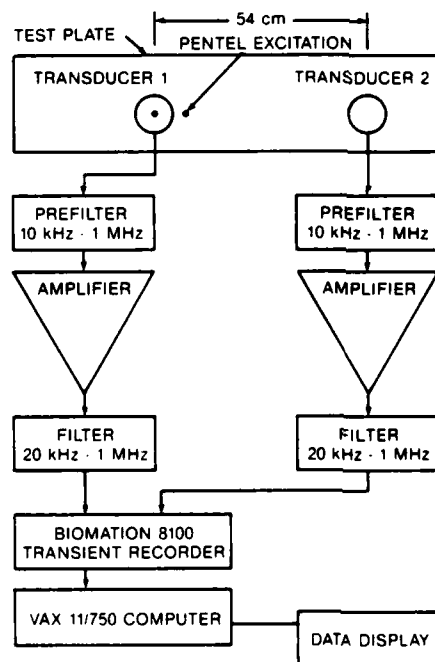


FIGURE 8 — BLOCK DIAGRAM OF THE DATA ACQUISITION SYSTEM.

Figure 7. Block diagram of the data acquisition system

### 3.1 Hardware

The hardware used in the acoustic emission experiments will be described in the subsequent sections essentially in the same order as the signal flow through the system. Figure 8 shows the experimental setup; the thick test plate can clearly be seen, as can the transducers, the Biomation transient recorders, the analog instrumentation, and the VAX 11/750 control/analysis computer in the background.



**Figure 8.** The experimental setup, showing the thick test plate, the Biomation transient recorders, the analog instrumentation, and VAX 11/750 control/analysis computer

#### 3.1.1 Specimens

A series of experimental specimens was prepared to provide a reasonable range of expected reverberation conditions; a key element in the design was the notion that it should be possible to increase gradually, and in a controlled manner, the physical complexity of the experiments, so that new conditions were added one at a time.

**3.1.1.1 Workpieces.** Two aluminum plates were selected to conduct the experiments. The first one was a relatively thick plate with dimensions of  $27 \times 5 \times 2$  in. The thick plate was chosen to allow clean near-field measurements without the confusing effects of reverberations from the back wall and other reflections. It functions in a near-field sense as an infinite half-sphere, and has known solutions for the expected AE due to a Pentel pencil lead break. The first plate is set up on a 2 in. thick foam pad to isolate

from ambient vibrations. The second plate had dimensions of  $27 \times 25 \times 1/4$  in. and was chosen to allow the investigation of the effects of plate bending vibrations and near-field resonance, as well as edge reflections. Only the thick plate was used for most of the experiments described in this report.

**3.1.1.2 C-Block Stress Corrosion AE Sources.** Notched aluminum C-blocks, under tensile stress and subject to a corrosive solution, were used to generate natural acoustic emissions. The notched C-blocks were clamped to the test plate and coupled with glycerine, so that good transmission of the generated AE pulses was achieved. A sodium chloride solution was subsequently used to induce stress corrosion cracking in the notched block. These sources provided reliable, natural AE signals, and could be easily controlled by small adjustments in the stress and/or salt solution so as to provide a wide range of natural AE signals, ranging from weak, individual pulses to nearly continuous trains of strong pulses.

### **3.1.2 Transducers**

Both conventional resonant transducers (Physical Acoustics Corporation, Model PAC R15) as well as conical low-resonance wide-band transducers (Industrial Quality Inc., Gaithersburg, Md.) were used to detect the acoustic events. The resonant transducers have a band frequency response between 100 and 300 kHz. The conical transducers have a flat frequency response up to 1 MHz.

### **3.1.3 Analog Signal Conditioning Equipment**

The signals from the transducers were first amplified using a Tektronix Model 502 amplifier with internal low-pass filter set at 1 MHz, or lower, depending on the nature of the sensing transducer. These signals were further processed using a Kronhite programmable filter (48 dB/octave/section) set up to low pass below 1 MHz. The output of the filter was then connected to the Biomation transient recorder system. A Nicolet digital Explorer scope was used for a visual display of the signals prior to transmitting them into the computer.

### **3.1.4 Transient Waveform Recorders**

Biomation 8100 digital transient recorders were used to record and sample the acoustic emission signals. Using a single Biomation 8100, 2048 samples can be obtained in single-channel mode; in dual-channel mode, two parallel channels of 1024 samples can be obtained. The recorder operates at up to a maximum sampling rate of 100 MHz, if desired. For single-channel operation, a 2 MHz sampling rate results in the acquisition of nearly 1 ms of data. Longer records needed to verify the Phase I results were obtained by ganging together four of these systems in series. Using this arrangement, nearly 4 ms of continuous data could be acquired from a single AE event, if required. A VAX-Biomation interface board was designed and fabricated to provide both automated experiment control by the VAX, as well as a data path for the transmission of transducer signals from the Biomation recorder into the DEC VAX 11/750 computer for later postprocessing. The hardware and software for this data acquisition system were designed and put together during the duration of this contract specifically for these AE experiments.

### 3.1.5 Stress-Corrosion Event Monitor

A conventional AE transducer, in conjunction with a Dunegan/Endevco 920 Distribution Analyzer and a Tektronix 604 Monitor, was used to monitor the rate of arrival and the intensity of the stress-corrosion events. This information allowed us to categorize easily the state of the C-block stress corrosion AE source, as well as providing a useful reference standard against which to measure the new analysis methods under evaluation.

## 3.2 Software

A variety of software modules were employed or developed to serve the needs of the experimental effort. Software modules were developed or used for data acquisition, data validation, algorithm identification, homomorphic data analysis, feature extraction, and pattern recognition. Much of the data analysis was carried out using an interactive, interpretive, high-level digital signal processing language available from Signal Technology Inc. (Santa Barbara, Calif.) called ILS. "Recipes," consisting of DEC DCL command procedures interpretable by ILS, are provided for all the analysis carried out using ILS. In some cases, where ILS did not provide appropriate functional modules, new ILS procedures were developed to provide the needed functionality. In either case, the modules used will be functionally described below; source code, and documentation for those developed under this contract, can be found in the appendices.

### 3.2.1 Data Acquisition

Software was developed to interface the Biomation hardware with the VAX 11/750 computer and to provide some measure of validation for the experimental data. Additional software was developed to prepare the input data in a form compatible with the high-level signal processing language (ILS) which contains a variety of software modules for data analysis.

*3.2.1.1 VAX-Biomation Interface.* A data acquisition/control software system was put together to acquire data with, and control up to, four Biomation 8100 transient recorders with a VAX 11/750. The Biomation interface software is MACRO-based and controls a DR-11C parallel interface board. The interface permits VAX-based software control of the Biomation recorders. All of the Biomation front panel switches and controls can be set via software commands originating in the VAX. In use, the Biomation transient recorders are operated in parallel, with one of the recorders providing triggering synchronization signals for the others. The trigger delays were set so that the captured data records would overlap slightly. The captured data records were processed by a VAX FORTRAN program that carried out correlation calculations on the overlapped regions to verify that the record segments were correctly synchronized, making appropriate corrections if required. In addition, the overlapped regions were used to adjust the gain and offset on each record to compensate for record-to-record variations. Finally, graphics software permitted the experimenter to view the captured records, or any subset of them, prior to proceeding. Complete documentation on the system, including functional descriptions, schematics, and software listings can be found in Appendix A.

### 3.2.2 Analysis

The basic approach in developing analysis software was to carry out as much of the analysis as possible directly in the ILS (see Appendix B) digital signal processing language; ILS is an expressive language, licensed by Signal Technology Incorporated, and provides a flexible ensemble of digital signal processing primitives built into its structure. In addition, ILS provides many kinds of generalized support functions, including data editing capability, simulation software, graphics software, and extensive pattern recognition software. ILS graphics capability, for example, supports graphics primitives for either interactive or hardcopy display, along with the associated support functionality.

Many complex functions can be constructed by stringing together primitive functional modules into DCL command procedure recipes. These recipes are executable either on a stand-alone basis, or as new primitive modules in more complex primitives. All the recipes used in analyzing the data in this report are included, where appropriate, with the figures in Section 4 displaying processed data. Finally, it is relatively easy to add entirely new modules to the ILS system.

Perhaps the most important feature of ILS is that it is intended to be used by nonprogrammers; thus it allows a researcher familiar with digital signal processing techniques to explore new techniques very quickly. This approach allowed considerable flexibility, and proved to be a very effective way to analyze and process quickly the acoustic emission data acquired under this contract.

**3.2.2.1 Homomorphic Analysis.** A software package was developed under Phase I at Rensselaer Polytechnic Institute to carry out simulation studies on the application of homomorphic analysis to acoustic emission signals. This package was based on the routines published in the IEEE volume *Programs for Digital Signal Processing*.<sup>5</sup> After some changes and modifications, the package was installed in the VAX computer system and integrated with the ILS digital signal processing package as the XCP module. Besides altering the code to conform to the ILS functional conventions, two basic changes were made to the existing code. The first change consisted of adding variable exponent exponential weighting capability to convert the acquired waveforms to minimum phase sequences, which are analytically more stable than mixed phase sequences. (See Section 2.4 for a description of the theory behind exponential weighting.) The second change involved adding bandpass mapping, necessary to deal with the signals produced by narrow-band resonant transducers. (See Section 2.5 for a description of the theory behind bandpass mapping.) The source code and documentation for the XCP ILS module can be found in Appendix B.

**3.2.2.2 Feature Extraction and Pattern Recognition.** A feature extraction and pattern recognition package was put together to investigate the separability of the different groups of waveforms. The package was based generally on the pattern recognition software available in ILS. Since the ILS pattern recognition modules were originally designed for speech and speaker recognition, minor changes had to be made to some of the modules to make them suitable for this application. The process followed in using the pattern recognition package to analyze acoustic emission waveform data was the traditional training/test set approach. A set of waveforms is analyzed and used to train the system; subsequent sets are tested against the training set for statistical compatibility. Summarizing the process, each of the waveforms in the group is Fourier transformed to extract a number of Fourier components. These Fourier components



are then processed to yield the mean waveform, the covariance matrix, the Fischer linear discrimination matrix, and the eigenvalues and eigenvectors for each group of waveforms. Using these intermediate results, principal component analysis is carried out on sets of waveforms. In essence, principal component analysis projects the groups of waveforms onto a discriminant plane with maximum separation between the groups. A scatter plot is then made using the first two principal components, and the statistical properties of each of the groups can be computed. Finally, the mean separation between groups can be expressed in statistical terms and used to characterize the probability that a given test value is in fact a member of the training set. A complete description of the software modifications to the ILS modules can be found in Appendix B.

## Section 4

### EXPERIMENTAL RESULTS

The block diagram of the experimental layout, hardware, and associated electronics was shown in Figure 7 of the previous section.

A problem encountered very early in the analysis was that our available methods for recording and for carrying out analog-to-digital conversion of the data did not provide sufficiently long record lengths for our analysis. Initially, the data acquisition system utilized a single Biomation 8100 interface providing a pair of 1024-point-long records. For sampling rates of 10 MHz, this system provided records approximately 0.1 ms in duration, insufficiently long for reverberations to decay in the experimental fixtures we employed. Accordingly, several different hardware approaches for recording and analog-to-digital conversion were explored to try to identify a more suitable means of acquiring the large experimental data sets needed to confirm the results of Phase I. The first approach explored was to reduce the sampling frequency to 2 MHz and to use the Biomation recorder in single-channel mode, so that a single 2048-point record is obtained instead of a pair of 1024-point records. This technique yields a record duration of roughly 1 ms, which—although still too short to record the entire ringdown train—provided enough data to begin analysis.

Later the record length was increased fourfold to 8096 points by connecting four Biomation recorders in series, as described above in Section 3.1.4. At a sampling rate of 2 MHz, this arrangement gives a record length of about 4 ms, which was adequate for subsequent analysis.

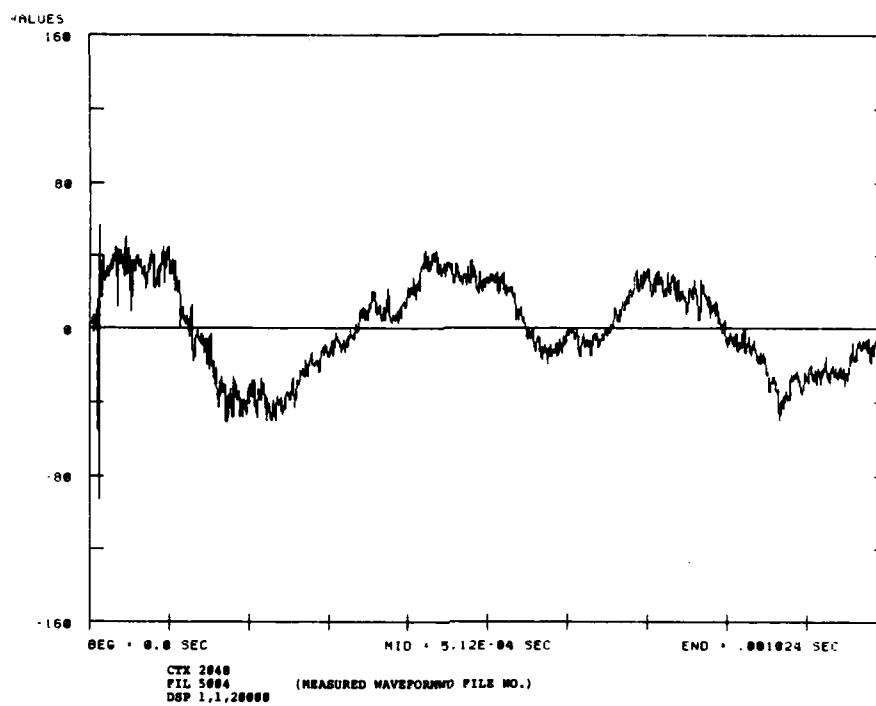
The command procedures for generating the ILS plots shown in Figures 9 and following are included in the illustrations. The DSP command at the end of most of the instruction sets displays the result in sampled data form.

#### 4.1 Calibration

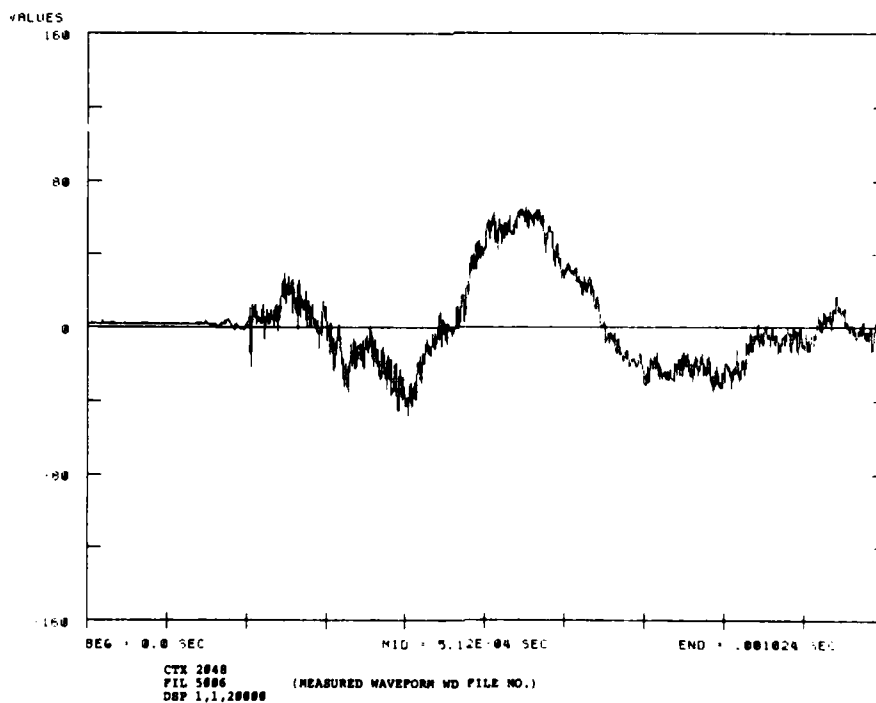
The breakage of Pentel pencil leads was employed as a calibration source. In the initial phase of the experiments, the 0.5 mm HB leads were broken manually.

Two typical lead-breaking waveforms are shown in Figures 9 and 10. They are two separate events taken with two of the wide-band, NBS-style conical transducers spaced about 54 cm apart on the  $27 \times 5 \times 2$  in. aluminum block. Pentel lead breaks were generated relatively close to transducer 1 at the location indicated. Signals from transducer 1 triggered the Biomation to accept data. The Biomation was set to the pretrigger mode. The data were low-passed at 1 MHz with the Kronhite filter, sampled at 2 MHz, for a 2048-point record. The waveform shown in Figure 9 is the output from transducer 1, while the waveform shown in Figure 10 is from transducer 2.

The time difference between the two pulses in Figures 9 and 10 is about 200  $\mu$ s, which is reasonably close to the value of 176  $\mu$ s calculated from the geometry of the setup, assuming the shear wave speed in aluminum to be 3.1 km/s. The waveforms exhibit low-frequency ringdown due to the reflections in the aluminum block.



**Figure 9. A typical lead-break waveform recorded by an NBS conical transducer**



**Figure 10. Another lead-break waveform recorded by an NBS conical transducer located 54 cm from the one used in Figure 9**

Homomorphic deconvolution was performed on the waveform in Figure 10. The cepstrum of the waveform is shown in Figure 11. The result of passing the cepstrum through the low-time window  $w(t)$ :

$$w(t) = \begin{cases} 1 & -9.5 \mu s < t < 9.5 \mu s \\ 0 & \text{otherwise} \end{cases} \quad (5)$$

and then inverse transforming is shown in Figure 12. An impulse type of function is recovered, but the result is not very clean: the ringing in the background is quite serious.

A large improvement was brought about by employing the technique of exponential weighting to minimize the effects of reverberation. The result of applying exponential weighting to the waveform in Equation 10 with  $\alpha = 0.9955$  and using the low-time window given by Equation 5 is shown in Figure 13. The recovered signal is much sharper and cleaner, and in fact compares favorably with the theoretical Pekar's solution, which is shown in Figure 14.

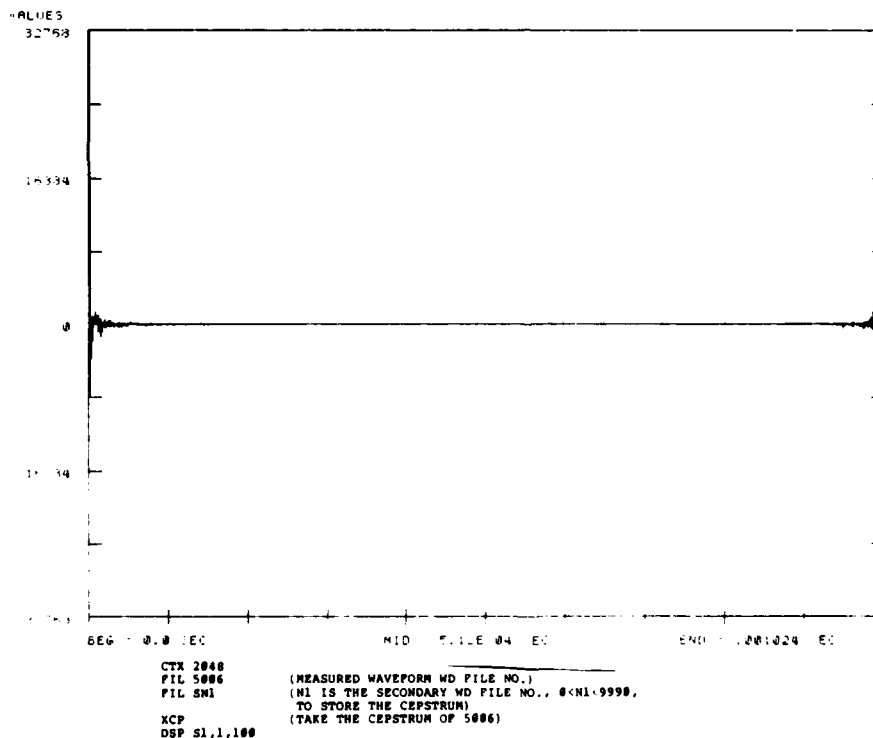


Figure 11. The cepstrum of the waveform in Figure 10

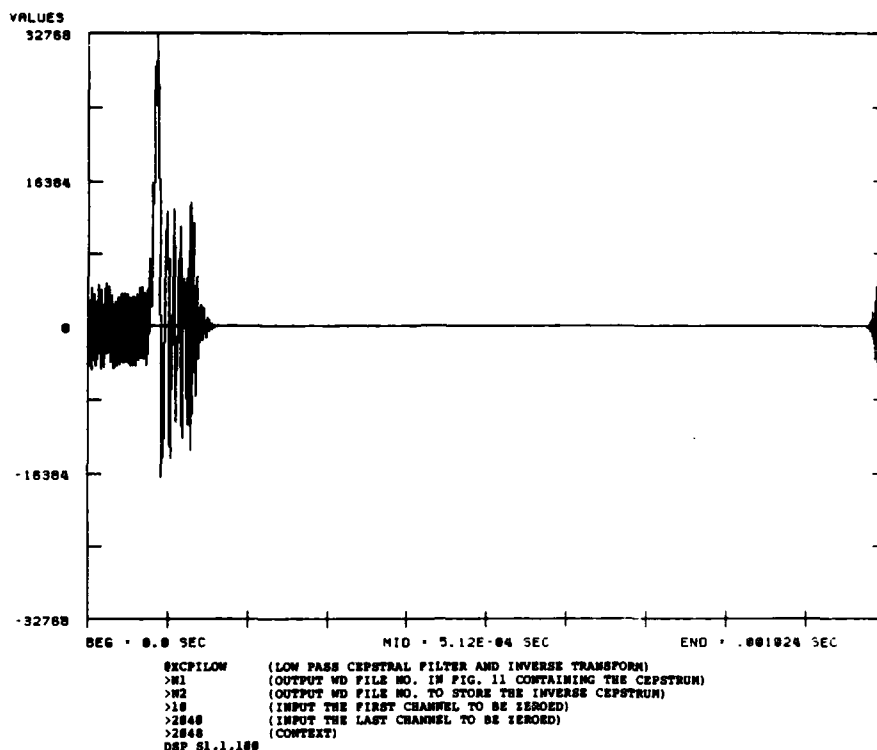


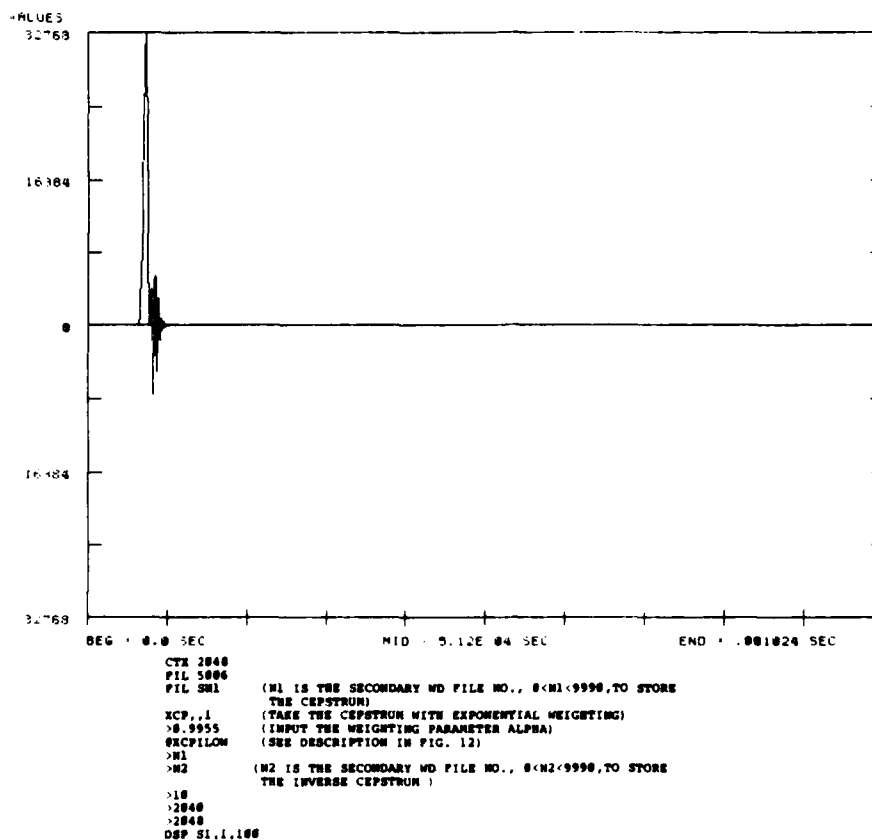
Figure 12. The result in the time domain of low-time pass filtering the cepstrum of the waveform in Figure 10

#### 4.2 Effect of Record Length

Another problem encountered was that the ringdown response of the medium could last substantially longer than the maximum record length, for as long, in fact, as several seconds under some circumstances. The record length of a waveform determines the fraction of ringdown captured, and hence the amount of leakage error in the Fourier transformation operation in homomorphic analysis. One would certainly expect the results to improve substantially when the record length is increased. As we shall see, however, if the effect of the record length is not significant enough, the improvement brought about by longer record length could easily be masked by other sources of errors.

The effect of the record length is demonstrated in the results shown in the sequence of Figures 15–26. Figures 15–18 are four lead-breaking waveforms captured with a conical transducer, the source being located at distances 45, 45, 30, and 15 cm, respectively, from the transducer. The signals were low-pass filtered at 1 MHz, sampled at 2 MHz, and are 8096 points in length. Each of these waveforms was transformed to the cepstral domain, then low-time filtered with the  $w(t)$ :

$$w(t) = \begin{cases} 1 & -4.5 \mu s < t < 4.5 \mu s \\ 0 & \text{otherwise} \end{cases} \quad (6)$$



**Figure 13. The improved result obtained through the use of exponential weighting;  
 $\alpha = 0.9955$**

and finally inverse-transformed to the time domain. The value of  $\alpha$  used in exponential weighting is 0.9985. Figures 19 through 22 show the results obtained in this way using all of the 8096 samples of the input waveforms in Figures 15 through 18. In comparison, the results obtained using only the first 2048 samples of the input waveforms are shown in Figures 23 through 26. Identical values of  $\alpha$  and  $w(t)$  are used in both cases. In comparing the two sets of results, although Figure 19 is much cleaner than Figure 23, the improvement caused by the longer record length is less obvious in the other three cases. From these results it may be concluded that the leakage error in Fourier transformation in going from 4.096 ms (corresponding to 8192 points at 2 MHz sampling rate) to 1.024 ms is not significant compared to the other sources of errors in the experiments and analysis. The lack of improvement with longer record length may also be due partially to some transient instabilities in the gain and noise characteristics of the four Biomatron recorders. At any rate, the record length of 1.024 ms provided by one Biomatron recorder at a sampling rate of 2 MHz appeared to be long enough for our purpose. The waveforms used in some of our later analyses were 1.024 ms in length.

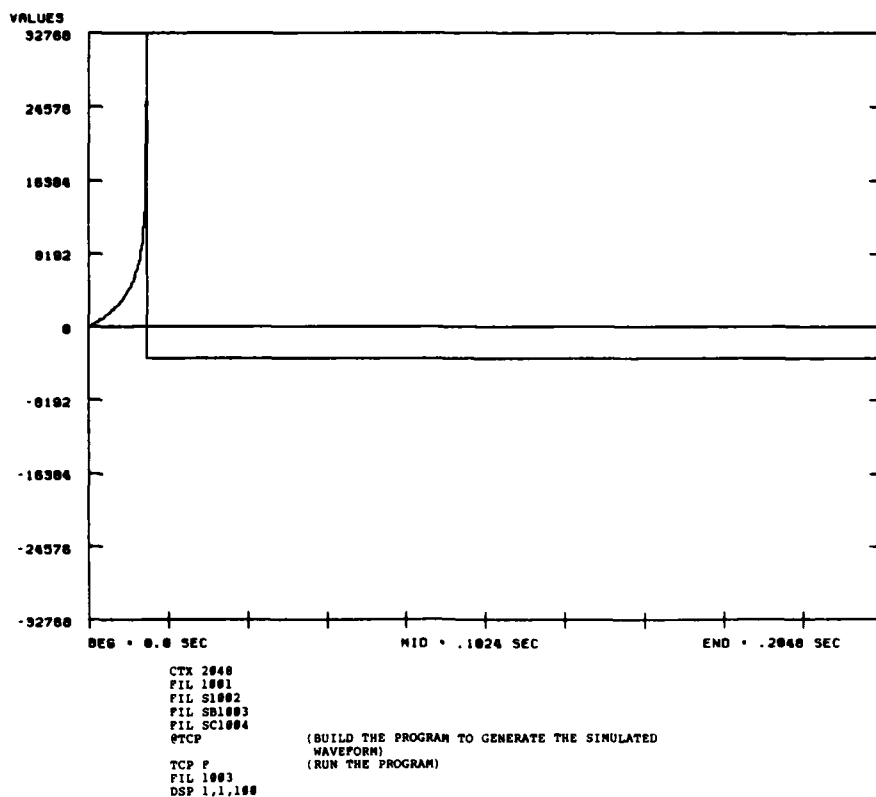


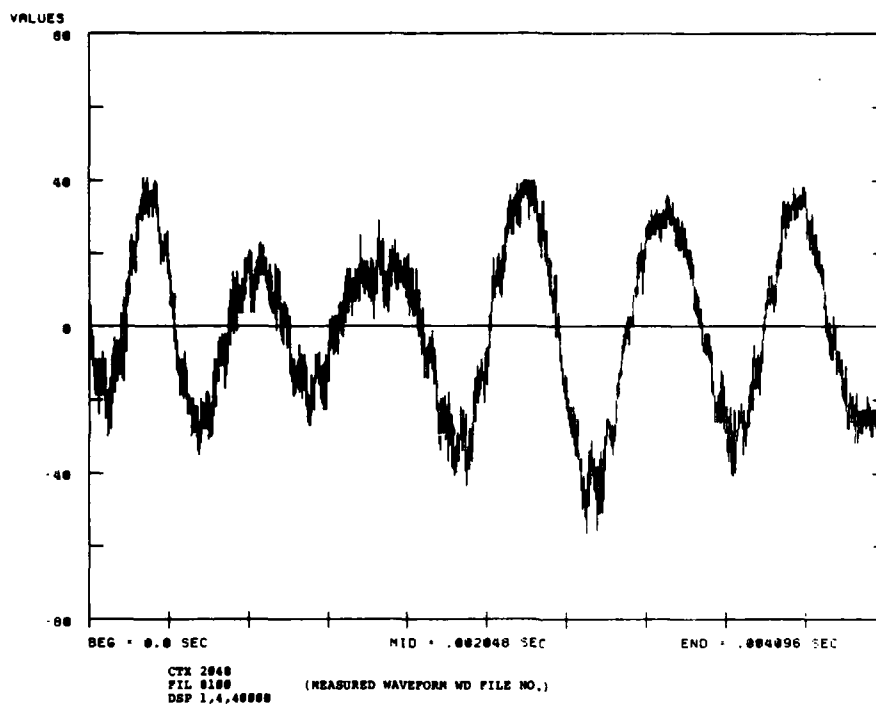
Figure 14. The theoretical seismic surface pulse according to Pekar

#### 4.3 Dependence on the Width of the Cepstral Filter

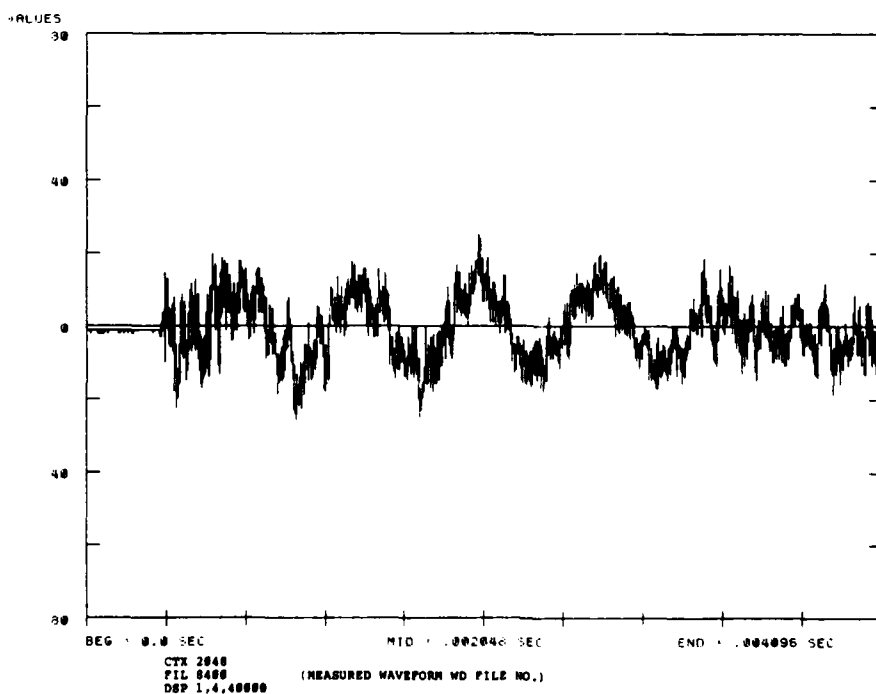
The ideal width of the cepstral filter used to separate out the signal impulses from the recorded waveforms was determined empirically. The results indicated that in most cases a filter centered around the origin in the cepstral domain with a width of about  $4.5 \mu\text{s}$  worked best. Two typical sets of results are illustrated in Figures 27–32. Figures 27–29 show the results of low-passing the cepstrum of the recorded waveform in Figure 10 using  $\alpha = 0.999$  in exponential weighting and cepstral filters of widths  $\pm 2.5$ ,  $\pm 4.5$ , and  $\pm 9.5 \mu\text{s}$ , respectively. Figures 30–32 are the results of applying the same procedure to the waveform in Figure 15 using  $\alpha = 0.9985$  in exponential weighting. In both cases, the cepstral filter with a width of  $\pm 4.5 \mu\text{s}$  yielded the best results.

#### 4.4 Dependence on Alpha

In the homomorphic analysis of single-event waveforms, exponential weighting converts the original time sequences  $h_p(n)$  and  $h_r(n)$  to the new ones  $\alpha^n h_p(n)$  and  $\alpha^n h_r(n)$ . The value of  $\alpha$  determines the distributions of the cepstra of the two exponentially weighted time sequences, and hence the results of the homomorphic deconvolution depend on the value of  $\alpha$  used in the exponential weighting. To study the extent of this dependence, homomorphic analysis was performed on the waveform in Figure 15 with the window  $w(t)$  in Equation 6 but with different values of  $\alpha$  in the



**Figure 15. A lead-break waveform recorded by an NBS conical transducer located 45 cm from the source**



**Figure 16. A lead-break waveform recorded by an NBS conical transducer located 45 cm from the source**



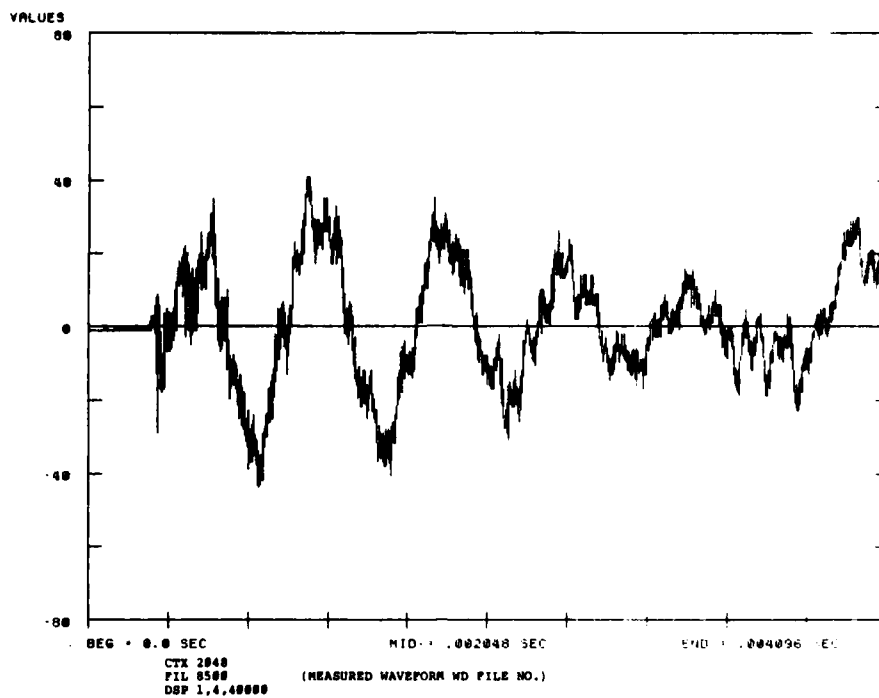


Figure 17. A lead-break waveform recorded by an NBS conical transducer located 30 cm from the source

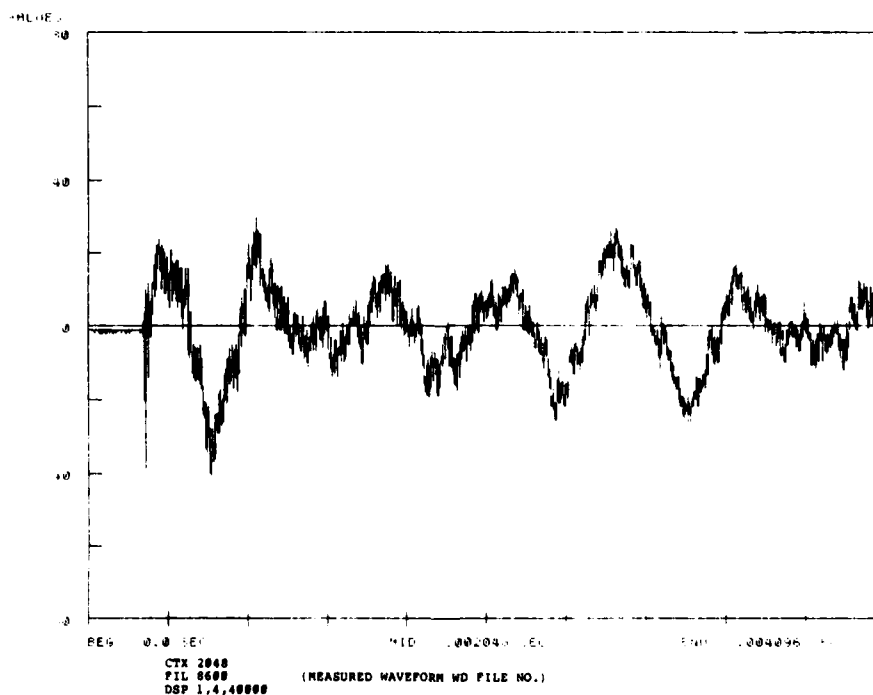


Figure 18. A lead-break waveform recorded by an NBS conical transducer located 15 cm from the source

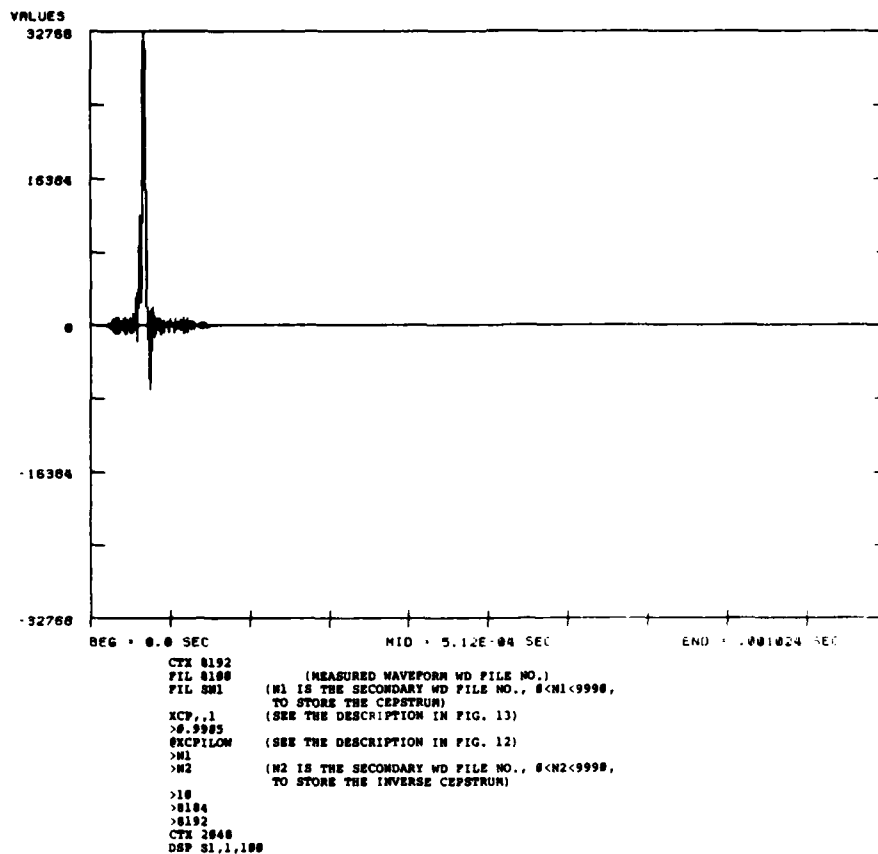


Figure 19. The result of low-time pass filtering the cepstrum of the waveform in Figure 15, using all of the 8096 samples as input;  $\alpha = 0.9985$

exponential weighting. The results corresponding to  $\alpha = 0.9985$ , 0.998, and 0.9975 are shown in Figures 33, 34, and 35, respectively. The results indicate that even a slight change in the value of  $\alpha$  affects the results significantly.

#### 4.5 Fourier Deconvolution

Initial efforts to use the transfer function obtained in the homomorphic deconvolution of one Pentel lead-breaking waveform to deconvolve other Pentel lead-breaking waveforms were met with little success. To understand the difficulties, the repeatability of the manually performed lead-breaking waveforms was examined in detail. It was found that the lead-breaking signals depend critically on (1) the precise angle the lead makes with the surface and (2) possible secondary Pentel tip impact after the lead break.

These findings were illustrated in Figures 36–38, which are manually performed lead-breaking waveforms captured with a conical transducer. Figures 36 and 37 illustrate the dependence of the signal on the angle between the lead and the surface. The waveform in Figure 36 was generated with an angle of about  $65^\circ$ , and the one in Figure 37 with an angle of about  $40^\circ$ . The waveforms are markedly different. It appears

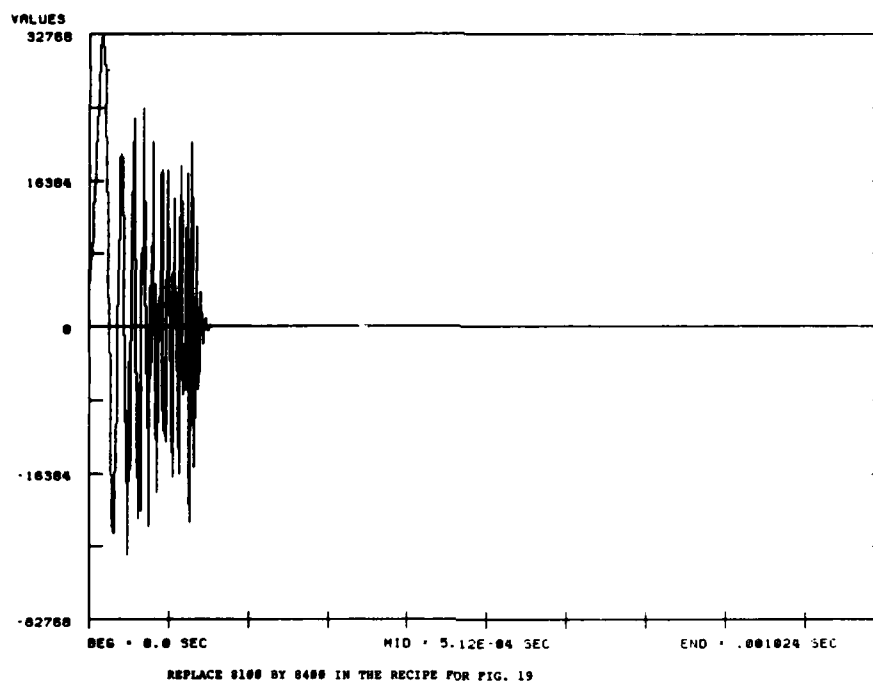


Figure 20. The result of low-time pass filtering the cepstrum of the waveform in Figure 16, using all of the 8096 samples as input;  $\alpha = 0.9985$

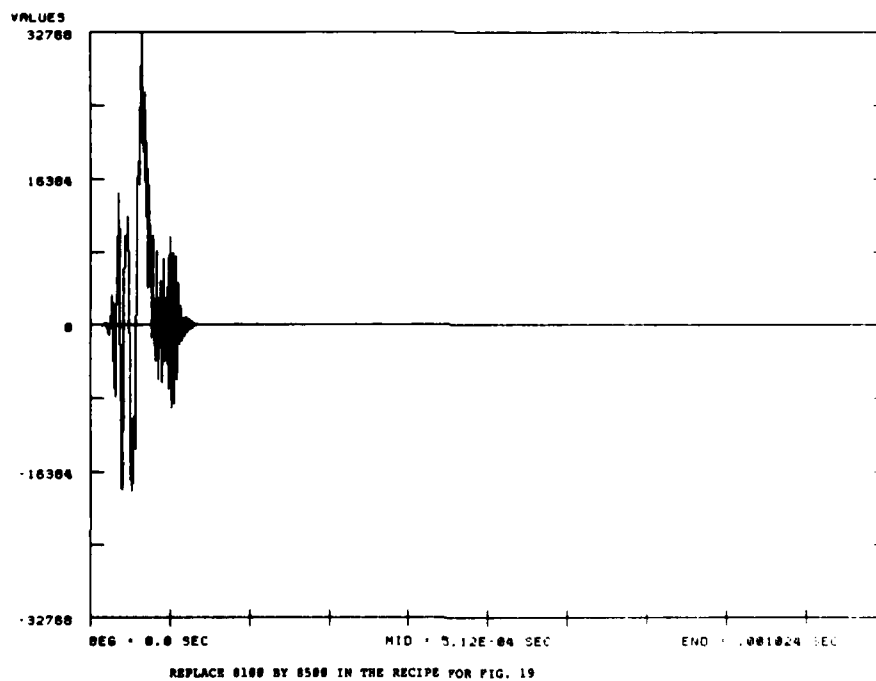


Figure 21. The result of low-time pass filtering the cepstrum of the waveform in Figure 17, using all of the 8096 samples as input;  $\alpha = 0.9985$

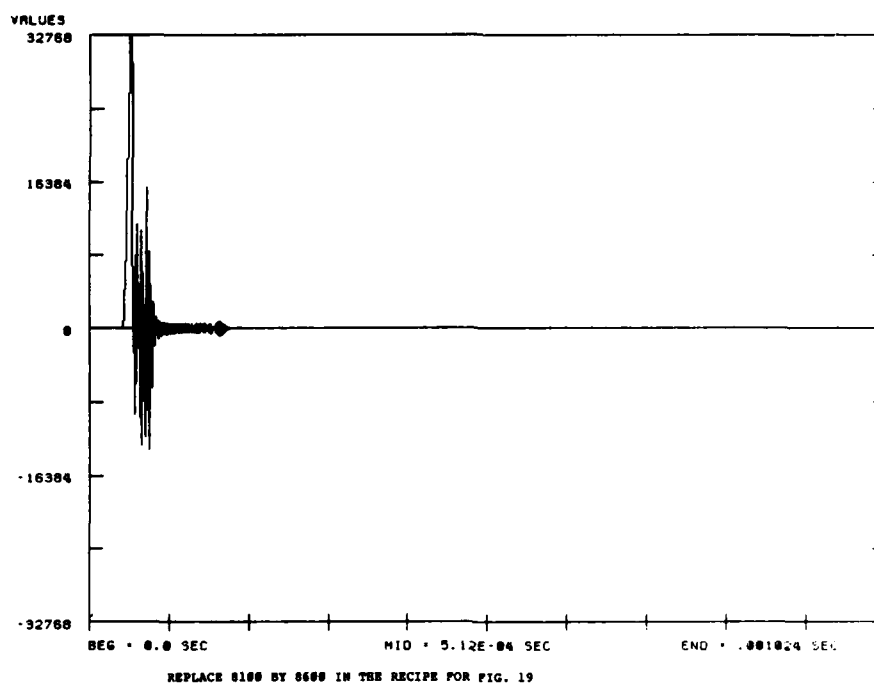


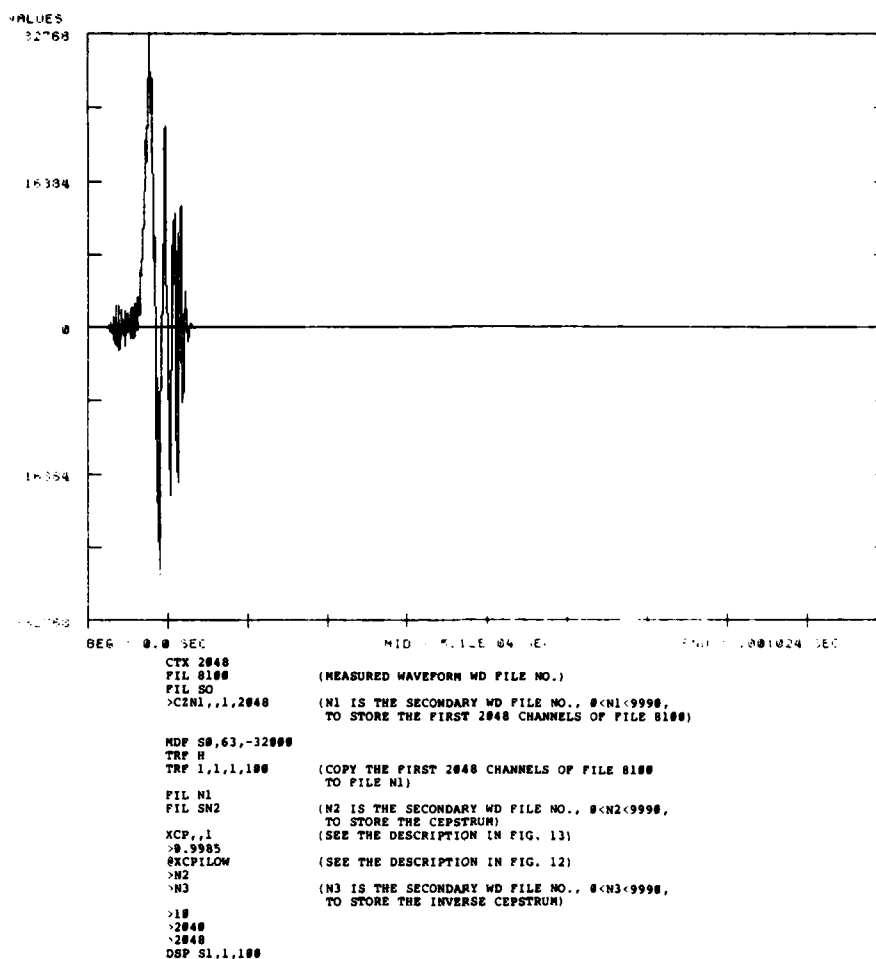
Figure 22. The result of low-time pass filtering the cepstrum of the waveform in Figure 18, using all of the 8096 samples as input;  $\alpha = 0.9985$

that the angle determines the amount of shear and transverse waves generated in each event. A shallow-angle ( $<45^\circ$ ) event contains relatively more low-frequency components and is relatively featureless. On the other hand, a steep angle ( $\sim 60^\circ$ – $70^\circ$ ) produces strongly featured signals with much modulation. A Pentel lead break with secondary Pentel tip impact at large angle is shown in Figure 38. The secondary impact generated additional AE activity in the midregion of the waveform.

In an effort to produce repeatable lead-breaking signals, a fixture for breaking leads in a well-controlled manner was acquired. The fixture allowed precise control over the angle of the lead during breakage and effectively eliminated problems with periodic secondary tip contact. Further studies on the feasibility of Fourier deconvolution were carried out using waveforms generated with the help of the fixture. It was found that the waveforms were highly repeatable. The repeatability is demonstrated in Figures 39 and 40, where 200 points of two such waveforms are shown near the initial impulses. These waveforms were recorded by a conical transducer and sampled at 2 MHz. Figures 41 and 42 show another 200 points in the later part of the two waveforms. Minor differences between the two waveforms are visible after a considerable delay, but the overall similarity is still very impressive.

Fourier deconvolution using these mechanically generated waveforms yielded encouraging results. Figure 43 shows the impulse response obtained by the following procedure:

- Averaging 33 of such waveforms (Pentel lead breaking by a mechanical device recorded by a conical transducer), each of the waveforms being 1024 points in length



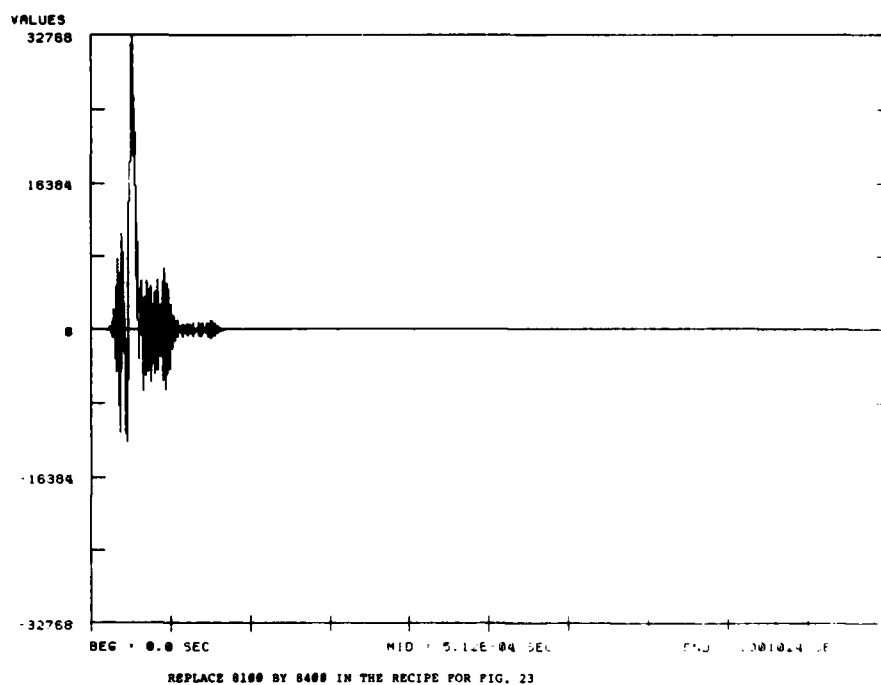
**Figure 23. The result of low-time pass filtering the cepstrum of the waveform in Figure 15, using only the first 2048 samples as input;  $\alpha = 0.9985$**

- Calculating the cepstrum of the averaged waveform,  $\alpha = 0.999$
- High-pass filtering the cepstrum, i.e., setting the low-time portion of the cepstrum to zero for  $|t| < 4.5 \mu s$
- Inverse-transforming the cepstrum to the time domain

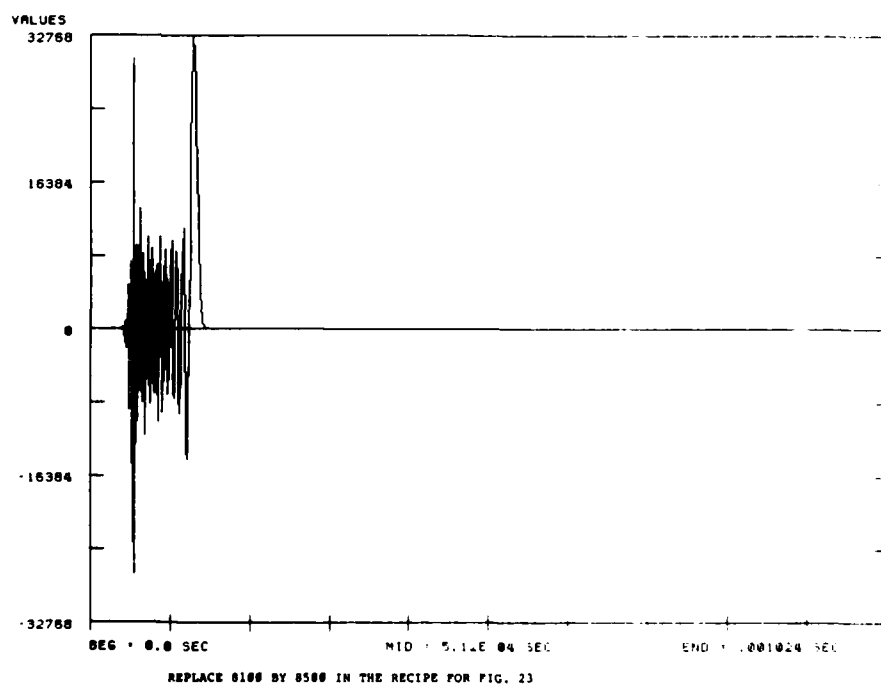
The result of deconvolving one of the 33 waveforms used in averaging, the one shown in Figure 44, by the impulse response in Figure 43 is illustrated in Figure 45. The result was improved by removing the high-frequency noise with the elliptical frequency filter shown in Figure 46 with  $-60$  dB cut-off at  $700$  kHz; the filtered result is shown in Figure 47.

The deconvolution results shown in Figures 45 and 47 were obtained under close to ideal conditions:

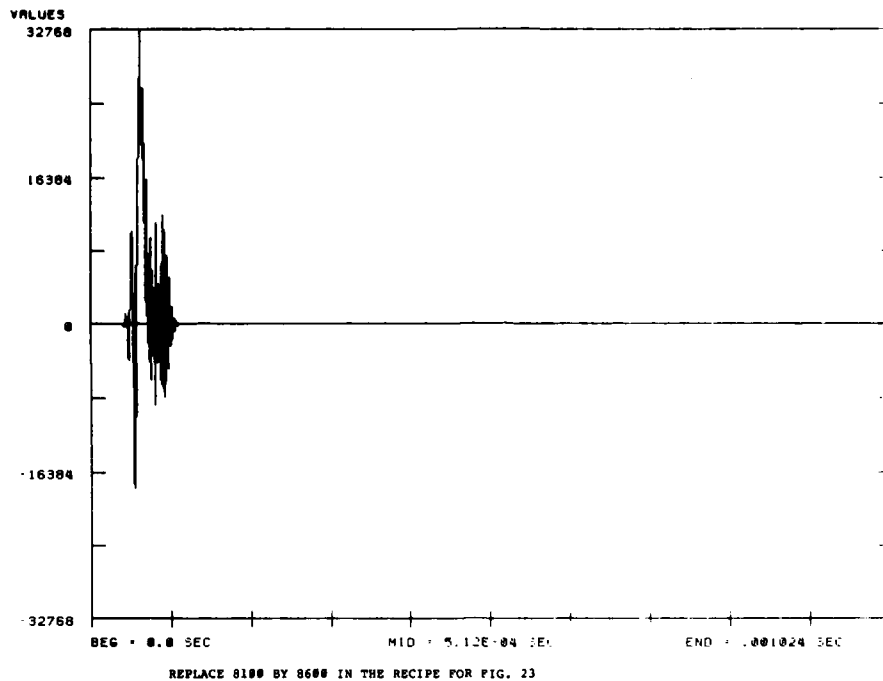
- Obtaining highly repeatable waveforms by breaking Pentel leads with a mechanical device



**Figure 24.** The result of low-time pass filtering the cepstrum of the waveform in Figure 16, using only the first 2048 samples as input;  $\alpha = 0.9985$



**Figure 25.** The result of low-time pass filtering the cepstrum of the waveform in Figure 17, using only the first 2048 samples as input;  $\alpha = 0.9985$



**Figure 26.** The result of low-time pass filtering the cepstrum of the waveform in Figure 18, using only the first 2048 samples as input;  $\alpha = 0.9985$

- Recording the waveforms with a wide-band conical transducer
- Using a relatively short record of 1024 points, where the similarity between waveforms is extremely good

In Figures 45 and 47, the presence of the impulse is obvious, yet the quality of the pulse shape does not appear to be good enough for identification purposes, even after low-pass filtering to remove high-frequency noise. These results indicate that Fourier deconvolution is too sensitive to the minor differences in the recorded waveforms to be of any value for event identification in practical situations.

#### **4.6 Effects of a Limited-Frequency Band**

Realistic acoustic emission events were generated using aluminum C-blocks subject to stress and a salt solution. It was found that the amplitude of stress-corrosion events generated in this way was about 60 dB lower than the amplitude from Pentel lead breaks; this finding can be seen clearly in the amplitude plot shown in Figure 48. Unfortunately, since the sensitivities of the conical transducers are much lower than those of the resonant transducers, the stress-corrosion signals could not be detected by the conical transducers at all. For this reason, resonance transducers were used for detection.

The data taken with the resonance transducers are band-limited between 100 kHz and 300 kHz. The power spectrum of one of the stress-corrosion waveforms captured with a resonance transducer is shown in Figure 49. As mentioned in Section 2.5, such data require bandpass mapping to suppress the effects of noise propagation in the

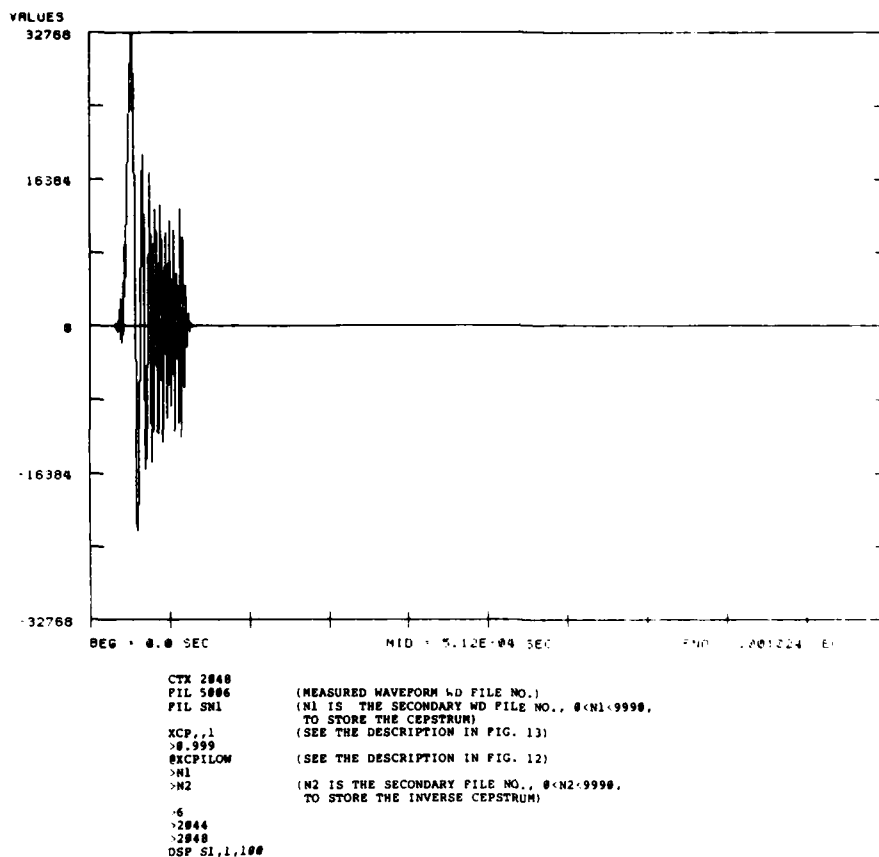


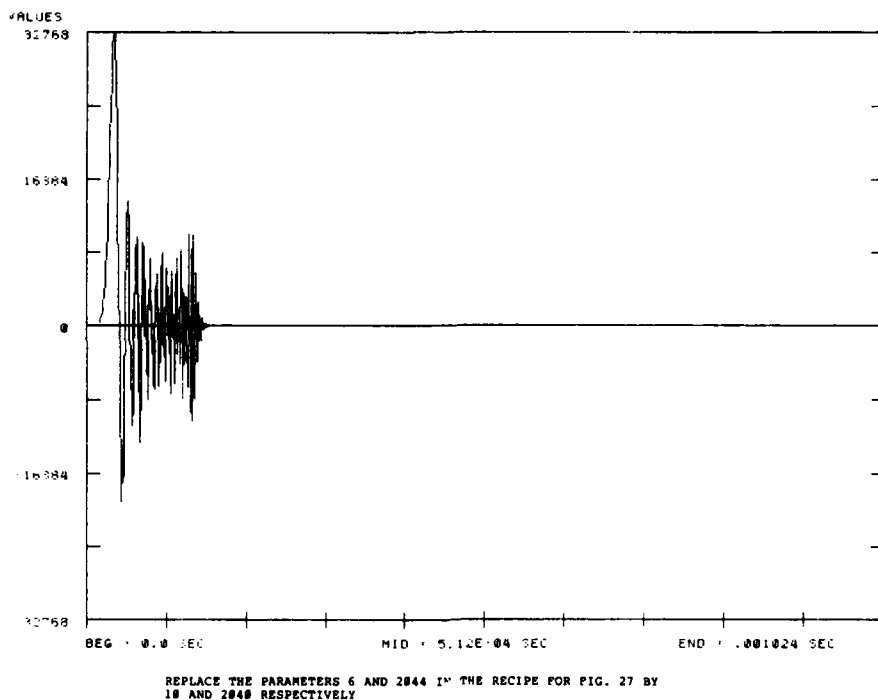
Figure 27. The result of low-time pass filtering the cepstrum of the waveform in Figure 10,  $\alpha = 0.999$ , width of cepstral filter =  $\pm 2.5 \mu s$

unoccupied frequency band in the process of phase unwrapping. Figure 50 shows the average of 31 stress-corrosion waveforms with their initial impulses lined up at the same location, as mentioned in Section 2.3.2. The waveforms were captured with a resonance transducer, were sampled at 2 MHz, and were 2048 points in length. The result of low-pass filtering the cepstrum of the averaged waveform without bandpass mapping is shown in Figure 51. The value of  $\alpha$  used in exponential weighting is 0.999, and the low-time cepstral filter  $w(t)$  is given by

$$w(t) = \begin{cases} 1 & -4.5 \mu s < n < 4.5 \mu s \\ 0 & \text{otherwise} \end{cases}$$

Figure 52 shows the improved result brought about by bandpass mapping between 100 kHz and 300 kHz. Even with bandpass mapping, there is still severe ringing and broadening in the impulse signal caused by the lack of information in the frequency band with zero energy.



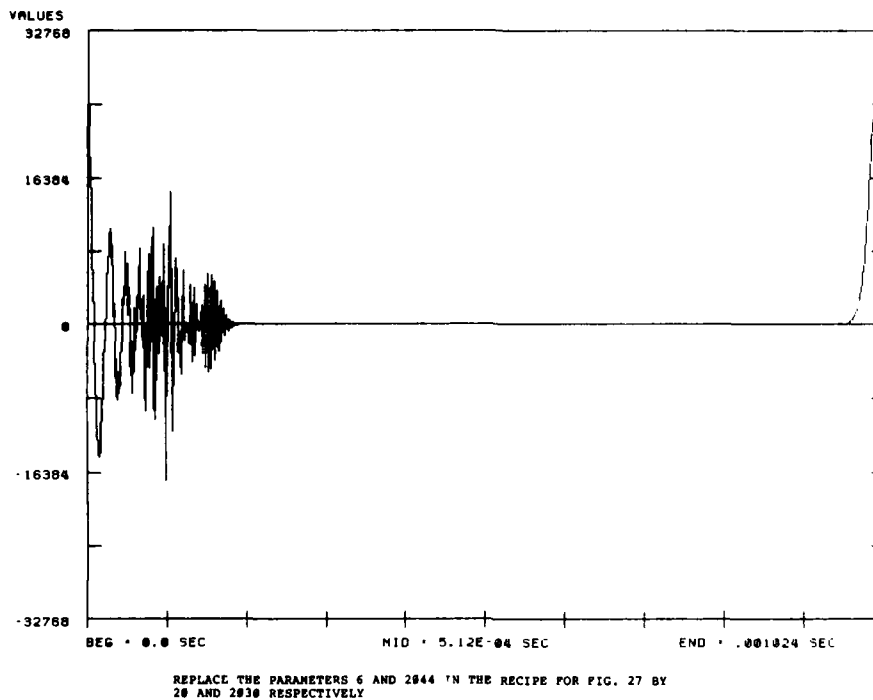


**Figure 28.** The result of low-time pass filtering the cepstrum of the waveform in Figure 10,  $\alpha = 0.999$ , width of cepstral filter =  $\pm 4.5 \mu s$

Figure 51 indicates that the band-limited nature of the waveforms recorded using resonance transducers gives rise to severe distortions in the impulses obtained by low-pass filtering the cepstrum of the averaged waveform in Figure 50. Bandpass mapping cleans up the result considerably by suppressing the error propagation in the unoccupied frequency regions, but severe ringing and broadening is still present as shown in Figure 52. This ringing and broadening is due to intrinsic limitations of the band-limited nature of the data—resulting from the use of resonance transducers—and therefore cannot be removed by signal processing.

#### 4.7 Pattern Recognition

The quality of the homomorphically deconvolved signals obtained above does not give one much hope for accomplishing source identification. This observation was confirmed in the following pattern recognition analysis using two sets of waveforms from different sources, with nine waveforms in each set. In one set are waveforms generated by rubbing sandpaper on the edge of one end of the aluminum block, while the ones in the other set were generated by rubbing the sandpaper on the face of the same end of the block. The waveforms were sampled at 2 MHz and are 2048 points in length. Each of the waveforms was homomorphically deconvolved to recover the low-time component in the cepstral domain, using a weighting of 0.9985 and a cepstral cutoff time of  $\pm 4.5 \mu s$ . These deconvolved signals were Fourier analyzed, and the 20 Fourier coefficients from 976.56 Hz to 594.7 kHz at 31.25 kHz intervals were extracted. Principal component analysis<sup>6</sup> was performed on these 20 Fourier coefficients of



**Figure 29.** The result of low-time pass filtering the cepstrum of the waveform in Figure 10,  $\alpha = 0.999$ , width of cepstral filter =  $\pm 9.5 \mu s$

the two sets of 9 waveforms in each set using the software in the ILS signal analysis package. The first two principal components are plotted in Figure 53. (See Appendix C for a description on pattern recognition and examples of command procedures for Figures 53 and following.) The results in Figure 53 indicate the two sets of waveforms cannot be distinguished from each other. In comparison, the first two principal components of the raw data without cepstral filtering are plotted in Figure 54. The two sets of waveforms become less separated after the cepstral filtering.

Similar results were obtained when the above procedure was repeated with an additional set of nine lead-breaking waveforms. The plots of the first two principal components with and without cepstral filtering are illustrated in Figures 55 and 56. Again, cepstral filtering failed to improve the separability of the three sets of waveforms.

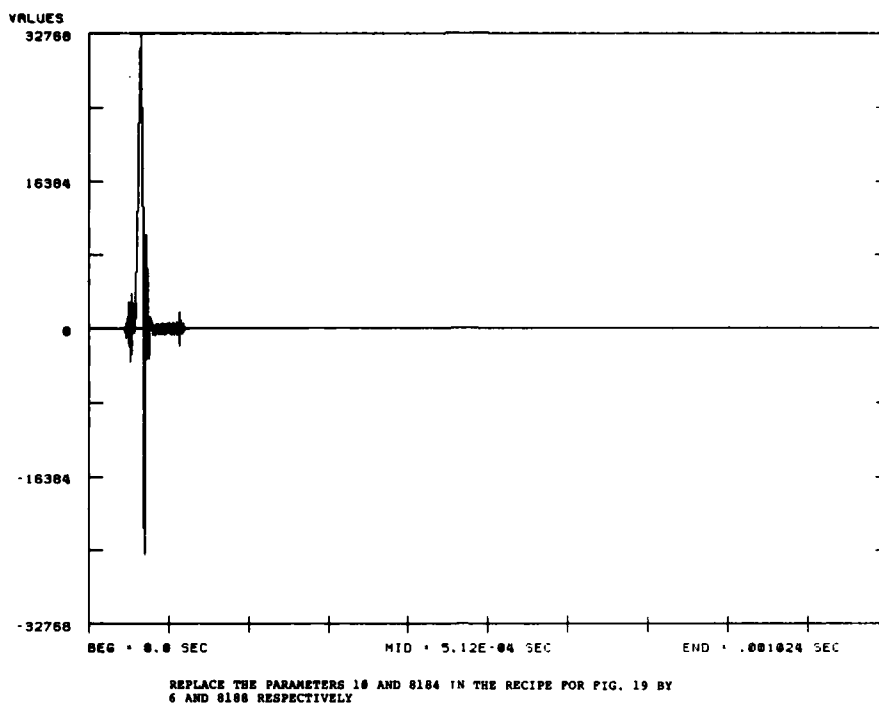


Figure 30. The result of low-time pass filtering the cepstrum of the waveform in Figure 15,  $\alpha = 0.9985$ , width of cepstral filter =  $\pm 2.5 \mu s$

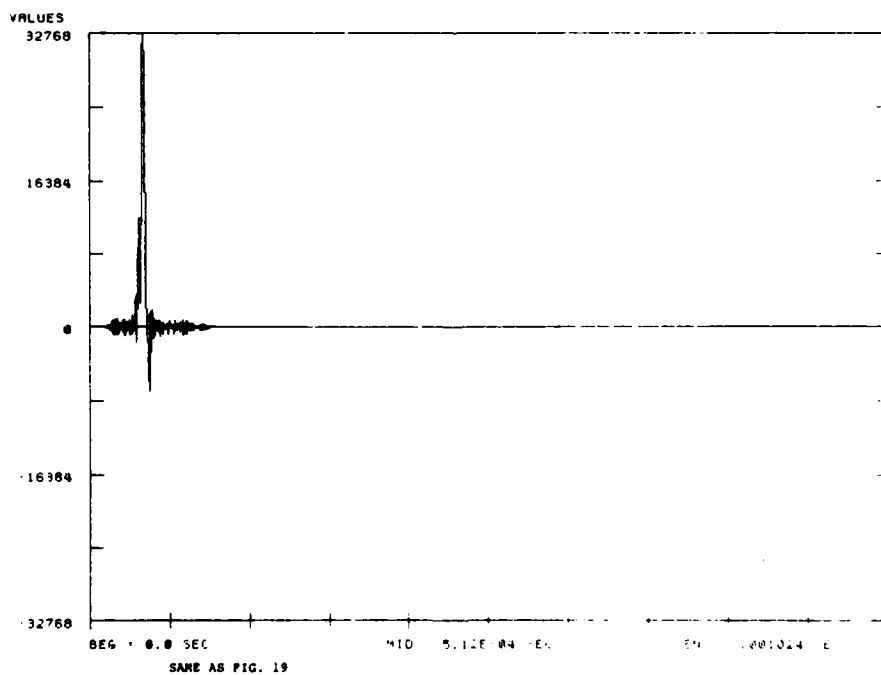
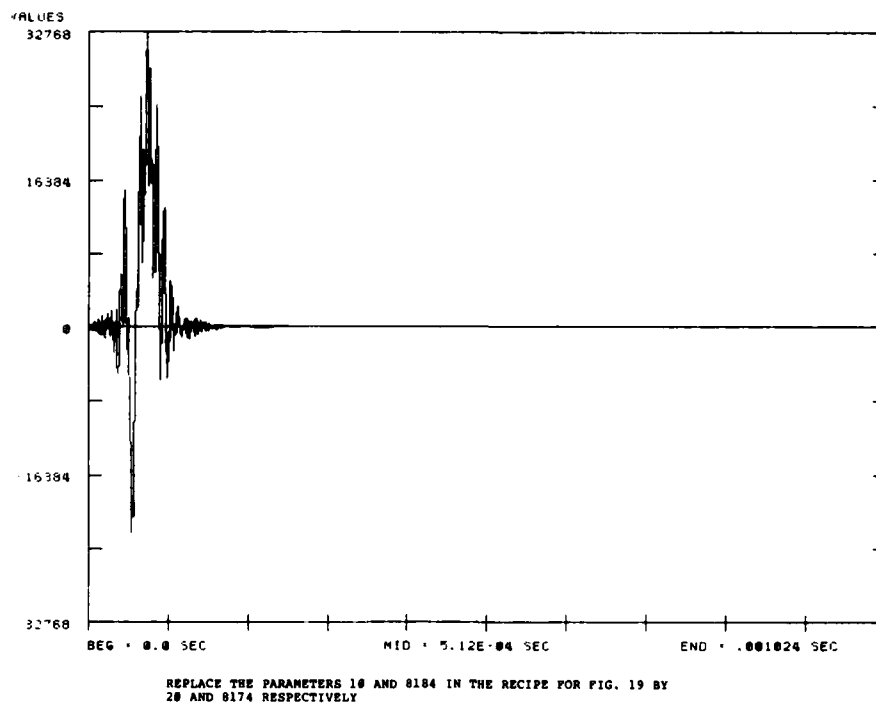
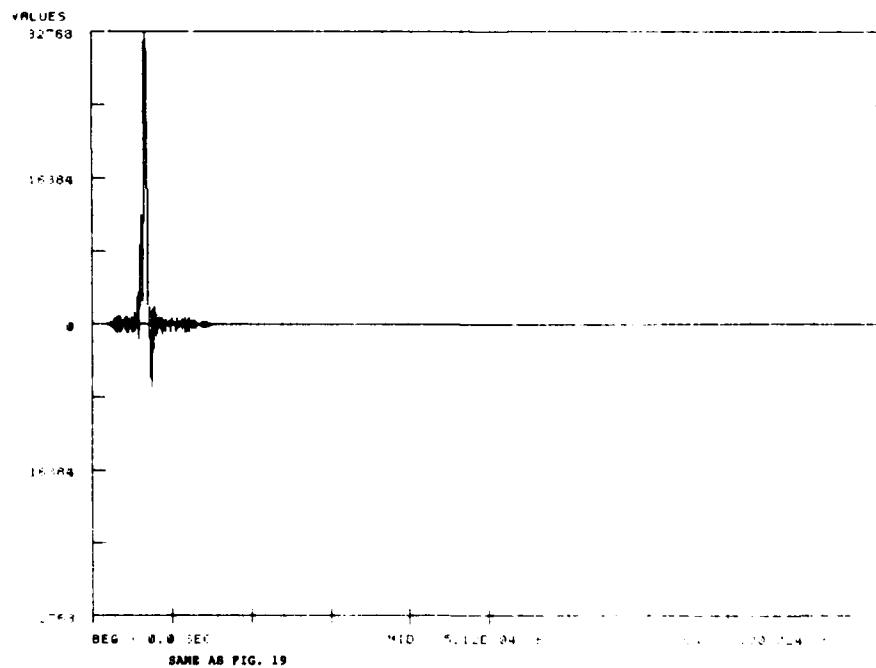


Figure 31. The result of low-time pass filtering the cepstrum of the waveform in Figure 15,  $\alpha = 0.9985$ , width of cepstral filter =  $\pm 4.5 \mu s$



**Figure 32.** The result of low-time pass filtering the cepstrum of the waveform in Figure 15,  $\alpha = 0.9985$ , width of cepstral filter =  $\pm 9.5 \mu s$



**Figure 33.** The result of low-time pass filtering the cepstrum of the waveform in Figure 15,  $\alpha = 0.9985$

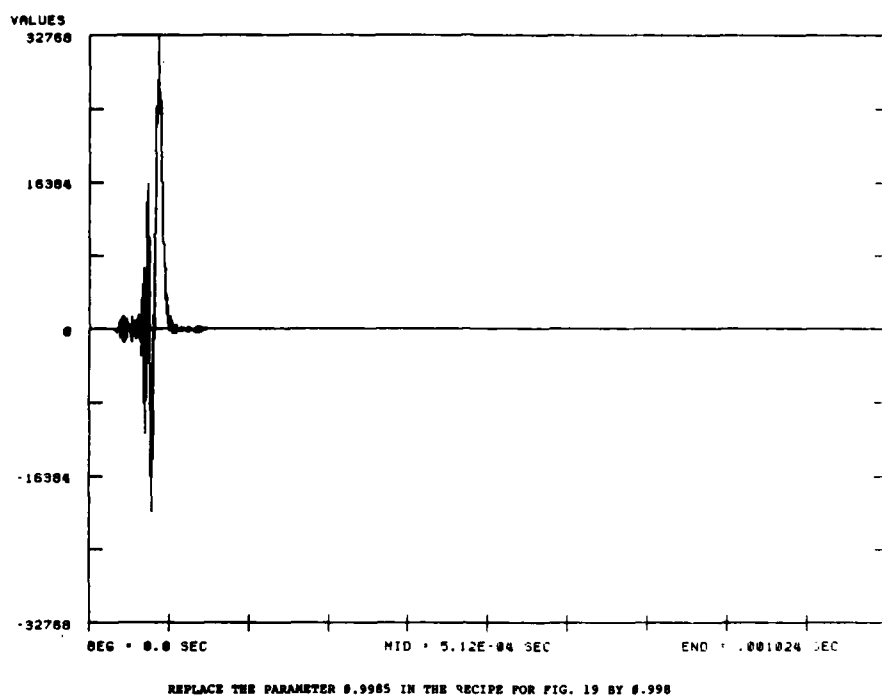


Figure 34. The result of low-time pass filtering the cepstrum of the waveform in Figure 15,  $\alpha = 0.9980$

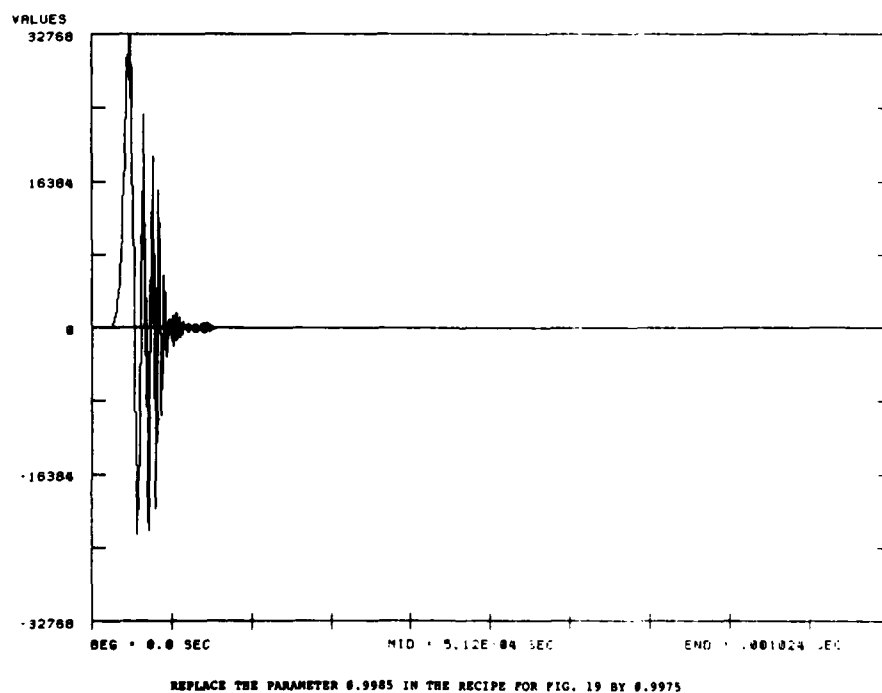
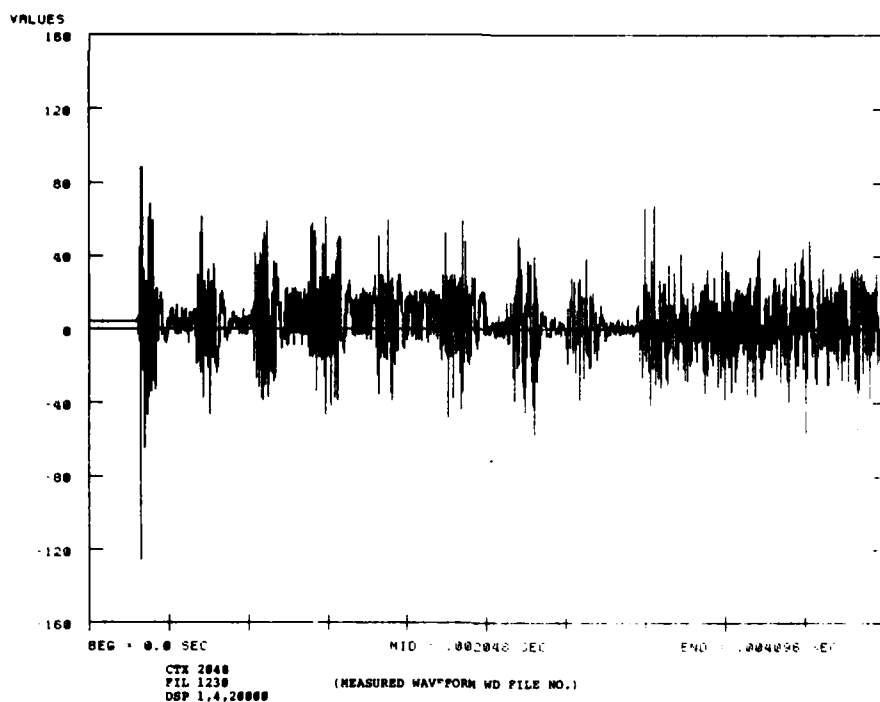
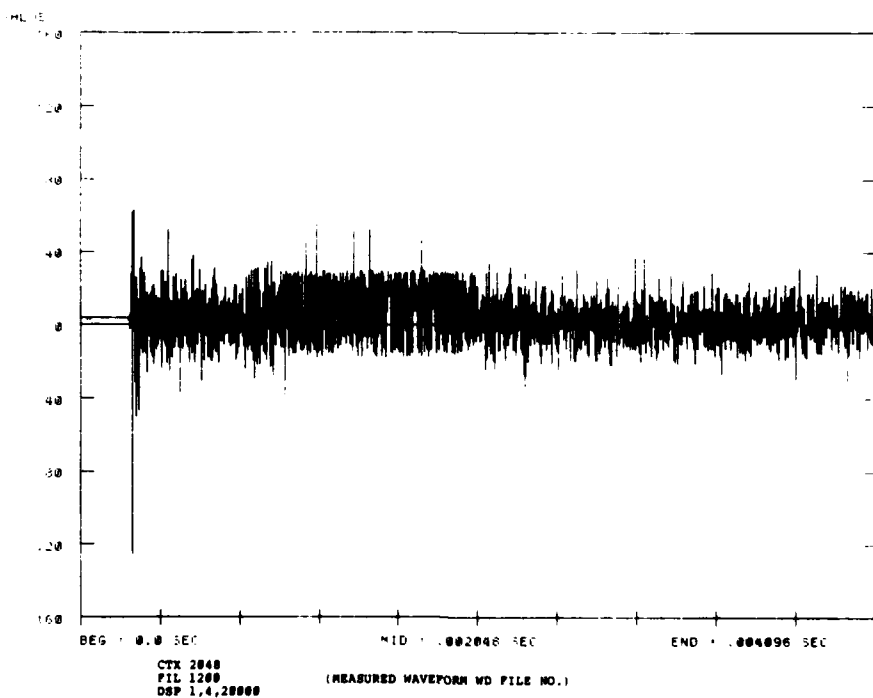


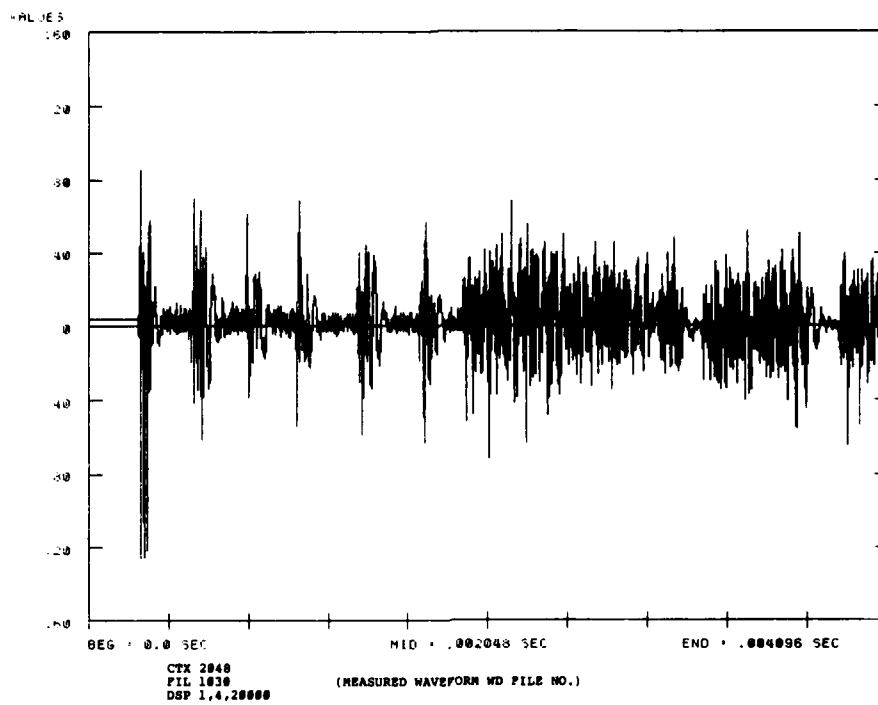
Figure 35. The result of low-time pass filtering the cepstrum of the waveform in Figure 15,  $\alpha = 0.9975$



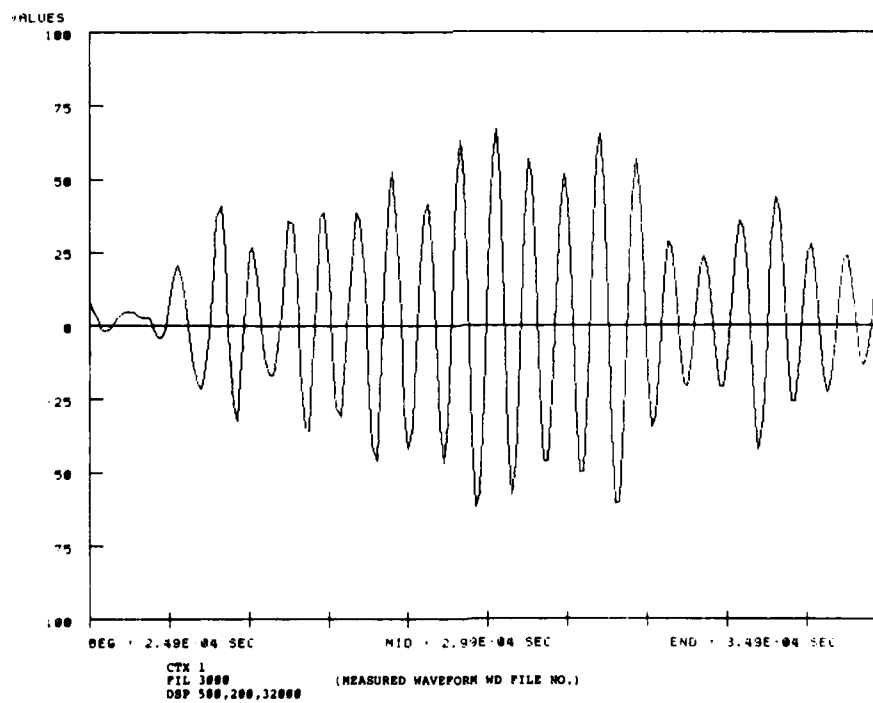
**Figure 36. A typical large-angle Pentel lead-break waveform; the angle between the lead and the surface is  $65^\circ$**



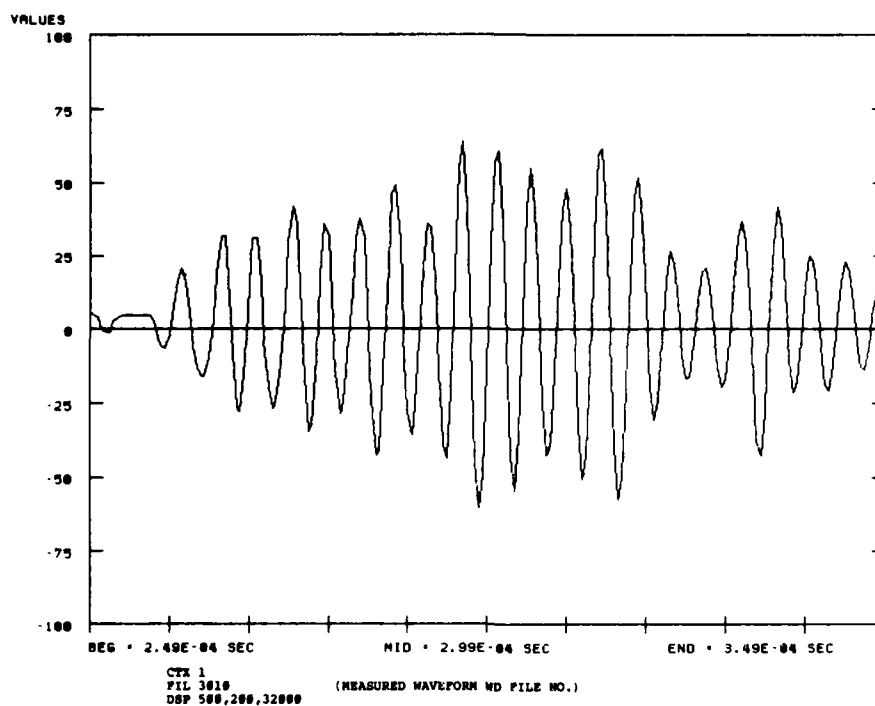
**Figure 37. A typical small-angle Pentel lead-break waveform; the angle between the lead and the surface is  $40^\circ$**



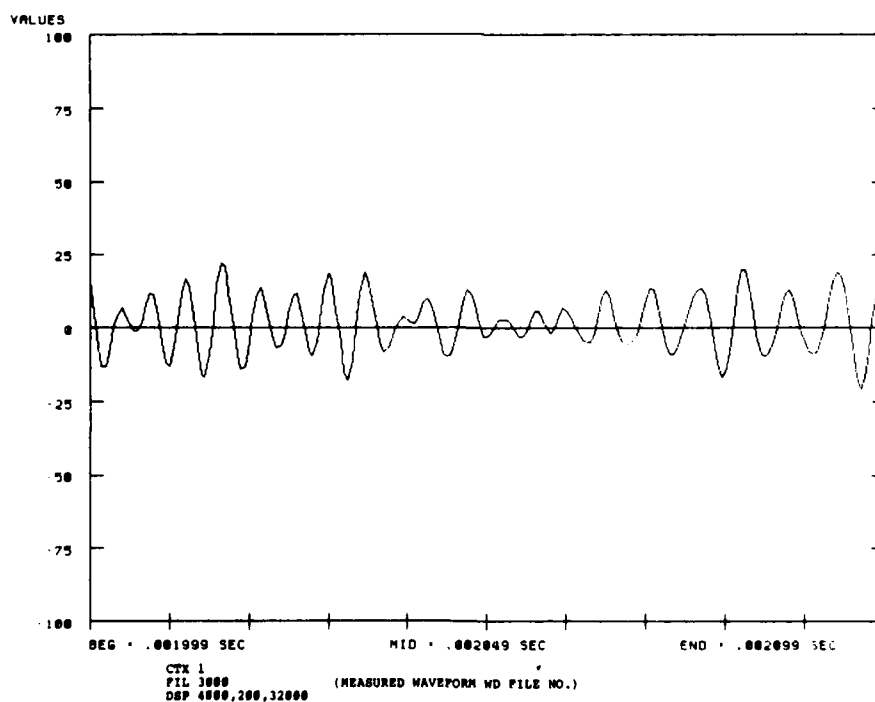
**Figure 38. A Pentel lead-break waveform with secondary Pentel tip impact at a large angle**



**Figure 39. A 200-point sample of a fixture-generated lead-break waveform near the initial impulse**

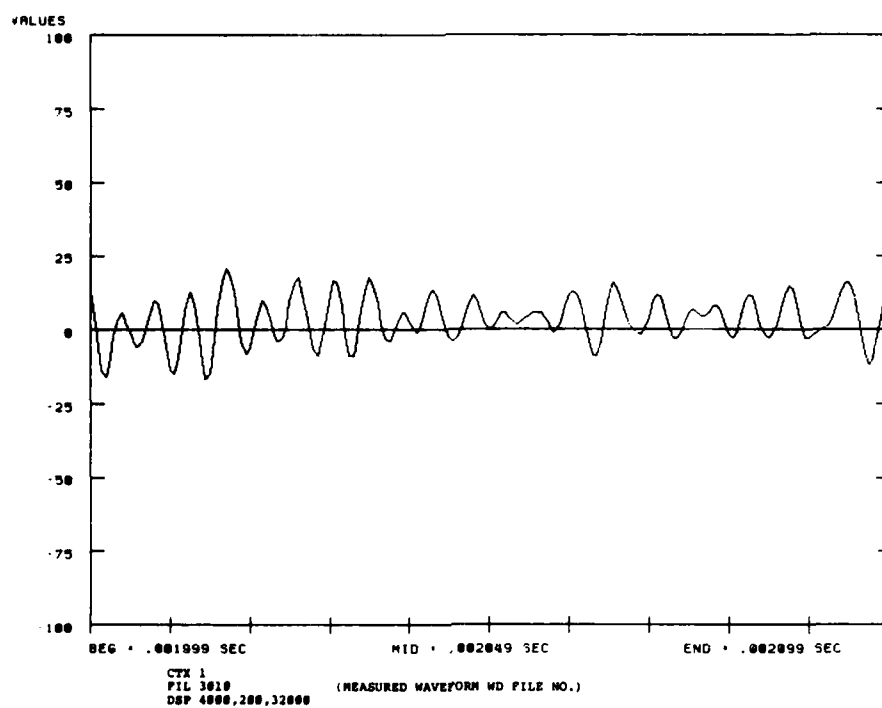


**Figure 40. A 200-point sample of another fixture-generated lead-break waveform near the initial impulse**

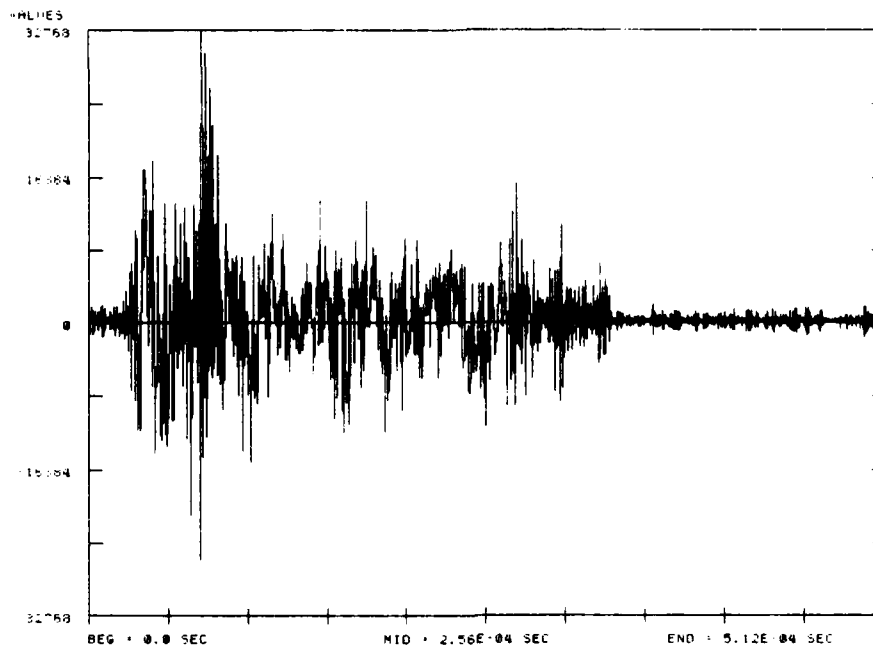


**Figure 41. Another 200-point sample of the waveform in Figure 39 near the later part of the waveform**





**Figure 42. Another 200-point sample of the waveform in Figure 40 near the later part of the waveform**



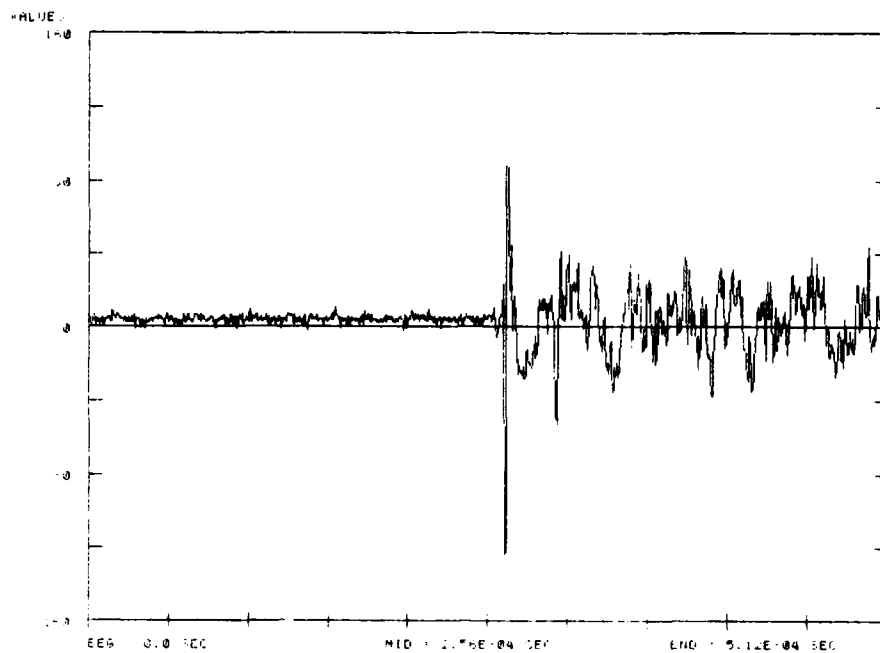
```

CTX 6192
@AVERG      (AVERAGE A NUMBER OF WAVEFORMS)
>FN         (INPUT FILE NAME CONTAINING THE WD NOS. OF THE FILES TO
              BE AVERAGED)
>N1         (OUTPUT SECONDARY WD FILE NO. TO STORE THE AVERAGED WAVEFORM)
CTX 1824
FIL N1
FIL SO
>CIN2,,1,1824      (N2 IS THE SECONDARY WD FILE NO., 0<N2<9999,
                    TO STORE THE FIRST 1824 CHANNELS OF FILE N1)

NDF S0,63,-32888
TRF 0
TRF 1,1,1,188      (COPY THE FIRST 1824 CHANNELS OF FILE N1 TO
                    FILE N2)
FIL N2
FIL SN3            (N3 IS THE SECONDARY WD FILE NO., 0<N3<9999,
                    TO STORE THE CEPSTRUM)
KCP,,1
>0.999
@XCPHIGH         (HIGH-PASS CEPSTRAL FILTER AND
                  INVERSE TRANSFORM)
>N3
>N4              (N4 IS THE SECONDARY WD FILE NO., 0<N4<9999,
                  TO STORE THE INVERSE CEPSTRUM)
>9
>1817
>1824
DSP S1,1,188

```

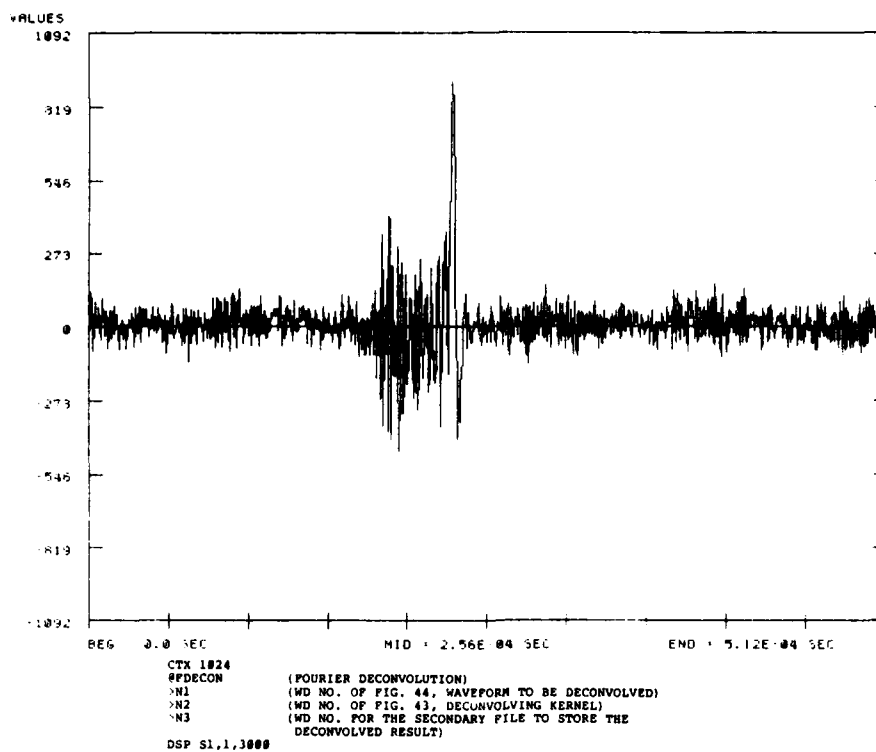
**Figure 43.** The impulse response obtained by high-time pass filtering the cepstrum of the average of 33 mechanically generated lead-break waveforms;  $\alpha = 0.999$



CTX 1824  
 FIL 3200 (MEASURED WAVEFORM WD FILE NO.)  
 FIL 50  
 >CZNI,,1,1824 (N1 IS THE SECONDARY WD FILE NO., <N1<9999,  
 TO STORE THE FIRST 1824 CHANNELS OF FILE 3200)

NDF 58,63,-32000  
 TRF H  
 TRF 1,1,1,180 (COPY THE FIRST 1824 CHANNELS OF FILE 3200 TO  
 FILE N1)  
 DSP S1,1,20000

**Figure 44. One of the 33 mechanically generated lead-break waveforms used in the averaging**



**Figure 45. The result of deconvolving the waveform in Figure 44 by the impulse response in Figure 43**

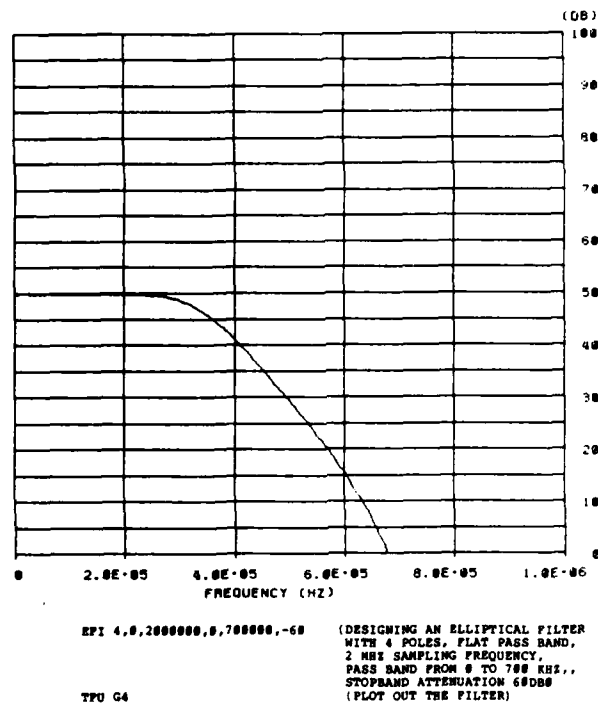
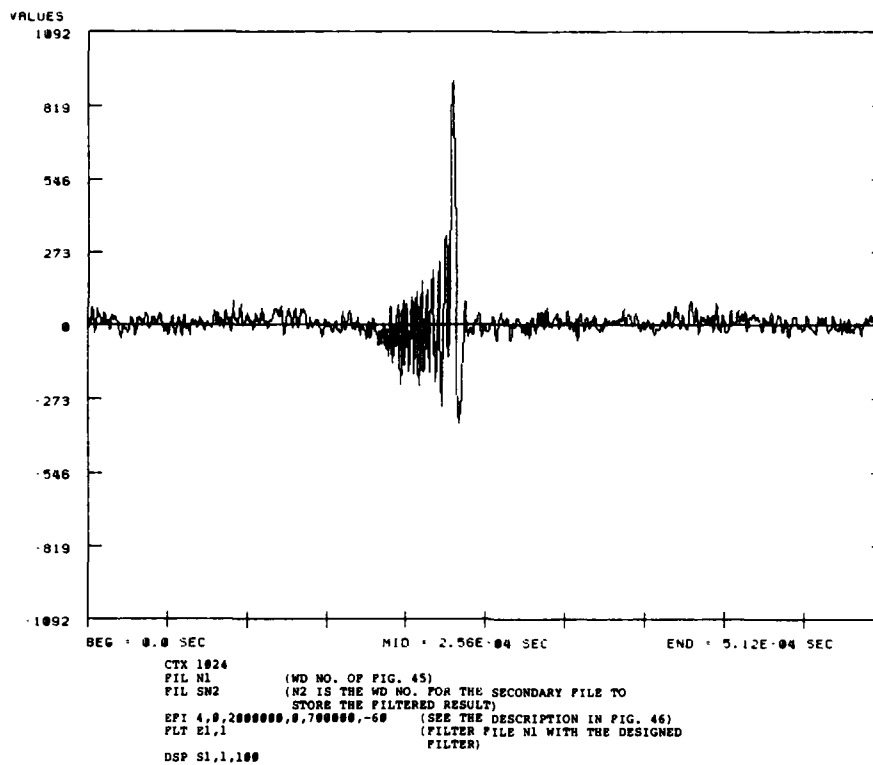
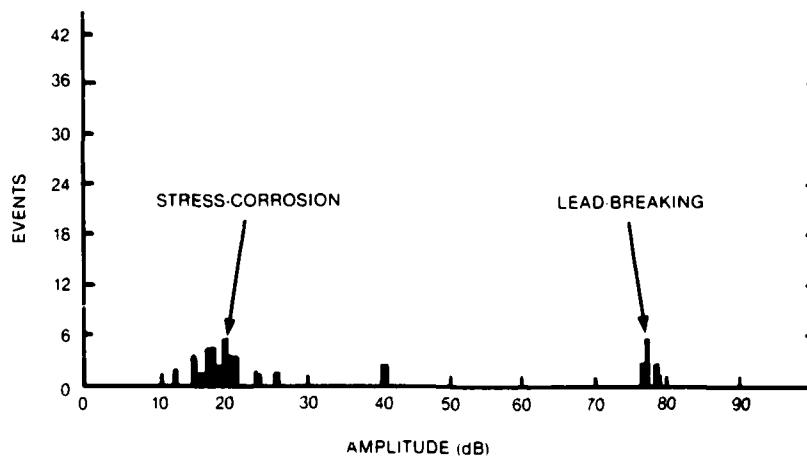


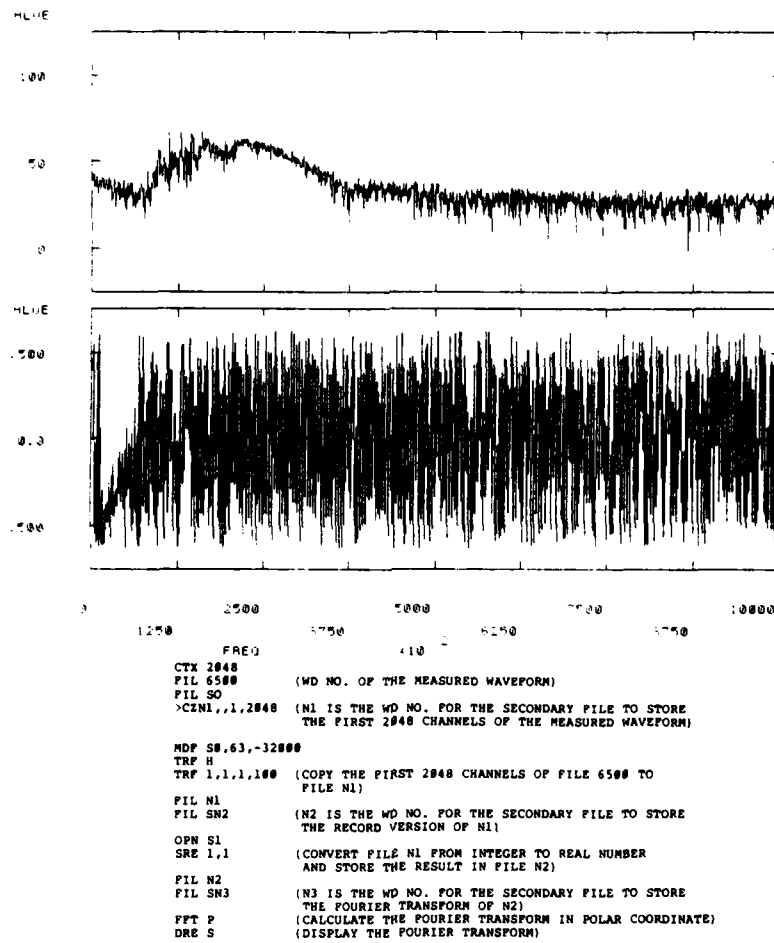
Figure 46. An elliptical frequency filter used to reduce the high-frequency noise in the result in Figure 45; -60 dB cut-off at 700 kHz



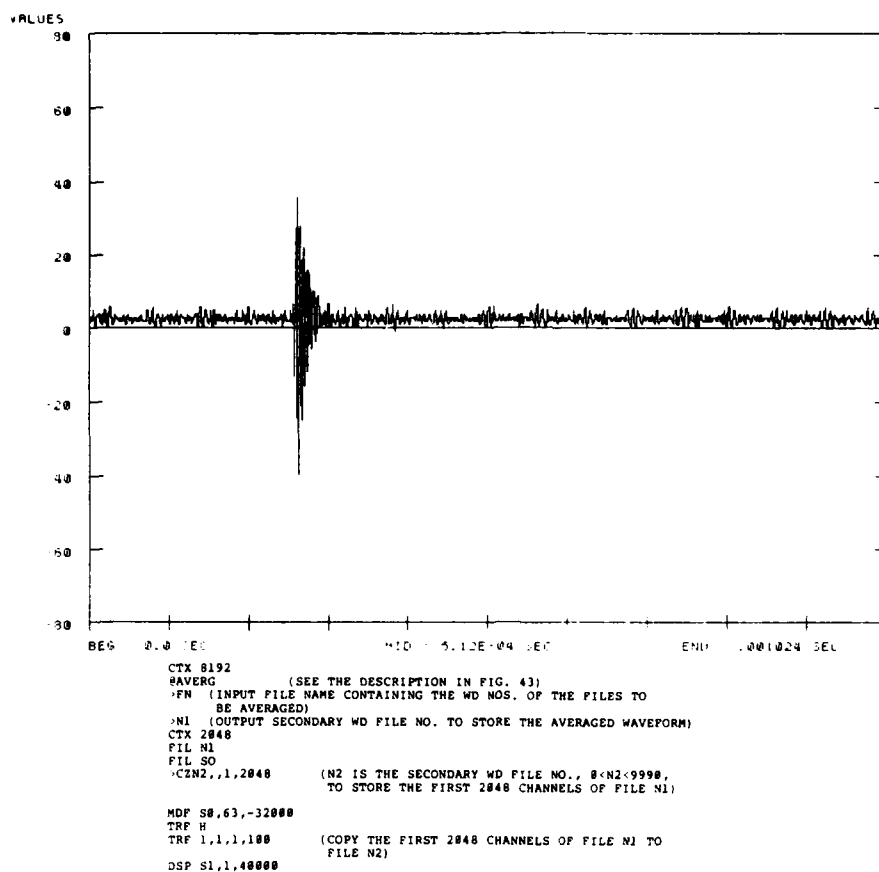
**Figure 47.** The result of filtering the waveform in Figure 45 with the elliptical filter in Figure 46



**Figure 48.** Relative amplitude of the stress corrosion events and the lead-breaking events



**Figure 49. Power spectrum of a stress-corrosion event captured by a resonance transducer**



**Figure 50. The average of 31 stress-corrosion waveforms with their initial impulses lined up at the same location; the waveforms were captured with a resonance transducer**



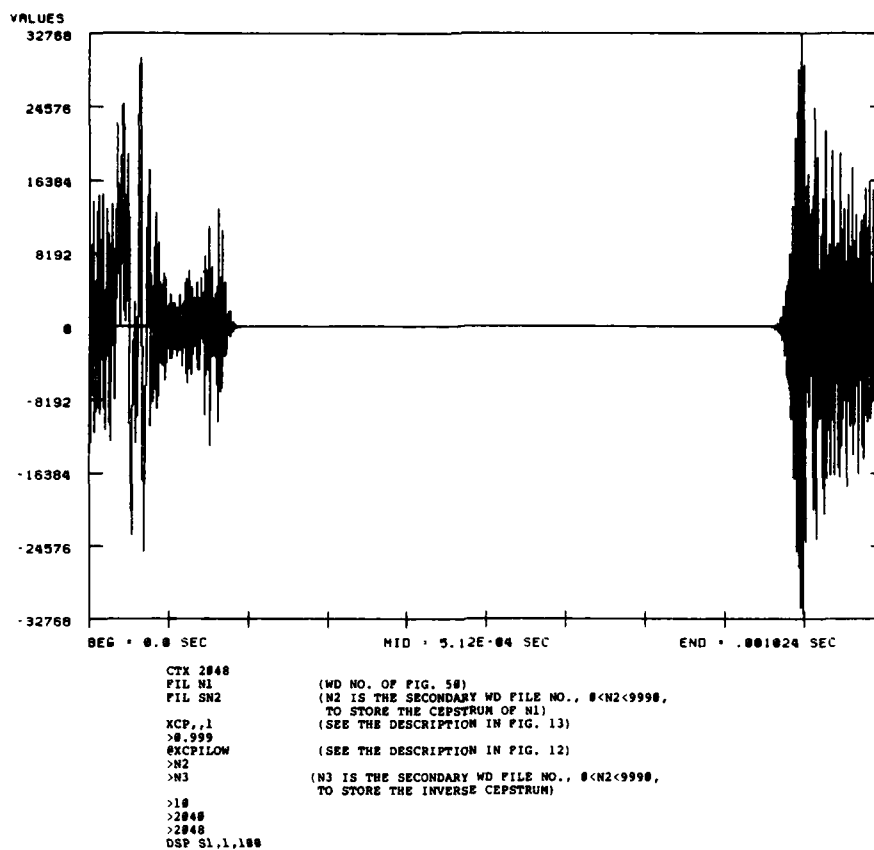


Figure 51. The result of low-time pass filtering the cepstrum of the averaged waveform in Figure 50; no band-pass mapping

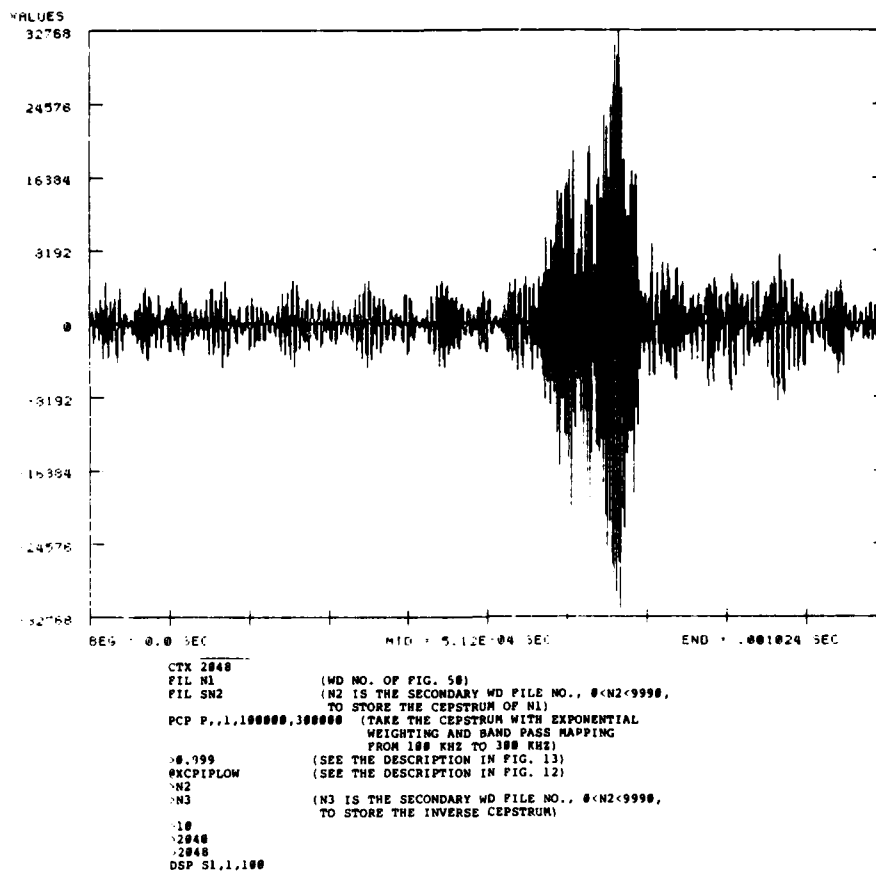
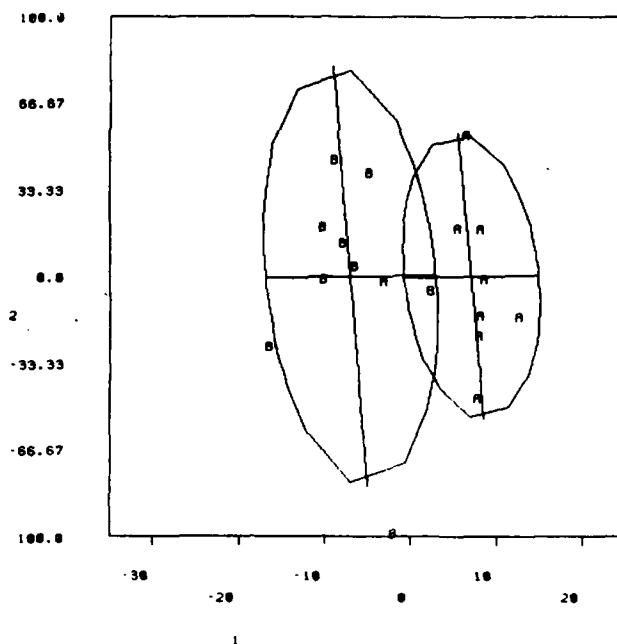
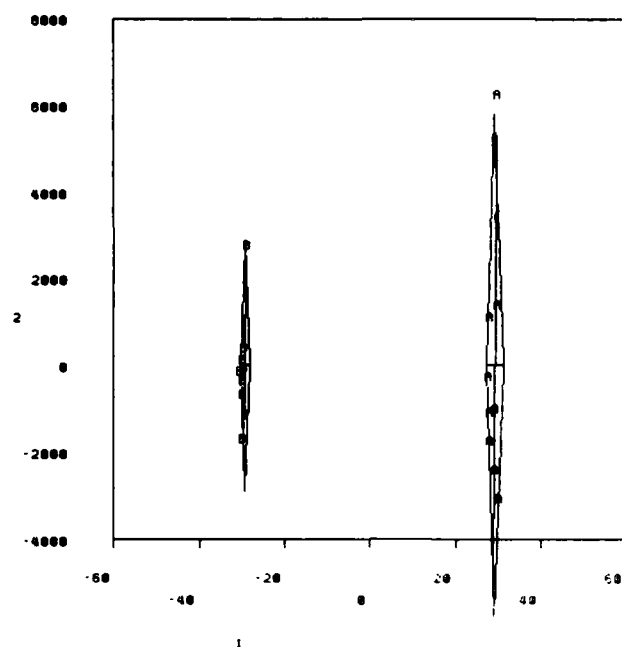


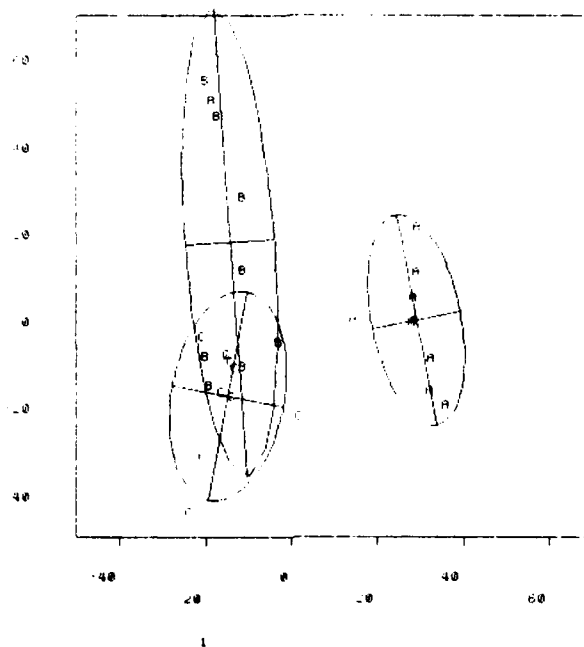
Figure 52. The result of low-time pass filtering the cepstrum of the averaged waveform in Figure 50; band-pass mapping between 100 kHz and 300 kHz



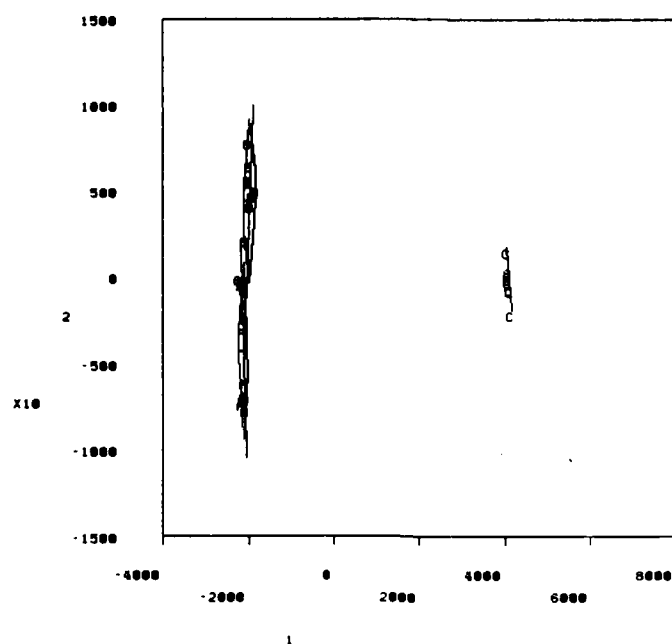
**Figure 53.** Scatter plots of the first two principal components of the frequency components of two sets of events after low-time pass filtering in the cepstral domain. The two sets of events were generated by (a) rubbing a sandpaper on the edge of one end of the aluminum block and (b) rubbing the sandpaper on the facet of the same end of the block



**Figure 54.** Scatter plots of the first two principal components of the frequency components of the two sets of events in Figure 53 without filtering in the cepstral domain



**Figure 55.** Scatter plots of the first two principal components of the frequency components of three sets of events after low-time pass filtering in the cepstral domain; the three sets of events were generated by (a) rubbing a sandpaper on the edge of one end of the aluminum block, (b) rubbing the sandpaper on the facet of the same end of the block, and (c) lead-breaking



**Figure 56. Scatter plots of the first two principal components of the frequency components of the three sets of events in Figure 55 without filtering in the cepstral domain**

## Section 5

### CONCLUSION

The key results obtained in this study are summarized below:

- Adaptive homomorphic deconvolution seems to work relatively well for AE signals captured with a wide-band flat response transducer with good S/N. This technique is able to clean up multipath and multimode reverberation and ring-down effects rather well. The technique does, however, have some unpleasant ad hoc elements, in that several key parameters cannot be determined *a priori*. Specifically, the decay exponent, needed for exponential weighting to make the sequence a well-behaved minimum phase sequence, is a problem, as are the linear filter coefficients needed for filtering the cepstral waveform. The appropriate constants can quickly be found with a little experimentation, but the process is non-unique.
- Adaptive homomorphic deconvolution seems to work to some degree with a conventional resonant transducer, particularly when bandpass mapping is employed. The success of the deconvolution is severely limited, however, because of the limited signal bandwidth information provided by the resonant transducer. As can be seen in plots shown in Section 4.6, although pulse compression has been achieved, the reconstructed signal suffers from serious ringing effects due to the very limited transducer bandwidth.
- Adaptive homomorphic deconvolution, implemented by high-pass filtering the averaged cepstral domain signal, seems to be effective in providing approximate estimates of the workpiece transfer function, particularly when the original signal is obtained with a wide-band transducer. As noted above, the technique is not effective when narrow-band resonant transducers are employed, because of the limited illumination of the cepstral domain afforded by those transducers. Unfortunately, the wide-band transducers employed in this study are not suitable for use in practical applications because of their fragility and severely limited sensitivity. Their sensitivity is not high enough to be used to obtain signals from typical, real AE sources; specifically, they are too insensitive for use with aluminum stress corrosion V-block sources, or most practical AE sources in general.
- The final step in employing adaptive homomorphic deconvolution to analyze acoustic emission signals is to employ ordinary Fourier deconvolution to eliminate the effects of transducer ringdown, multipath, etc. Fourier deconvolution, when used to eliminate the workpiece transfer function effects, is very sensitive to the precise details of the transfer function. In practical terms, this means that the transfer functions must be precisely repeatable from pulse to pulse. In fact, the data obtained in this study leads one to believe that there are substantial variations in the transfer function obtained from both artificial and natural AE sources, particularly when relatively long records of the pulse ringdown signals are obtained. The transfer functions are more repeatable when relatively short records are used. Preliminary analysis suggests that slight variations in the direction of the signal propagation may be responsible for these effects. A slight variation in the direction of propagation, or in the mix of longitudinal and shear

waves, will result in changes in the workpiece ringdown response function; these changes get larger and larger as the ringdown proceeds. Thus, the later segments of long records will suffer progressively increasing distortion from inaccurate deconvolution. Natural AE signals, since they emanate from evolving cracks where the crack propagation follows natural grain boundaries, may be expected to offer some variation in the strength and direction of the emitted shear and longitudinal waves. Surprisingly, even AE signals carefully generated via artificial means (Pental lead breaks) seem to vary in the precise details of their ringdown responses over relatively long record lengths. The net result is that only relatively short ( $\sim 2048$  points) record lengths can be used successfully, either with natural or artificial AE signal sources.

- Pattern recognition, employed as a means of identifying and characterizing AE sources through differences in the microstructure of their waveforms, does not seem to be particularly effective, either when applied to the original waveforms, or when applied to waveforms restored using adaptive homomorphic deconvolution. The original waveforms are heavily contaminated with the workpiece transfer function effects, and microstructure effects are difficult to separate. The reconstructed waveforms suffered from ringing distortion because of the limited bandwidth of the transducers employed. They were also very difficult to separate.

In conclusion, the analytical techniques explored in this research provide several new options in processing acoustic emission signals. Specifically, they provide the means to obtain and employ workpiece transfer functions in deriving the underlying acoustic emission waveforms. Successful application of these techniques requires the use of wide-band transducers that are faithful to the original waveforms and can only be employed on relatively short records. Any practical application must await the development of wide-band accurate transducers with much improved sensitivity over currently available transducers.



## Section 6

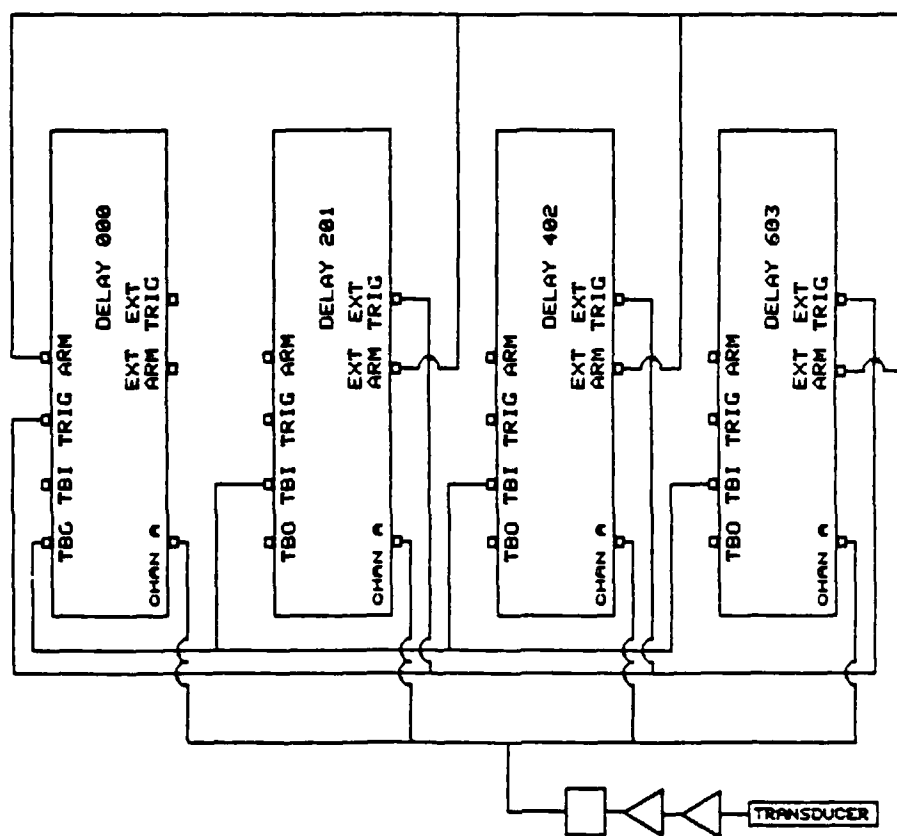
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2. Tribolet, J.M., *Seismic Applications of Homomorphic Signal Processing* (Englewood Cliffs, N.J.: Prentice-Hall, 1979).
3. Oppenheim, A.V., and Schaefer, R.W., *Digital Signal Processing* (Englewood Cliffs, N.J.: Prentice-Hall, 1975).
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5. Weinstein, Clifford J., et al., *Programs for Digital Signal Processing* (IEEE Press, 0-87942-127-4).
6. Chen, Y.T., *Interactive Pattern Recognition* (New York: Marcel Dekker, 1978).

**Appendix A**  
**BIOMATION 8100 DR11-C INTERFACE**

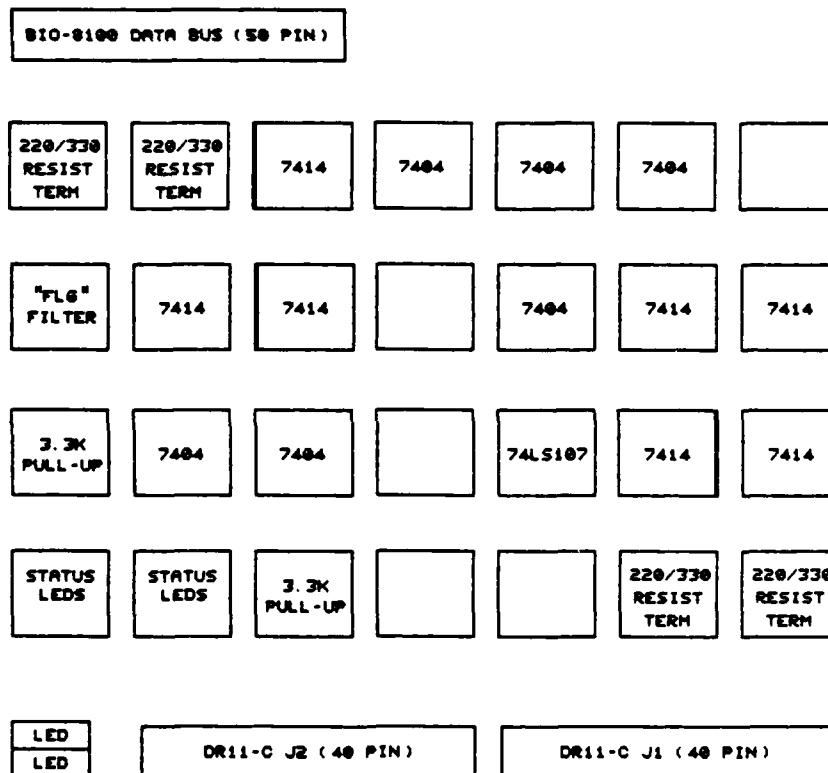
**PRELIMINARY DOCUMENTATION**

**BIOMATION 8100  
DR11-C  
INTERFACE**



PRELIMINARY 9/1/83

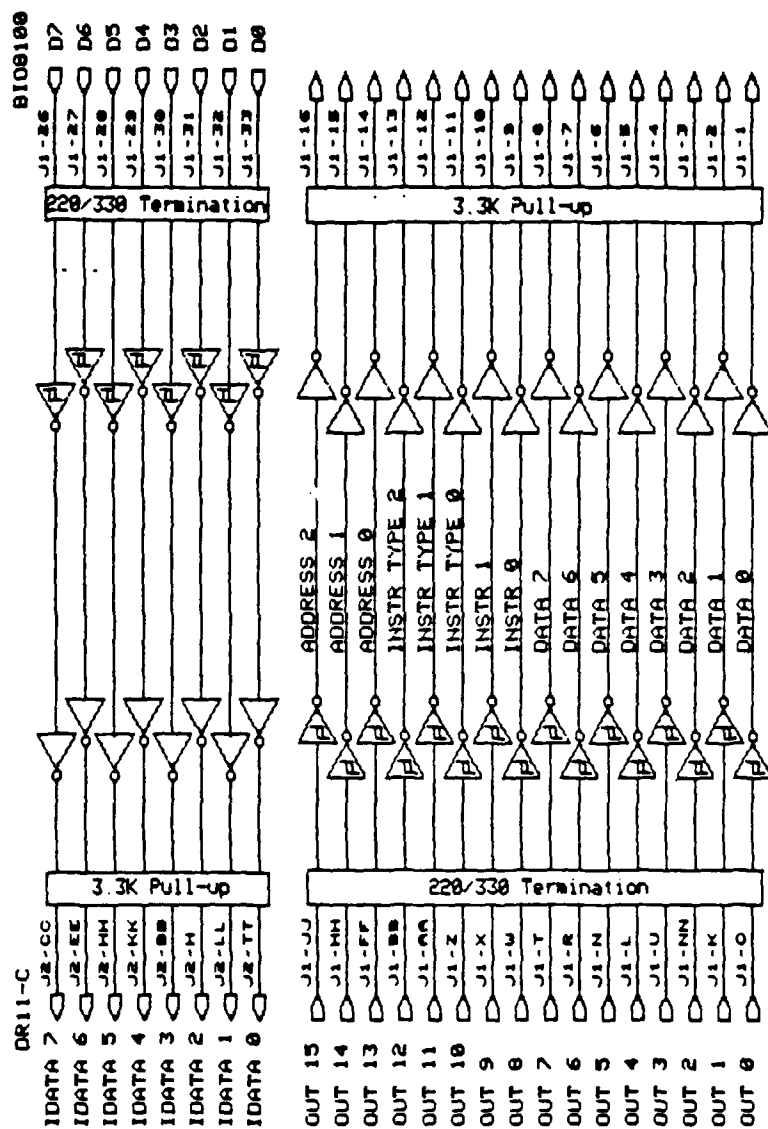
BIOMATION 8100 /DR11-C INTERFACE MODULE



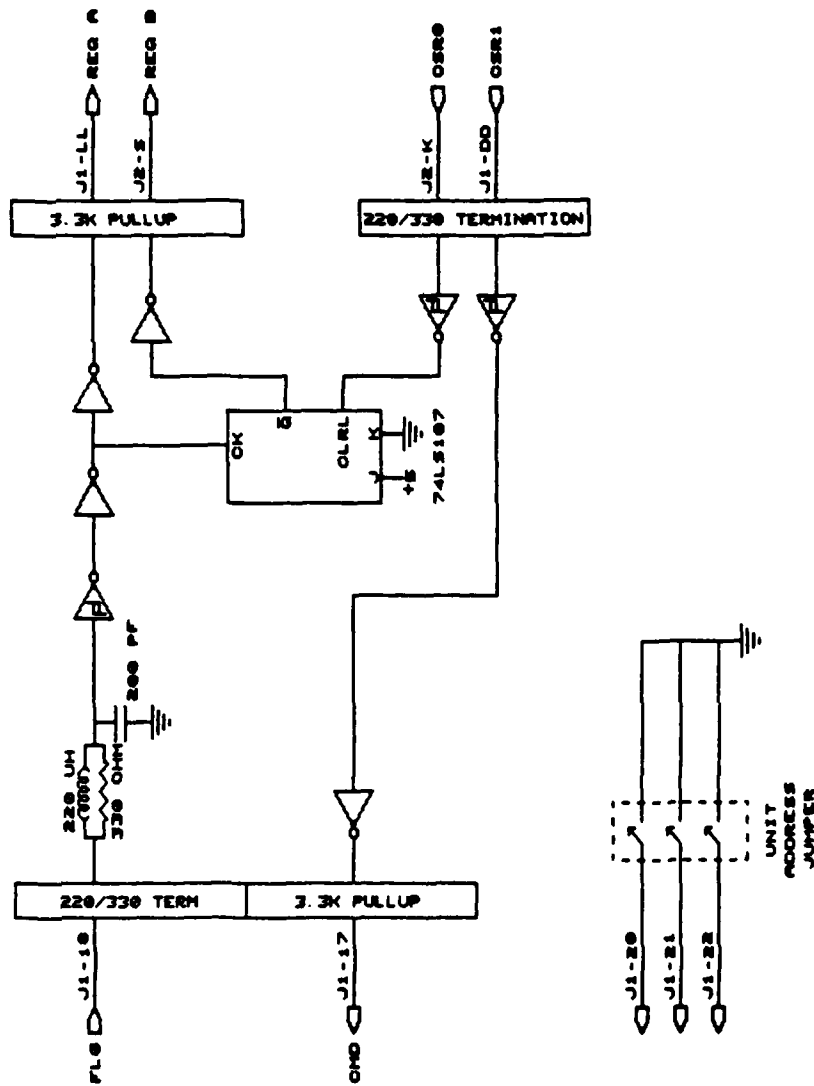
MD8-M91UW

PRELIMINARY 9/1/83

# BIOMATION 8100 / DR11-C CABLE MAPPING



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0001 PROGRAM BI08100V2
0002 . IDENT /V2-5a/
0003
0004 Software interface between VAX and Biomation B100
0005 transient recorder
0006
0007 This version implements full computer control of the
0008 B100s.
0009
0010 This version allows the user to disable correlation
0011 checking of the overlap region
0012
0013 IMPLICIT LOGICAL*4 (S)
0014
0015 C C C C C C C C C C C C C C C C C C C C C C C C
0016 C
0017 I/O page mapping
0018
0019 PARAMETER IPROT = '0080'X !S00M:RMED prot mask
0020 PARAMETER IO_SPACE_BASE = 'FF0000'X
0021 PARAMETER IPGENT = 1
0022 PARAMETER IUNIBUS_BASE = '760000'D
0023 PARAMETER IDRI1_CSR = '767770'D
0024 PARAMETER IBUGSIZ = 8192
0025 INCLUDE 'BI081STR.INC/LIST'
0026
0027 1 C C C C C C C C C C C C C C C C C C C C C C C C
0028 1 C
0029 1 C Biomation B100 digital control word format:
0030 1 C
0031 1 C 15 14 13 12 11 10 9 8 7 6 5 4 3 2 1 0
0032 1 C UNIT ADDR INSTR TYP INST ..... DATA .....
0033 1 C
0034 1 C C C C C C C C C C C C C C C C C C C C C C C C
0035 1 C Biomation B100 control instructions
0036 1 C
0037 1 C Instruction types:
0038 1 C
0039 1 C
0040 1 C
0041 1 C PARAMETER IT_CONTROL = '0800'X
0042 1 C PARAMETER IT_ORHDE = '0C00'X
0043 1 C PARAMETER IT_CHANB = '1000'X
0044 1 C PARAMETER IT_CHANA = '1400'X
0045 1 C PARAMETER IT_ARM = '1800'X
0046 1 C PARAMETER IT_TRIGGER = '1C00'X
0047 1 C
0048 1 C Instructions:
0049 1 C
0050 1 C PARAMETER I_10 = '0000'X
0051 1 C PARAMETER I_11 = '0100'X
0052 1 C PARAMETER I_12 = '0200'X
0053 1 C PARAMETER I_13 = '0300'X
0054 1 C
0055 1 C Data bit positions:
0056 1 C
0057 1 C PARAMETER IBIT_0 = '0001'X

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0058 1      PARAMETER IBIT_1 = '0002'X
0059 1      PARAMETER IBIT_2 = '0004'X
0060 1      PARAMETER IBIT_3 = '0008'X
0061 1      PARAMETER IBIT_4 = '0010'X
0062 1      PARAMETER IBIT_5 = '0020'X
0063 1      PARAMETER IBIT_6 = '0040'X
0064 1      PARAMETER IBIT_7 = '0080'X
0065 1      PARAMETER IBIT_A = '00FF'X
0066 1 C
0067 1 C      Biomation B100 instruction summary
0068 1 C
0069 1 C      B0-B7 are output bit masks
0070 1 C      D0-D7 are input bit masks
0071 1 C
0072 1 C C      CONTROL FUNCTIONS - IT_CONTROL
0073 1 C      (2)
0074 1 C
0075 1 C      I_10 - Set program mask
0076 1 C
0077 1 C      B7 - Trigger control group
0078 1 C      B6 - Arm control group
0079 1 C      B5 - Channel B control group
0080 1 C      B4 - Channel A control group
0081 1 C      B3 - Time base, output mode, record mode groups
0082 1 C      B2-B0 NOP
0083 1 C
0084 1 C      I_11 - Set function
0085 1 C
0086 1 C      NOTE: Only ONE bit may be set per instruction
0087 1 C      load.
0088 1 C
0089 1 C      B7 - Clear and update "status" word
0090 1 C      B6 - Plot
0091 1 C      B5 - NOP
0092 1 C      B4 - Arm
0093 1 C      B3 - Trigger
0094 1 C      B2 - NOP
0095 1 C      B1 - Switch to alternate time base
0096 1 C      B0 - Reset
0097 1 C
0098 1 C      I_12 - Read status
0099 1 C
0100 1 C      D7 - Record in progress
0101 1 C      D6 - Output data ready
0102 1 C      D5 - Ready to Arm
0103 1 C      D4 - Ready to Trigger
0104 1 C      D3 - Offscale Chan A (+)
0105 1 C      D2 - Offscale Chan A (-)
0106 1 C      D1 - Offscale Chan B (+)
0107 1 C      D0 - Offscale Chan B (-)
0108 1 C
0109 1 C      I_13 - Enable output data
0110 1 C
0111 1 C      D7-D0 output data
0112 1 C
0113 1 C C      OUTPUT MODE, TIME BASE, RECORD MODE - IT_OMODE
0114 1 C      (3)

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0115 1 C
0116 1 C
0117 1 C
0118 1 C
0119 1 C
0120 1 C
0121 1 C
0122 1 C
0123 1 C
0124 1 C
0125 1 C
0126 1 C
0127 1 C
0128 1 C
0129 1 C
0130 1 C
0131 1 C
0132 1 C
0133 1 C
0134 1 C
0135 1 C
0136 1 C
0137 1 C
0138 1 C
0139 1 C
0140 1 C
0141 1 C
0142 1 C
0143 1 C
0144 1 C
0145 1 C
0146 1 C
0147 1 C
0148 1 C
0149 1 C
0150 1 C
0151 1 C
0152 1 C
0153 1 C
0154 1 C
0155 1 C
0156 1 C
0157 1 C
0158 1 C
0159 1 C
0160 1 C
0161 1 C
0162 1 C
0163 1 C
0164 1 C
0165 1 C
0166 1 C
0167 1 C
0168 1 C
0169 1 C
0170 1 C
0171 1 C

I_10 - Main time base
      B7  B6  SOURCE
      0  0  Internal
      0  1  External < 25 ms
      1  0  External > 25 ms
      1  1  External .25 ms - 1 ms

      B5  B4  UNITS
      0  0  us
      0  1  ms
      1  0  s
      1  1  NOP

      B3-B0  HEX  RANGE
      0  .01
      1  .02
      2  .05
      3  .1
      4  .2
      5  .5
      6  1
      7  2
      8  5
      9  10
      A-F  UNDEFINED

I_11 - Alternate time base
      Identical to Main Time Base

I_12 - Output mode
      B7-B2  NOP
      B1  B0  MODE
      0  0  OFF
      0  1  AUTO
      1  0  EDIT
      1  1  UNDEFINED

I_13 - Record mode
      B7-B6  NOP
      B5  B4  START RECORD
      0  0  Arm
      0  1  Delayed Arm
      1  0  Trigger
      1  1  Delayed Trigger

      B3  B2  SWITCH TO ALT TIMEBASE
      0  0  NOP
      0  1  Delayed Arm
      1  0  Trigger
      1  1  Delayed Trigger

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0172 1 C
0173 1 C
0174 1 C
0175 1 C
0176 1 C
0177 1 C
0178 1 C
0179 1 C C C
0180 1 C
0181 1 C
0182 1 C
0183 1 C
0184 1 C
0185 1 C
0186 1 C
0187 1 C
0188 1 C
0189 1 C
0190 1 C
0191 1 C
0192 1 C
0193 1 C
0194 1 C
0195 1 C
0196 1 C
0197 1 C
0198 1 C
0199 1 C
0200 1 C
0201 1 C
0202 1 C
0203 1 C
0204 1 C
0205 1 C
0206 1 C
0207 1 C
0208 1 C
0209 1 C
0210 1 C
0211 1 C
0212 1 C
0213 1 C
0214 1 C
0215 1 C
0216 1 C
0217 1 C
0218 1 C
0219 1 C
0220 1 C
0221 1 C
0222 1 C
0223 1 C
0224 1 C
0225 1 C
0226 1 C
0227 1 C
0228 1 C C C

      B1      B0      STOP RECORD
      0      0      End-of-sweep
      0      1      NOP
      1      0      Trigger
      1      1      Delayed Trigger

      CHANNEL A FUNCTIONS - IT_CHANA
      (4)

      I_10 - Range and coupling

      B7      B6      + INPUT COUPLING
      0      0      Off
      0      1      AC
      1      0      DC
      1      1      NOP

      B5      B4      - INPUT COUPLING
      0      0      Off
      0      1      AC
      1      0      DC
      1      1      NOP

      B3-B0      HEX      RANGE
      0      0      +- .05 (V)
      1      1      +- .1
      2      2      +- .2
      3      3      +- .5
      4      4      +- 1.
      5      5      +- 2.
      6      6      +- 5.
      7-F      7-F      UNDEFINED

      I_11 - Input offset magnitude
      ex:      BCD coded in lower byte
               value 99 = '99'X

      I_12 - Input mode and offset sign

      B7-B5      NOP
      B4      MODE
      0      Off
      1      Input
      B3-B1      NOP
      B0      POLARITY
      0      -
      1      +

      I_13 - NOP

      CHANNEL B FUNCTIONS - IT_CHANB
  
```

```

0229 1 C      (5)
0230 1 C      Identical to Channel A functions
0231 1 C
0232 1 C
0233 1 C      I_10 - Range and coupling
0234 1 C      I_11 - Input offset magnitude
0235 1 C      I_12 - Input mode and offset sign
0236 1 C      I_13 - NOP
0237 1 C
0238 1 C C C  ARM FUNCTIONS - IT_ARM
0239 1 C      (6)
0240 1 C
0241 1 C      I_10 - Delay magnitude, 2nd and least digits
0242 1 C
0243 1 C      BCD coded in lower byte
0244 1 C
0245 1 C      I_11 - Delay magnitude, most significant digit
0246 1 C
0247 1 C      BCD coded in lowest 4 bits of lower byte
0248 1 C
0249 1 C      I_12 - Level magnitude
0250 1 C
0251 1 C      BCD coded in lower byte
0252 1 C
0253 1 C      I_13 - Arm mode, source, coupling, slope and level
0254 1 C      polarity
0255 1 C
0256 1 C      B7 MODE
0257 1 C      0 Input
0258 1 C      1 Auto
0259 1 C
0260 1 C      B6 NOP
0261 1 C
0262 1 C      B5 SOURCE
0263 1 C      0 External
0264 1 C      1 Internal
0265 1 C
0266 1 C      B4 Internal Source
0267 1 C      0 Chan A
0268 1 C      1 Chan B
0269 1 C
0270 1 C      B3 SLOPE
0271 1 C      0 +
0272 1 C      1 -
0273 1 C
0274 1 C      B2 COUPLING
0275 1 C      0 DC
0276 1 C      1 AC
0277 1 C
0278 1 C      B1 NOP
0279 1 C
0280 1 C      B0 LEVEL POLARITY
0281 1 C      0 -
0282 1 C      1 +
0283 1 C
0284 1 C C C  TRIGGER FUNCTIONS - IT_TRIGGER
0285 1 C      (7)

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[illegible]



BI08100V2

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0400 C
0401 20 WRITE (6,25)
0402 25 FORMAT ('$Starting WD file number: ')
0403 READ (5,17,ERR=20,END=20) IFIRST
0404 IF (IFIRST.EQ.9999) THEN
0405 TYPE *, 'INVALID WD FILE NUMBER: ', IFIRST
0406 GO TO 20
0407
0408 C
0409 30 WRITE (6,35)
0410 35 FORMAT ('$WD file number increment: ')
0411 READ (5,17,ERR=30,END=30) INCR
0412 IF (INCR.EQ.0) INCR=1
0413
0414 C
0415 40 WRITE (6,41)
0416 41 FORMAT ('$Enter numeric portion of sample interval: ')
0417 READ (5,42) INPUT_LINE
0418 FORMAT (A)
0419 STATUS=STRTRIM(INPUT_LINE,INPUT_LINE,INLEN)
0420 IF (INLEN.EQ.0) GO TO 40
0421 IPOS=INDEX(INPUT_LINE(1:INLEN),'.')
0422 IF (IPOS.EQ.0) THEN
0423 INLEN=INLEN+1
0424 INPUT_LINE(INLEN:INLEN)='.'
0425
0426 C
0427 44 READ (INPUT_LINE(1:INLEN),44,ERR=40) SAMP
0428 FORMAT (F)
0429
0430 C
0431 Is this value in the list?
0432 ISNR=0
0433 DO I=1,10
0434 IF (SAMP.EQ.SAMPINT(I)) ISNR=I
0435 ENDDO
0436 IF (ISNR.EQ.0) THEN
0437 WRITE (6,46) SAMP
0438 FORMAT (' INVALID PARAMETER: ',F)
0439 GO TO 40
0440
0441 C
0442 50 WRITE (6,51)
0443 format ('Units: S - 1',/,
0444 1 , mS - 2',/,
0445 2 , uS - 3')
0446
0447 C
0448 52 WRITE (6,52)
0449 FORMAT ('$Select [1-3]: ')
0450 READ (5,17,ERR=50,END=50) IMUL
0451 IF (IMUL.LT.1. OR IMUL.GT.3) GO TO 50
0452
0453 C
0454 Biomatron B100 will not do dual channel operation
0455 at a rate above 10 Mhz
0456
0457 C
0458 IF (INCHANS.EQ.2 AND (ISNR.LT.4) AND IMUL.EQ.3) THEN
0459 WRITE (6,62)
0460 FORMAT (' ERROR TWO CHANNEL OPERATION MAX 10Mhz')
0461 STOP
0462
0463 C
0464 ENDIF

```

**A-14**



```

00514 DATSEC(2)=DATSEC(2)*'200'X
00515 IF (DATSEC(1).NE.DATSEC(2)) THEN
00516   WRITE (6,100) DATSEC(1)
00517   FORMAT (' ERROR - ARRAY TO BE MAPPED IS NOT',
00518         ' PAGE ALIGNED: ADDR=',ZB.8)
00519   STOP 'ERROR ABORT'
00520 ENDIF
00521
00522 Set flags for writable private section. using page
00523 frame mapping
00524
00525 SEC_FLO=XLOC(SECON_WRT)+XLOC(SECON_PFNMAP)
00526
00527 Increase the working set size as much as the system
00528 allows
00529
00530 MAXWSL=3072-WSETLM
00531 SUCCESS=SYSADJWSL(MAXWSL,WSETLM)
00532 TYPE *,MAXWSL,WSETLM
00533
00534 Perform the actual mapping of the desired page of
00535 IO space
00536
00537 WRITE (6,110) VBN,DATSEC(1),DATSEC(2)
00538 FORMAT (' VBN=',Z4.4,Z(4,ZB.8))
00539 STATUS=SYSBCRMPBC(DATSEC,DATRET,%VAL(BEC_FLO),,
00540               'XVAL(1),XVAL(VBN),IPROT,.)
00541 IF (.NOT. STATUS)
00542   CALL FATAL_ERROR('Mapping IO page')
00543
00544 Lock the data array into memory
00545
00546 LCKREG(1)=XLOC(IDATA)
00547 LCKREG(2)=LCKREG(1)+IBUFSIZ*2-1
00548 STATUS=SYSBLCKPAQ(LCKREG,LCKRET,.)
00549 IF (.NOT. STATUS)
00550   CALL FATAL_ERROR('Locking data array')
00551
00552 WRITE (6,120) LCKRET(1),LCKRET(2)
00553 FORMAT (' Data area locked: ',ZB.8,' ',ZB.8)
00554
00555 Lock the io access code into memory
00556
00557 STATUS=PIO_CODE(CODEST,CODEND)
00558 CODREG(1)=(CODEST)
00559 CODREG(2)=(CODEND)
00560 STATUS=SYSBLCKPAQ(CODREG,CODRET,.)
00561 IF (.NOT. STATUS)
00562   CALL FATAL_ERROR('Locking i/o code')
00563
00564 WRITE (6,130) CODRET(1),CODRET(2)
00565 FORMAT (' Code area locked: ',ZB.8,' ',ZB.8)
00566
00567 C C C C C C C C C C C C C C C C C C C C C C C C C C C C
00568 C C C C C C C C C C C C C C C C C C C C C C C C C C C C
00569 C C C C C C C C C C C C C C C C C C C C C C C C C C C C
00570
00571 Main loop:
00572
00573 CONTINUE
00574 SDEBUG=.FALSE.
00575

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0571 WRITE (6,210)
0572 FORMAT ('OMenu: ',/)
0573 1 ' <c/r> - Acquire next waveform',/
0574 1 ' CORR - Toggle enable/disable correlation fitting',/
0575 1 ' CLEAR - Clear outputs',/
0576 1 ' NEW - Re-enter parameters',/
0577 1 ' <Z> - Exit',/
0578 WRITE (6,215)
0579 FORMAT ('$Function: ')
0580 READ (5,220,END=1000,ERR=200) COMMAND
0581 FORMAT (A)
0582 FORMAT (' ',A)
0583 STATUS=STRUPCASE(COMMAND,COMMAND)
0584 STATUS=STRSTRIM(COMMAND,COMMAND,LCOMM)
0585 IF (LCOMM.EQ.0) GO TO 250
0586
0587 IF (COMMAND(1:3).EQ.'NEW') THEN
0588 WRITE (6,222) COMMAND
0589 SINI=.TRUE.
0590 GO TO 1234
0591
0592 ENDIF
0593
0594 IF (COMMAND(1:3).EQ.'DEB') THEN
0595 WRITE (6,222) COMMAND
0596 SDEBUO=.TRUE.
0597 GO TO 250
0598
0599 ENDIF
0600
0601 IF (COMMAND(1:3).EQ.'COR') THEN
0602 CORRCHK=.NOT.CORRCHK
0603 IF (CORRCHK) THEN
0604 WRITE (6,222) 'Correlation ENABLED!'
0605 ILBHDR(63)=-32000
0606
0607 ELSE
0608 WRITE (6,222) 'Correlation DISABLED!'
0609 ILBHDR(63)=0
0610
0611 ENDIF
0612 GO TO 200
0613
0614 IF (COMMAND(1:3).EQ.'CLE') THEN
0615 WRITE (6,222) COMMAND
0616 DO IVALUE=1,NUNITS
0617 STATUS=UNIT_SELECT(UNUM(IVALUE))
0618
0619 Enable programming of output mode
0620 INSTR=IT_CONTROL+I_IO+IBIT_3
0621 CALL LOAD_CTRLW_HS(INSTR,11,12)
0622
0623 Select EDIT output mode
0624 INSTR=IT_OMODE+I_I2+IBIT_1
0625 CALL LOAD_CTRLW_HS(INSTR,11,12)
0626
0627 Return to output mode = OFF

```

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BIOB100V2
C
0628 INSTR=IT_OMODE+1_12
0629 CALL LOAD_CTRLW_HS(INSTR,11,12)
0630
C
0631 Return control of output mode to front panel
C
0632 INSTR=IT_CONTROL+1_10
0633 CALL LOAD_CTRLW_HS(INSTR,11,12)
0634
C
0635 ENDDO
0636 GO TO 200
0637
C
0638 ENDIF
0639
C
0640 Unknown command
C
0641
C
0642 WRITE (6,245) COMMAND
0643 FORMAT (' UNKNOWN COMMAND: ',A)
0644 GO TO 200
0645
C
0646 Bump process to REALTIME priority
0647
C
0648 STATUS=SYSETPRI(.,%VAL(31).)
0649 IF (.NOT. status)
0650 1 CALL FATAL_ERROR (status, 'SETTING RT PRI0')
0651
C
0652 Perform the data transfer
0653
C
0654 WRITE (6,222) 'DATA TRANSFER'
0655 DO IVALUE=1,NUNITS
0656 WRITE (6,234) UNUM(IVALUE),IVALUE,NUNITS
0657 FORMAT (' Selecting unit ',i11,' of ',i11,
0658 1 STATUS=UNIT_SELECT(UNUM(IVALUE))
0659
C
0660 Enable programming of output mode
0661
C
0662 INSTR=IT_CONTROL+1_10+IBIT_3
0663 CALL LOAD_CTRLW_HS(INSTR,11,12)
0664
C
0665 Select EDIT output mode
0666
C
0667 INSTR=IT_OMODE+1_12+IBIT_1
0668 CALL LOAD_CTRLW_HS(INSTR,11,12)
0669
C
0670 Data transfer:
0671
C
0672 Enable digital output and transfer the digital memory
0673 to the host
C
0674 IF (.NOT. SDEBUE) THEN
0675 STATUS=BIO_SYNCH(IDATA(1+(IVALUE-1)*2048),2030)
0676 ELSE
0677 STATUS=BIO_IN(IDATA(1+(IVALUE-1)*2048),2030)
0678 ENDIF
0679
C
0680 Return to output mode = OFF
0681
C
0682
C
0683
C
0684

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8-Sep-1983 12:36:51  
29-Aug-1983 13:16:22

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0685 INSTR=IT_ONMODE+I_I2
0686 CALL LOAD_CTRLW_MB(INSTR,I1,I2)
0687
0688 Return control of output mode to front panel
0689
0690 INSTR=IT_CONTROL+I_I0
0691 CALL LOAD_CTRLW_MB(INSTR,I1,I2)
0692
0693 ENDDO
0694
0695 Drop back to normal priority
0696
0697 STATUS=SYSSETPRI(,XVAL(4),RTPRIOR)
0698 IF (.NOT. STATUS)
0699 1 CALL FATAL_ERROR (STATUS, 'CLRING RT PRIORITY')
0700
0701 Properly sign extend the 8 bit values to 16 bits
0702
0703 DO I=1,IBUFSIZ
0704   IDX=(I*2)-1
0705   BVAL=8BUF(IDX)
0706   IDATA(I)=BVAL
0707
0708 ENDDO
0709
0710 APPLY GAIN AND OFFSET CORRECTIONS
0711
0712 DO J=1,NUNITS
0713   DO I=1,2048
0714     IZLCH=UNUM(J)
0715     DO ICRUD=1,4
0716       IF (ONOSIDX(ICRUD).EQ. IZLCH) THEN
0717         ICR=ICRUD
0718       ENDIF
0719     ENDDO
0720     IDX=(J-1)*2048+I
0721     TVAL=FLOAT(IDATA(IDX))
0722     TVAL=(TVAL+OFFSET(ICR))*GAIN(ICR)
0723     IDATA(IDX)=NINT(TVAL)
0724   ENDDO
0725
0726 If correlation checking is disabled, save the raw data
0727
0728 IF ( NOT CORRCHK) THEN
0729   DO I=1,10240
0730     TESTDATA(I)=IDATA(I)
0731   ENDDO
0732
0733 ENDDO
0734
0735 Validate the two channel synchronization and merge the
0736 multi-unit records
0737
0738 (Overlapped portions should overlay exactly)
0739
0740 DO I=1,NUNITS-1
0741   MO=I*2048+1
0742   MI=I*2048+1

```

```

0742 DO J=-3,3
0743   CALL CORREL(IDATA(MO+4), IDATA(M1+J+4), 10, CORCO(J+6))
0744 ENDDO
0745 C D
0746   WRITE (6,314) UNUM(1), UNUM(1+1)
0747   FORMAT (' ', 111, '-', 111)
0748 C
0749   Highest correlation will occur at best match
0750   VMAX=-100000.
0751   IOS=0
0752   DO K=-5,5
0753 C d
0754   WRITE (6,315) K, CORCO(K+6)
0755   FORMAT (' ', 112, ' ', F)
0756   IF (VMAX LT CORCO(K+6)) THEN
0757     VMAX=CORCO(K+6)
0758     IOS=K
0759   ENDIF
0760 ENDDO
0761 C
0762   If the best match yielded a low correlation, then the data
0763   was probably not properly synched on acquisition
0764   IF (VMAX LE 0.0) THEN
0765     write (6,316) unum(1), unum(1+1)
0766     FORMAT (' No correlation between units ', 2(1x, 111))
0767     GO TO 200
0768   ENDIF
0769   M1=M1+IOS
0770   write (6,315) ios, vmax
0771   WRITE (6,318) (IDATA(J), J=MO, MO+13)
0772   WRITE (6,318) (IDATA(J), J=M1, M1+13)
0773   FORMAT (' ', 1415)
0774   DO L=10, 2024
0775     IDATA(MO+L)=IDATA(M1+L)
0776   ENDDO
0777   INVALID=.FALSE.
0778   IF (INVALID) THEN
0779     WRITE (6,320) UNUM(1), UNUM(2)
0780     FORMAT (' Channel synch lost unit ', 112, '-', 112)
0781     NVALCNT=NVALCNT+1
0782     GO TO 250
0783   ENDIF
0784 C
0785   Plot the entire buffer
0786 C
0787   IF (IGRAF EQ 1) THEN
0788     CALL INIT(120)
0789     NUMPTS=(NUMITS*2024)/NCHANS
0790     CALL BINITT
0791     CALL NPTS(NUMPTS)
0792     CALL XNEAT(O)
0793     CALL YNEAT(O)
0794     IF (NCHANS EQ 1) THEN
0795       DO I=1, NUMPTS
0796         XVALUES(I)=FLOAT(I-1)
0797         YVALUES(I)=FLOAT(IDATA(I))
0798       END DO
0799     END IF

```

```

0799      ENDDO
0800      YVALUES(1)=128.
0801      YVALUES(2)=-128.
0802
0803      ELSE
0804
0805          DO I=0,NUMPTS-1
0806              XVALUES(I+1)=FLOAT(1)
0807              YVALUES(I+1)=IDATA(2+I*2)
0808
0809          ENDDO
0810          YVALUES(1)=128.
0811          YVALUES(2)=-128.
0812
0813      ENDIF
0814      CALL SLIMX(150,850)
0815      CALL SLIMY(245,420)
0816      CALL CHECK(XVALUES,YVALUES)
0817      CALL DISPLAY(XVALUES,YVALUES)
0818      IF (NCHANS.EQ.2) THEN
0819          DO I=0,NUMPTS-1
0820              XVALUES(I+1)=FLOAT(1)
0821              YVALUES(I+1)=FLOAT(IDATA(1+I*2))
0822
0823          ENDDO
0824          YVALUES(1)=128.
0825          YVALUES(2)=-128.
0826      CALL BINITT
0827      CALL NPTS(NUMPTS)
0828      CALL XNEAT(0)
0829      CALL YNEAT(0)
0830      CALL SLIMX(150,850)
0831      CALL SLIMY(10,185)
0832      CALL XLAB(0)
0833      CALL CHECK(XVALUES,YVALUES)
0834      CALL DISPLAY(XVALUES,YVALUES)
0835
0836      ENDIF
0837      CALL FINITT(0,700)
0838
0839      ENDDO
0840
0841      Open the file and write the data to disk
0842
0843      IF (SDEBUQ) GO TO 200
0844      CALL BUILD_FLNM(FILENAME,IFIRST,ILEN)
0845      NBLKS=NUMPTS*8
0846      OPEN (FILE=FILENAME(1:ILEN),UNIT=1,STATUS='NEW',
0847            1 ACCESS='DIRECT',RECL=128,ERR=500)
0848      WRITE (1,1) (ILSHDR(J),J=1,128)
0849      IF (CORRCHK) THEN
0850          DO I=1,NBLKS
0851              WRITE (1,I+1) (OBUF(J,1),J=1,128)
0852          ENDDO
0853      ELSE
0854          DO I=1,NBLKS
0855              WRITE (1,I+1) (TBUF(J,1),J=1,128)
0856          ENDDO
0857      ENDIF
0858      INQUIRE (UNIT=1,NAME=FULL_FILE)
0859      STATUS=STRIM(FULL_FILE,FULL_FILE,ILEN)
0860      CLOSE (UNIT=1)
0861      WRITE (6,410) FULL_FILE(1:ILEN)
0862      FORMAT (' File generated. ',A)
0863
0864      410

```

BIOB100V2

```

0854 IFIRST=IFIRST+INCR
0857 IF (IFIRST.GT.9998) THEN
0858   WRITE (6,420)
0859   FORMAT (' NO MORE WD NUMBERS AVAILABLE')
0860   CALL EXIT
0861 ENDIF
0862 GO TO 200
0863
0864 Error opening desired file.
0865
0866 Two possible events:
0867 1. File already exists. Warn the user and then
0868   open a new version (not ILS compatible)
0869 2. Any other error. Fatal error.
0870
0871 INQUIRE (FILE=FILENAME(1:ILEN),
0872           EXIST=SEXTISTS)
0873 IF (SEXTISTS) THEN
0874   WRITE (6,510) 7,FILENAME(1:ILEN)
0875   FORMAT (' ',1A1,'WARNING! PRE-EXISTING ',
0876         'FILE ',A)
0877   WRITE (6,520)
0878   FORMAT (' NEW VERSION BEING CREATED!')
0879   ILEN=ILEN-1
0880   OPEN (FILE=FILENAME(1:ILEN),UNIT=1,STATUS='NEW',
0881         ACCESS='DIRECT',RECL=128,ERR=500)
0882   GO TO 405
0883 ELSE
0884   STOP 'ERROR - CANNOT OPEN OUTPUT FILE'
0885 ENDIF
0886
0887 Normal termination
0888
0889 IF (IORAF.EQ.1) THEN
0890   ENDIF
0891   CALL EXIT
0892   END

```







```
0000 1 TITLE BIOSUBSV2
0000 2 IDENT /V2-3/
0000 3
0000 4 $SSDEF
0000 5
0000 6
0000 7
0000 8 I/O subroutines for software interface between VAX and
0000 9 Biomation 8100 transient recorders
0000 10
0000 11
0000 12
0000 13 Symbolic offsets into IO space
0000 14
0000 15
0000 16
0000 17
0000 18
0000 19
0000 20 DR11_CSR = ^X1FB
0000 21 DR11_OBUF = DR11_CSR+2
0000 22 DR11_IBUF = DR11_CSR+4
0000 23
0000 24
0000 25
0000 26
0000 27
0000 28
0000 29
0000 30
0000 31
0000 32
0000 33
0000 34
0000 35
0000 36
0000 37
0000 38
0000 39
0000 40
0000 41
0000 42
0000 43
0000 44 DR11M_REQB = ^XB000
0000 45 DR11M_REGA = ^XB0
0000 46 DR11M_INTENA = ^X40
0000 47 DR11M_INTENB = ^X20
0000 48 DR11M_CSRI = ^X2
0000 49 DR11M_CSRO = ^X1
0000 50 DELAY = 100
0000 51
0000 52
0000 53
0000 54
0000 55
0000 56
0000 57
```

000001FB  
000001FA  
000001FC

00008000  
00000080  
00000040  
00000020  
00000002  
00000001  
00000064

DR11-C CSR is at UNIBUS address 767770. That page  
begins at address 767000. DR11 CSR is offset  
by 770 (1FB hex) from the base address of the  
page

DR11-C CSR bit assignments

15	14	13	12	11	10	09	08	07	06	05	04	03	02	01	00
R	E	G	U	S	T	B		R	I	I			C	C	
								E	N	N			S	B	
								Q	T	T			R	R	
								U	E	E			I	O	
								S	N	N					
								T	A	B					
								A							

BIOMATION 8100 CONTROL WORD FORMAT

15	14	13	12	11	10	09	08	07	06	05	04	03	02	01	00

```
0000 58 ; U U I I I I D D D D D D D
0000 59 ; N N N N N N A A A A A A A
0000 60 ; I I I S S S S T T T T T T T
0000 61 ; T T T T T T A A A A A A A
0000 62 ; N N N T T T R R R
0000 63 ; R R R Y Y Y C C C
0000 64 ; P P P T T T
0000 65 ; E E E I I I
0000 66 ;
0000 67 ;
0000 68 ;
0000 69 ;
0000 70 ;
0000 71 ;
0000 72 ;
0000 73 ;
0000 74 ;
0000 75 ;
0000 76 ;
00000000 77 ;
00000400 78 ;
0000 79 ;
0000 80 ;
0000 81 ;
0000 82 ;
0000 83 ;
00000000 84 ;
0000 85 ;
0000 86 ;
0000 87 ;
0001 88 ; CSR_SET0: WORD <DR11EM_CSR0> ;(<DR11EM_CSR1>)
0002 89 ; CSR_SET1: WORD DR11EM_CSR1
0000 90 ; CSR_RESET: WORD 0
0000 91 ; UNIT_NR: WORD 0
0000 92 ;
0000 93 ; OUTPUT_ENABLE: WORD ~B0000101100000000
0000 94 ;
0000 95 ;
0000 96 ;
0000 97 ;
00000000 98 ; PSECT CODE PIC, SHR, NOWRT, LONG
0000 99 ;
0000 100 ;
0000 101 ;
0000 102 ; ENTRY BIO_CODE ~MC>
0002 103 ;
0002 104 ; Return the address range of i/o access code.
0002 105 ; This code will be locked into memory during
0002 106 ; operation
0002 107 ;
0002 108 ; MOVL CODE_START, E4(AP)
000A 109 ; MOVL CODE_END, E8(AP)
0012 110 ; MOVL #SS$NORMAL, R0
0015 111 ; RET
0016 112 ;
0016 113 ;
0016 114 ;
```

**A-26**

```

0000 0090 172  ENTRY UNIT_SELECT ^M<
      0092 173  MOVW  @4(AP), RO
      0096 174  ASHL  #13, RO, RO
      009A 175  MOVW  RO, UNIT_NR
      00A1 176  MOVW  #SS$NORMAL, RO
      00A4 177  RET
      00A5 178
      00A5 179
      00A5 180
      00A5 181
      00A5 182
      00A5 183
      00A5 184
      00A5 185
      00A5 186
      00A5 187
      00A5 188
      00A5 189
      00A5 190
      00A5 191
      0000 00A7 192
      00000000'EF DE 00A7 193
      50 04 BC D0 00AE 194
      00000004'EF D0 00B2 195
      00B9 196
      00B9 197
      00B9 198
      50 00000006'EF AB 00B9 199
      00C0 200
      00C0 201
      00C0 202
      08 BC 30 3C 00C0 203
      00C4 204
      00C4 205
      00C4 206
      01FA C1 50 B0 00C4 207
      00C9 208
      00C9 209
      00C9 210
      50 00000000'EF B0 00C9 211
      01FB C1 50 B0 00D0 212
      00000004'EF B0 00D5 213
      01FB C1 50 B0 00DC 214
      00E1 215
      00E1 216
      00E1 217
      50 00000002'EF B0 00E1 218
      01FB C1 50 B0 00EB 219
      00ED 220
      00ED 221
      00ED 222
      56 01FB C1 B0 00ED 223 120$
      56 8000 BF B3 00F2 224
      03 12 00F7 225
      F1 52 F5 00F9 226
      228 130$

```

ENTRY UNIT\_SELECT ^M<

MOVW @4(AP), RO

ASHL #13, RO, RO

MOVW RO, UNIT\_NR

MOVW #SS\$NORMAL, RO

RET

Load a single control word and wait for response

CALL LOAD\_CTRLW\_HS(IWORDIN, IWORDOUT, IDUMMY)

IWORDIN is the command (less address field) to be loaded

IWORDOUT is the actual command (including address) that was loaded

IDUMMY is the timeout count (not normally used)

ENTRY LOAD\_CTRLW\_HS ^M<

MOVAL IOPAGE, R1

MOVL @4(AP), RO

MOVL #DELAY, R2

"OR" in the unit address to each command word

BISW UNIT\_NR, RO

Return the actual control word to the calling program

MOVZWL RO, @B(AP)

Load the command word

MOVW RO, DR11\_OBUF(R1)

Clear REQB flipflop by pulsing CSRO

MOVW CSR\_SET0, RO

MOVW RO, DR11\_CSR(R1)

MOVW CSR\_RESET, RO

MOVW RO, DR11\_CSR(R1)

Raise CMD signal (CSR1)

MOVW CSR\_SET1, RO

MOVW RO, DR11\_CSR(R1)

Wait for REQUEST B

MOVW DR11\_CSR(R1), R6

BITW #DR11\$M\_REQB, R6

BNEQU 130\$

SUBQTR R2, 120\$



```

33 08 BC D0 0145 286
    0149 287
    0149 288
    0149 289
    50 0000000A'EF B0 0149 290
    0150 291
    0150 292
    0150 293
    50 00000006'EF AB 0150 294
    0157 295
    0157 296
    0157 297
    0157 298
    01FA C1 50 0157 299
    015C 300
    015C 301
    015C 302
    015C 303
    015C 304
    015C 305
    55 01FB C1 B0 015C 306
    55 0080 BF B3 0161 307
    000001BA'EF 17 0166 308
    1C 12 016C 309
    016E 310
    016E 311
    016E 312
    016E 313
    016E 314
    016E 315
    01FB C1 50 B0 0175 316
    50 FFFF BF B0 017A 317
    01FA C1 50 B0 017F 318
    50 20D4 BF B0 0184 319
    04 B0 0189 320
    018A 321
    018A 322
    018A 323
    018A 324
    018A 325
    018A 326
    018A 327
    018A 328
    018A 329
    018A 330
    018A 331
    018A 332
    01FB C1 50 B0 0191 333
    50 00000004'EF B0 0196 334
    01FB C1 50 B0 019D 335
    50 00000002'EF B0 01A2 336
    01FB C1 50 B0 01A9 337
    01AE 338
    01AE 339
    01AE 340
    01AE 341
    50 00000064 BF D0 01AE 342

    MOVL 8B(AP), R3
    Enable digital output
    MOVW OUTPUT_ENABLE, R0
    "OR" in the unit address to each command word
    BISH UNIT_NR, R0
    Load the command word
    MOVW R0, DR11_OBUF(R1)
    Selected unit should be in digital output mode
    with the first data value ready. Check the
    status of the device
    Check "READY" bit in CSR (actually FLO signal)
    MOVW DR11_CSR(R1), R5
    BITW #DR11EM_REGA, R5
    jmp 300$
    BNEGU 300$
    Device not ready
    Select unit 7 (not normally used) and clear CSR
    MOVW CSR_RESET, R0
    MOVW R0, DR11_CSR(R1)
    MOVW #-1, R0
    MOVW R0, DR11_OBUF(R1)
    MOVW #SS6_DEVINACT, R0
    RET
    Transfer the first data value to the user buffer
    MOVW DR11_OBUF(R1), R5
    MOVW R5, (R2)+
    Clear REQ8 flipflop by pulsing CSRO
    Raise CMD signal (CSR1)
    MOVW CSR_SET0, R0
    MOVW R0, DR11_CSR(R1)
    MOVW CSR_RESET, R0
    MOVW R0, DR11_CSR(R1)
    MOVW CSR_SET1, R0
    MOVW R0, DR11_CSR(R1)
    Wait for REQUEST B
    MOVW #DELAY, R0

```

56	FD 50	F5	01B5	343	321\$	SOBQTR	RO, 321\$
56	01FB C1	B0	01B5	344		MOVW	DR11_CSR(R1), R6
56	8000 8F	B3	01B8	345		BITW	#DR11\$M_REQB, R6
	EA	13	01C2	347		BEGLU	320\$
			01C4	348			
			01C4	349			BIO 8100 has responded. Store data value.
			01C4	350			
55	01FC C1	B0	01C4	351		MOVW	DR11_IBUF(R1), R5
	82 55	B0	01C9	352		MOVW	R5, (R2)+
			01CC	353			Reset CMD bit
			01CC	354			
			01CC	355			
50	00000004'EF	B0	01CC	356		MOVW	CSR_RESET, RO
	01FB C1	50	01D3	357		MOVW	RO, DR11_CSR(R1)
	AF 53	F4	01D8	358		SOBQEQ	R3, 310\$
			01D8	359			Clear CSR
			01D8	360			
			01D8	361			
50	00000004'EF	B0	01D8	362		MOVW	CSR_RESET, RO
	01FB C1	50	01E2	363		MOVW	RO, DR11_CSR(R1)
	50 01	B0	01E7	364		MOVW	#SS\$_NORMAL, RO
		04	01EA	365		RET	
			01EB	366			
			01EB	367			CODEND:
			01EB	368			
			01EB	369			END



**Appendix B**  
**ILS SOFTWARE MODULES**

## Appendix B

### ILS SOFTWARE MODULES

The ILS Interactive Laboratory System™ interactive digital signal processing software package (available through software leasing agreements from Signal Technology Inc., Santa Barbara, California) was used as the primary software tool in developing the signal processing software used in this study. The basic ILS package supplies a very flexible and powerful set of modular software components, implemented in an interactive environment featuring considerable graphics feedback. The ILS software provides many generic digital signal processing modules useful in a wide variety of application areas. When we found that ILS did not provide the necessary functionality needed in the acoustic emission application, new ILS modules were written, or modifications were made to existing ILS modules to implement the new functionality. These software modules and documentation are included on the following pages of this Appendix.

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RESEARCH AND DEVELOPMENT SCHEMECTA. J W ERKES ET AL.  
DEC 84 845RD002 N00014-82-C-2031 F/G 20/1

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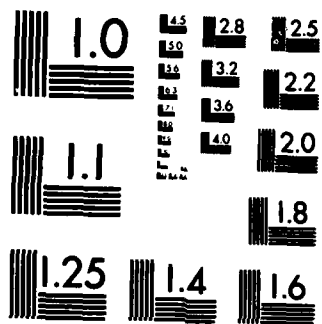
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F/G 20/1

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END



MICROCOPY RESOLUTION TEST CHART  
NATIONAL BUREAU OF STANDARDS-1963-A

Existing ILS commands as supplied by STI:

ADF	ARITHMETIC FUNCTIONS ON INTEGER DATA
AFP	3-D AREA FUNCTION PLOT OF ANALYSIS DATA
ANL	LINEAR PREDICTION ANALYSIS USING AUTOCORRELATION, COVARIANCE, OR BURG'S METHOD
API	INVERSE FILTER ANALYSIS USING AUTOCORRELATION METHOD WITH PITCH DETECTION USING CEPSTRAL METHOD
ASG	ASSIGN LOGICAL UNIT NUMBER FOR KEYBOARD
AVG	RECORD AVERAGING (MOVING OR EXPONENTIAL) INPUT IS PRIMARY FILE, OUTPUT SECONDARY
BOP	BINARY ARITHMETIC OPERATIONS ON FLOATING POINT DATA; OR COMBINE TWO REAL VECTORS INTO ONE COMPLEX VECTOR - TWO FILES INPUT, ONE OUTPUT
BPA	PATTERN MATCHING OF TEST AND REFERENCE DATA SUPPORTS DYNAMIC PROGRAMMING FOR APPLICATIONS SUCH AS WORD RECOGNITION
CEP	CEPSTRUM DISPLAY OF VARIABLE LENGTH FRAMES
CLA	CURSOR LABELING OF PERIODIC SAMPLED DATA
CNV	CONVOLUTION OF A LONG DURATION TIME SERIES WITH A FILTER IMPULSE RESPONSE USING SHORT OVERLAPPED FFT'S
COR	CORRELATION OF TWO LONG DURATION TIME SERIES USING SHORT OVERLAPPED FFT'S
CST	COMPUTATION OF SAMPLED DATA STATISTICS; PRINTS LOCATION OF PEAK VALUE
CTX	EXAMINE OR CHANGE CONTEXT, WHICH IS THE NUMBER OF SAMPLED DATA POINTS PER FRAME OF DATA
CUR	X-Y CURSOR, USED IN CONJUNCTION WITH DISPLAY
DPM	DISPLAY OF ANALYSIS DATA; SUPPORTS VARIABLE FRAME SIZES
DRE	DISPLAY REAL OR COMPLEX VECTORS FROM RECORD FILES
DSP	DISPLAY OF TIME SERIES; GRIDS AND FRAME LOCATIONS
EPI	ELLIPTIC, BUTTERWORTH AND CHEBYCHEV FILTER DESIGN
FDI	FREQUENCY SPECTRUM DISPLAY, FROM FFT ON SAMPLED DATA (MAXIMUM OF 512 POINTS)
FFT	FAST FOURIER TRANSFORM OF A TIME SERIES; SUPPORTS CIRCULAR SHIFTS OF THE DATA
FIL	SPECIFY, CREATE, DELETE BINARY DATA FILES
FLT	CONVOLUTION OF A TIME SERIES WITH DIGITAL FILTERS; SUPPORTS INTEGER OR FLOATING POINT DATA FILES
FPL	FREQUENCY PLOT OF SMOOTH SPECTRA FROM ANALYSIS FILE
FTR	FORMANT TRACKING OF SPEECH DATA; STORES BANDWIDTH INFORMATION AND SUPPORTS VARIABLE FRAME SIZES
GRD	PLOT A GRID FOR FREQUENCY OR AN AXIS FOR TIME
HHH	HELP WITH ILS COMMANDS
HIS	READ DATA FROM RECORD FILES AND PLOT A HISTOGRAM ON THE SCREEN
IDC	ENTER DATA INTO COMMON BLOCK

IFL IDEAL FILTER DESIGN; SUPPORTS CIRCULAR SHIFT  
 OF IMPULSE RESPONSE  
 ILS CREATE AND INITIALIZE USER COMMON FILE  
 INA INITIALIZATION OF SAMPLED DATA FILE HEADER;  
 SUPPORTS SPECIFICATION OF LOGICAL END OF FILE  
 WHICH IS USED IF STARTING FRAME =0 AND NUMBER  
 OF FRAMES=1 IN SUBSEQUENT PROGRAMS  
 LBA LABEL A SAMPLED DATA SEGMENT  
 LBF POINT TO A LABEL FILE  
 LCM LIST ILS COMMON (ALL OR PART)  
 LFI DESIGNS LINEAR PHASE FINITE IMPULSE RESPONSE  
 (FIR) FILTERS USING THE REMEX EXCHANGE ALGORITHM  
 LLA LISTS LABEL FILE  
 LRE LISTING OF FLOATING POINT DATA; INDEXES TIME  
 SERIES IN SECONDS AND SPECTRAL DATA IN HERTZ  
 LSN DIGITAL-TO-ANALOG CONVERSION \*NOT IMPLEMENTED\*  
 MDF MODIFY VALUES OF DATA POINTS IN FILES  
 WILL MODIFY FILE HEADERS AND RECORD HEADERS  
 MDX MULTIPLEXING AND DEMULTIPLEXING OF MULTI-  
 CHANNEL SAMPLED DATA FILES  
 MRE MANIPULATE RECORD FILES BY EXTRACTING RECORDS,  
 ITEMS AND ELEMENTS  
 MVF DATA MOVING WITHIN A FILE; SUPPORTS ANALYSIS  
 FRAMES AND ZEROING OF ORIGINAL DATA  
 NSI SIMULATION OF NOISY DATA BY ADDING NOISE TO A  
 SIGNAL; GENERATES PSEUDORANDOM NOISE  
 OPN ALLOCATES AND OPENS RECORD FILES  
 PAN PITCH SYNCHRONOUS ANALYSIS OF SAMPLED DATA  
 PCO PRINCIPAL COMPONENTS ANALYSIS  
 PLR SCATTER-PLOT OF DATA ELEMENTS IN RECORD FILES  
 PNS PITCH SYNCHRONOUS SYNTHESIS; INCLUDES IMPROVED  
 ALGORITHMS WITH GLOTTAL PULSE OR GAUSSIAN  
 NOISE EXITATION  
 PRT PRINT FROM BINARY FILES (OR HEADER)  
 QUR QUEUEING OF FEATURES IN TIME SERIES BASED ON  
 LABEL INFORMATION FOR PATTERN RECOGNITION.  
 INCLUDES GENERATION OF FEATURE MATRICES FOR  
 APPLICATIONS SUCH AS PATTERN RECOGNITION.  
 RAN RESIDUE ANALYSIS USING COEFFICIENTS FROM AUTO-  
 CORRELATION OR COVARIANCE ANALYSIS  
 REC ANALOG-TO-DIGITAL CONVERSION INTO A FILE  
 IS NOT IMPLEMENTED  
 EQUIVALENT HARDWARE AND SOFTWARE EXISTS  
 FOR SOME APPLICATIONS  
 RSO ROOT SOLVING FOR SPECTRAL RESONANCES; STORES  
 REAL ROOTS FOR HIGH QUALITY SPEECH SYNTHESIS  
 RVR REVERSE THE ORDER OF SAMPLED DATA POINTS  
 SDE SPECTRAL-DENSITY ESTIMATION COMMAND  
 (CROSS OR AUTO)  
 SDI 3-DIMENSIONAL DISPLAY OF SPECTRA; SUPPORTS  
 INTEGER TIME SERIES, FLOATING POINT TIME  
 SERIES OR FLOATING POINT SPECTRA INPUT FILES  
 SGM SPECTROGRAM DISPLAY OF ANALYSIS DATA; SUPPORTS  
 VARIABLE FRAME SIZES AND ERASES BEFORE DISPLAY

SIF	FUNDAMENTAL FREQUENCY EXTRACTION USING SIFT ALGORITHM
SME	CALCULATES STATISTICS OF DATA IN FEATURE RECORDS
SNS	SYNTHESIZE SAMPLED DATA FROM ANALYSIS DATA
SPL	3-DIMENSIONAL SPECTRAL PLOT OF ANALYSIS DATA; SUPPORTS VARIABLE FRAME SIZES
SRE	STORE RECORDS INTO SECONDARY FILE FROM KEYBOARD INPUT, PRIMARY SAMPLED DATA FILE OR PRIMARY ANALYSIS FILE
SSP	FREQUENCY SPECTRUM DISPLAY FROM FFT OF INVERSE FILTER COEFFICIENTS
TBL	SET UP DIRECTORY TABLE FOR ILS FILE SYSTEM
TFU	PROGRAM FOR CREATING TEST DATA IN A FILE (FILE MUST BE CREATED FIRST)
TLA	COPY LABELS MEETING A GIVEN SPECIFICATION INTO A SECONDARY LABEL FILE
TRE	TRANSFER OF FLOATING POINT DATA; SUPPORTS CHANGES IN RECORD SIZE
TRF	DATA TRANSFER BETWEEN FILES; SUPPORTS ANALYSIS FRAMES
TRM	SETTING OF TERMINAL CHARACTERISTICS; SUPPORTS NUMBER OF GRAPHIC INPUT TERMINATORS AND THE HP2648 TERMINAL IN NATIVE MODE. ALSO PRINTS SYSTEM TIME AND DATE AND WILL LABEL A PLOT. WILL CLEAR SCREEN OR PUT TERMINAL IN ALPHA MODE
TSI	TEST SIGNAL, GENERATE SAMPLED DATA, LABEL OR RECORD FILES.
TTL	TRANSFERS MARKED SECTION OF SAMPLED DATA TO THE SECONDARY FILE WITH OPTIONAL LABELING
UOP	UNARY OPERATIONS, PHASE UNWRAPPING OF FFT DATA; USES A MUCH IMPROVED ALGORITHM. MANY OTHER MANIPULATIONS OF FLOATING POINT DATA
VDI	VARIABLE DISTANCE THRESHOLD EVALUATION
VER	VERIFY HEADER BLOCK IN SAMPLED DATA OR ANALYSIS FILE
VTR	PLOT A VOCAL TRACT FROM SECONDARY ANALYSIS FILE OR USER'S COMMON
XPA	EXPAND, INTERPOLATE OR DOWNSAMPLE PRIMARY SAMPLED DATA FILE FOR HIGHER OR LOWER SAMPLING FREQUENCIES, BY INTERLEAVING ZEROS WITH THE ORIGINAL DATA OR SKIPPING DATA
XTR	READ DATA FROM PRIMARY RECORD FILE AND COMPUTE MAXIMUM OR MINIMUM VALUES

\*\*\*\*\*  
 New ILS commands developed and used in the course of this research.  
 \*\*\*\*\*

\$ APF - MERGES GRAPHIC OUTPUT FILES..WILL ALTERNATELY  
 MERGE INTO A USER CHOSEN FILE NAME.  
 \$ AMP - HISTOGRAMS WITH STATISTICS FOR SAMPLED DATA FILES

\$ CLR     PUTS AN ADM TERMINAL IN ALPHA MODE THEN  
           CLEARS SCREEN AFTER GRAPHICS SESSION  
 \$ CPF     - REASSIGNS GRAPHIC OUTPUT TO THE TERMINAL  
 \$ GRA     - PUTS AN ADM TERMINAL IN GRAPHICS MODE  
           THEN CLEARS SCREEN.....  
 \$ CXP     - MODIFICATION OF CEP TO WRITE DATA WHICH WAS  
           PREVIOUSLY PLOTTED, TO A FILE.

\$ HED     - LIST CONTENTS OF THE SPS PORTION OF ILS HEADER  
 \$ HFL     - APPEND HEADER TO YOUR DATA  
 \$ HPF     - PROCESSES GRAF.TMP FILES RECOGNIZING FORM FEED  
           NOT TO BE USED WITH OVERLAYS. DEVELOPED FOR  
           USE WITH HSP.  
 \$ HSP     - HISTOGRAM OF DATA PEAKS AND ENERGY CONTENT OF  
           DATA FILES.  
 \$ ITN     - INSERTS TEST NAME, COMMENTS AND ISPS HEADER  
           VALUES INTO ISPS PORTION OF HEADER  
 \$ LAB     - LABELS FOR PLOTS. CREATES A GRAF.TMP FILE TO  
           BE MERGED TO PROVIDE HORZ \_VERT LABELING.  
 \$ MST     - WRITES STATISTICS FROM SAMPLED DATA FILES  
           AS FROM ILS COMMAND "\$ CST", TO THE FILE  
           "CSTAT.TMP" WITH LEGEND AND HEADINGS.  
           SUBSEQUENT USES APPEND DATA.  
 \$ OPF     - ASSIGNS GRAPHIC OUTPUT TO A FILE NAMED GRAF.TMP  
 \$ PPF     - PROCESSES AND PRINTS GRAF.TMP FILES  
 \$ RR      - LISTS COMMANDS WHICH USE RECORD DATA  
 \$ RST     - WRITES STATISTICS FROM RECORD DATA FILES  
           AS FROM ILS COMMAND "\$ CST", TO THE FILE  
           "CSTAT.TMP" WITH LEGEND AND HEADINGS.  
           SUBSEQUENT USES APPEND DATA.  
 \$ RTS     - RECORD DATA TO SAMPLED DATA FOR DISPLAY PURPOSES  
 \$ STN     - SEARCHES A DIRECTORY FOR WD FILES FROM A  
           GIVEN TEST, ALTERNATELY USES WD NUMBERS FROM  
           A FILE.  
 \$ TRM     - TRM M WILL PRINT VERTICALLY IF N1 IS NEGATIVE  
 \$ WIN     - APPLIES FULL OR PARTIAL HANNING WINDOW  
           TO SAMPLED DATA  
 \$ WTN     - WRITES ISPS PORTION OF FILE HEADER TO  
           SECONDARY FILE  
  
 \$ XCP     - COMPUTES COMPLEX CEPSTRUM FROM INPUT WAVEFORM,  
           (INCLUDES EXPONENTIAL WEIGHTING, AND BANDPASS  
           MAPPING)

\*\*\*\*\*  
 Specialized Interactive "Recipes" using ILS commands  
 \*\*\*\*\*

\*\*\*\*\*



```
*****
XCPILOW.COM - an ILS "recipe" that implements a forward complex
cepstrum, low-pass filters the cepstrum, and finally does an
inverse complex cepstrum.
*****
```

```
$! LOW-PASS CEPSTRUM AND INVERSE TRANSFORM
$ ON CONTROLY THEN GOTO CLEANUP
$ ON ERROR THEN GOTO CLEANUP
$ IF P1 .EQS. "" THEN INQUIRE P1 "INPUT FILE NUMBER?"
$ IF P2 .EQS. "" THEN INQUIRE P2 "OUTPUT FILE NUMBER?"
$ IF P3 .EQS. "" THEN INQUIRE P3 "STARTING CHANNEL NO.?"
$ IF P4 .EQS. "" THEN INQUIRE P4 "ENDING CHANNEL NO.?"
$ IF P5 .EQS. "" THEN INQUIRE P5 "CONTEXT?"
$ CTX 'P5'
$ FIL SO
CZ9996,,1
CZ9997,,1

$ FIL 'P1'
$ FIL S9996
$ XCP Z'P3','P4'
$ FIL 9996
$ FIL S9997
$ XCP IS
$ RENAME WD9997. WD'P2'.
$CLEANUP:
$ dele/NOCONFIRM WD9996.;1
```

```
*****
XCPIHIGH.COM - an ILS "recipe" that implements a forward complex
cepstrum, high-pass filters the cepstrum, and finally does an
inverse complex cepstrum.
*****
```

```
$! HIGH-PASS CEPSTRUM AND INVERSE TRANSFORM
$ ON CONTROLY THEN GOTO CLEANUP
$ ON ERROR THEN GOTO CLEANUP
$ IF P1 .EQS. "" THEN INQUIRE P1 "INPUT FILE NUMBER?"
$ IF P2 .EQS. "" THEN INQUIRE P2 "OUTPUT FILE NUMBER?"
$ IF P3 .EQS. "" THEN INQUIRE P3 "U.P. OF 1st BAND?"
$ IF P4 .EQS. "" THEN INQUIRE P4 "L.B. OF 2nd BAND?"
$ IF P5 .EQS. "" THEN INQUIRE P5 "CONTEXT?"
$ CTX 'P5'
$ FIL SO
CZ9996,,1
CZ9997,,1
CZ9998,,1

$ FIL 'P1'
$ FIL S9996
```

```

$ XCP Z1,'P3'
$ FIL 9996
$ FIL S9997
$ XCP Z'P4','P5'
$ FIL 9997
$ FIL S9998
$ XCP IS
$ RENAME WD9998. WD'P2'.
$CLEANUP:
$ dele/NOCONFIRM WD9996.;1
$ dele/NOCONFIRM WD9997.;1

```

```

*****
XCPILOW.COM - an ILS "recipe" that implements bandpass mapping
for bandpass limited signals, followed by a forward complex
cepstrum, a low-pass filtering operation on the cepstrum, and
finally an inverse complex cepstrum.
*****

```

```

$! LOW-PASS CEPSTRUM AND INVERSE TRANSFORM
$! MODIFIED FOR BAND-PASS SYSTEM
$ ON CONTROLY THEN GOTO CLEANUP
$ ON ERROR THEN GOTO CLEANUP
$ IF P1 .EQS. "" THEN INQUIRE P1 "INPUT FILE NUMBER?"
$ IF P2 .EQS. "" THEN INQUIRE P2 "OUTPUT FILE NUMBER?"
$ IF P3 .EQS. "" THEN INQUIRE P3 "STARTING CHANNEL NO.?"
$ IF P4 .EQS. "" THEN INQUIRE P4 "ENDING CHANNEL NO.?"
$ IF P5 .EQS. "" THEN INQUIRE P5 "CONTEXT?"
$ CTX 'P5'
$ FIL SO
CZ9996,,1
CZ9997,,1

```

```

$ FIL 'P1'
$ FIL S9996
$ XCP PZ'P3','P4'
$ FIL 9996
$ FIL S9997
$ XCP ISP
$ RENAME WD9997. WD'P2'.
$CLEANUP:
$ dele/NOCONFIRM WD9996.;1

```

```

*****
XCPIPHIGH.COM - an ILS "recipe" that implements bandpass mapping
for bandpass limited signals, followed by a forward complex
cepstrum, a high-pass filtering operation on the cepstrum, and
finally an inverse complex cepstrum.
*****

```

```

$! HIGH-PASS CEPSTRUM AND INVERSE TRANSFORM
$! MODIFIED FOR BAND-PASS SYSTEM
$ ON CONTROLY THEN GOTO CLEANUP

```

```

$ ON ERROR THEN GOTO CLEANUP
$ IF P1 .EQS. "" THEN INQUIRE P1 "INPUT FILE NUMBER?"
$ IF P2 .EQS. "" THEN INQUIRE P2 "OUTPUT FILE NUMBER?"
$ IF P3 .EQS. "" THEN INQUIRE P3 "U.P. OF 1st BAND?"
$ IF P4 .EQS. "" THEN INQUIRE P4 "L.B. OF 2nd BAND?"
$ IF P5 .EQS. "" THEN INQUIRE P5 "CONTEXT?"
$ CTX 'P5'
$ FIL SO
CZ9996,,1
CZ9997,,1
CZ9998,,1

```

```

$ FIL 'P1'
$ FIL S9996
$ XCP PZ1,'P3'
$ FIL 9996
$ FIL S9997
$ XCP PZ'P4','P5'
$ FIL 9997
$ FIL S9998
$ XCP ISP
$ RENAME WD9998. WD'P2'.
$CLEANUP:
$ dele/NOCONFIRM WD9996.;1
$ dele/NOCONFIRM WD9997.;1

```

```

*****
FDECON.COM -an ILS "recipe" that implements Fourier Deconvolution
*****

```

```

$! FOURIER DECONVOLUTION
$ ON CONTROLY THEN GOTO CLEANUP
$ ON ERROR THEN GOTO CLEANUP
$ IF P1 .EQS. "" THEN INQUIRE P1 "INPUT A-FILE NUMBER?"
$ IF P2 .EQS. "" THEN INQUIRE P2 "INPUT B-FILE NUMBER?"
$ IF P3 .EQS. "" THEN INQUIRE P3 "OUTPUT FILE NUMBER?"
$ FIL S9993
$ OPN S5
$ FIL 'P1'
$ SRE 1,1
$ FIL 'P2'
$ FIL S9994
$ SRE 1,1
$ FIL 9993
$ FIL S9995
$ FFT
$ FIL 9994
$ FIL S9996
$ FFT
$ FIL 9995
$ FIL B9996
$ FIL S9997
$ BOP D
$ FIL 9997

```

```

$ FIL S'P3'
$ OPN S1
$ FFT I
$CLEANUP:
$ dele/NOCONFIRM WD9993.;1
$ dele/NOCONFIRM WD9994.;1
$ dele/NOCONFIRM WD9995.;1
$ dele/NOCONFIRM WD9996.;1
$ dele/NOCONFIRM WD9997.;1

```

```

*****
FDECONP.COM - sn ILS "recipe" that implements Fourier
Deconvolution on bandpass mapped waveforms.
*****

```

```

$! FOURIER DECONVOLUTION WITHIN A FREQUENCY BAND
$ IF P1 .EQS. "" THEN INQUIRE P1 "INPUT A-FILE NUMBER?"
$ IF P2 .EQS. "" THEN INQUIRE P2 "INPUT B-FILE NUMBER?"
$ IF P3 .EQS. "" THEN INQUIRE P3 "OUTPUT FILE NUMBER?"
$ IF P4 .EQS. "" THEN INQUIRE P4 "FREQUENCY LOWER LIMIT?"
$ IF P5 .EQS. "" THEN INQUIRE P5 "FREQUENCY UPPER LIMIT?"
$ ON CONTROLY THEN GOTO DELETECOMMAND
$ OPEN/WRITE OUT FDE.COM
$ WRITE OUT "$ ON CONTROLY THEN GOTO CLEANUP"
$ WRITE OUT "$ ON ERROR THEN GOTO CLEANUP"
$ WRITE OUT "$ FIL S9993"
$ WRITE OUT "$ OPN S5"
$ WRITE OUT "$ FIL 'P1'"
$ WRITE OUT "$ SRE 1,1"
$ WRITE OUT "$ FIL 'P2'"
$ WRITE OUT "$ FIL S9994"
$ WRITE OUT "$ SRE 1,1"
$ WRITE OUT "$ FIL 9993"
$ WRITE OUT "$ FIL S9995"
$ WRITE OUT "$ FFT"
$ WRITE OUT "$ FIL 9994"
$ WRITE OUT "$ FIL S9996"
$ WRITE OUT "$ FFT"
$ WRITE OUT "$ FIL 9995"
$ WRITE OUT "$ FIL B9996"
$ WRITE OUT "$ FIL S9997"
$ WRITE OUT "$ BOP DF"
$ WRITE OUT "$ 'P4'., 'P5'."
$ WRITE OUT "$ FIL 9997"
$ WRITE OUT "$ FIL S'P3'"
$ WRITE OUT "$ OPN S1"
$ WRITE OUT "$ FFT I"
$ WRITE OUT "$CLEANUP:"
$ WRITE OUT "$ dele/NOCONFIRM WD9993.;1"
$ WRITE OUT "$ dele/NOCONFIRM WD9994.;1"
$ WRITE OUT "$ dele/NOCONFIRM WD9995.;1"
$ WRITE OUT "$ dele/NOCONFIRM WD9996.;1"
$ WRITE OUT "$ dele/NOCONFIRM WD9997.;1"
$ CLOSE OUT

```

```
$ @FDE.COM
$DELETECOMMAND:
$ DELETE/NOCONFIRM FDE.COM;
```

```
*****
AVERG.COM - an ILS "recipe" that implements an averaging
operation on a series of candidate waveforms.
*****
```

```
$ ON CONTROLY THEN GOTO CLEANUP
$ ON ERROR THEN GOTO CLEANUP
$ IF P1 .EQS. "" THEN INQUIRE P1 "INPUT FILE?"
$! INPUT FILE CONTAINS WD NUMBERS, ONE PER LINE, OF FILES TO BE
  AVERAGED
$ IF P2 .EQS. "" THEN INQUIRE P2 "OUTPUT FILE NUMBER?"
$ OPEN/READ IN 'P1'
$ COUNT=0
$ FIL S9995
$ OPN S3
$LOOP:
$ READ/ENDOFFILE=THATSALL IN NUM
$ COUNT=COUNT+1
$ FIL 'NUM'
$ SRE 1,1
$ GOTO LOOP
$THATSALL:
$ FIL 9995
$ FIL S9996
$ AVG 1,'COUNT'
$ FIL 9996
$ FIL S9997
$ TRE 'COUNT',1
$ FIL 9997
$ FIL S'P2'
$ RTS 1,1
$CLEANUP:
$ dele/NOCONFIRM WD9995.;1
$ dele/NOCONFIRM WD9996.;1
$ dele/NOCONFIRM WD9997.;1
$ CLOSE IN
```

```
*****
Special Purpose ILS Software Modules created for this Study
*****
```

```
*****
XCP.FOR
*****
```

```

C...      INTERACTIVE LABORATORY SYSTEM
C...
C...      ILS COMMAND PROGRAM  ** XCP **
C...
C...      PROGRAM XCP
C...
C...      IMPLICIT INTEGER (I-N)
C...
C...      START OF DOCUMENTATION
C...
C...      SPECIAL VERSION TO USE ENTIRE FILE
C...
C...      COMMAND FORMAT:
C...
C...      XCP [Z,I,F][S]N1,N2
C...
C...      ALPHABETIC ARGUMENTS:
C...
C...      Z      - ZERO CEPSTRUM RECORD FROM N1 TO N2
C...      I      - INVERSE CEPSTRUM
C...      F      - CREATE TEST FILE
C...      S      - USED WITH [I], OPTIONAL PHASE SHIFT
C...      P      - PREPROCESS WITH FREQUENCY LIMIT
C...
C...      NUMERIC ARGUMENTS:
C...
C...      N1 - STARTING FRAME
C...      N2 - NUMBER OF FRAMES
C...      WITH OPTION [Z]
C...      N1 - FIRST POINT TO ZERO
C...      N2 - LAST POINT TO ZERO
C...
C...      N3 - ASK FOR ALPHA
C...      WITH OPTION [P]
C...      N4 - LOWER FREQUENCY
C...      N5 - UPPER FREQUENCY
C...
C...      END OF DOCUMENTATION
C...
C...      COMMON
C...      PI,TWOPI,THLINC,THLCON,NFFT,NDUM,NN,L,H,H1,DVTMN2,AMULT
C...      COMMON /CLBF/  INSFLG,LCLBF,ICLBF(40)
C...      COMMON /ILSA/  NBCW,NCWBK,NDPBK,NDPF,NBDP,NCWFH,
1      KBU,KBUIN,LPU,LUGI,LUGO,NSC,CWSC,
2      NBA2D,MIDA2D,ICTIM(4),ICDAT(6)
C...      COMMON /ILSB/  IA(4),N1,N2,N3,N4,N5,N6,N7,N8,N9,N10,N11,N12
C...      COMMON /ILSC/  IASAV(4),N1SAV,N2SAV,N3SAV,N4SAV,N5SAV,N6SAV,
1      N7SAV,N8SAV,N9SAV,N10SAV,N11SAV,N12SAV
C...      COMMON /ILSE/
C...      IFLPA(16),LENPA,LFILPA,NFLPA,IDKPA,IDKDPA,IDPA
C...      COMMON /ILSF/
C...      IFLSA(16),LENSA,LFILSA,NFLSA,IDKSA,IDKDSA,IDSA
C...      COMMON /ILSH/  FS,M,MP1,MO2,N,NSPBK,NSHFT,ICON,ISF,IHAM,LAN

```

```

COMMON /ILSI/ LRH,IEX,ISTAN,NAN,IDF(5),NP,IVL,IC
COMMON /ILSJ/ RC(30),A(30),R(30),F(10),B(10)
C...
COMMON /FLC1/ IFL1(16),LEN1,LFIL1,NFL1,IDK1,IDKD1,ID1
COMMON /FLCM/
IFLC(16),LENC,IFLALF(4),LENALF,IFLX,IFLV,LENCOM
C...
DIMENSION IY(16384),IR(16896),IS(16896),IX(16384)
DIMENSION Y(16384),X(16384),DUM(16384)
DIMENSION IHR(128),IHW(128)
DIMENSION IOPT(5),ICHSTR(5)
EQUIVALENCE (IHR(70),SCALR),(IHW(70),SCALW)
EQUIVALENCE (IHR(74),ACEP),(IHW(74),ALPHA)
C...
C
C...
DATA SCL1/1.0/
DATA I1,I2,I3,I4,I5,I11,I19,I30/1,2,3,4,5,11,19,30/
DATA ICHSTR(1),ICHSTR(2),ICHSTR(3),ICHSTR(4)/2HZ ,2HI ,2HS
,2HF /
DATA ICHSTR(5)/2HP /
DATA IHW(75),IHW(76)/0,0/
DATA IFLG,IIBW,IIBR,IOVERFLO/4*0/
DATA LIX,LIY,LIR,LIS/2*16384,2*16896/
DATA ICHA,ICHNP,ICHS,ICHR/2HA ,2HNP,2HS ,2HR /
DATA SCL0/0.0/
C...
CALL RCOMM
C...
C...
C...
INITIALIZATION DATA
CALL AFARG(IA,I1,I4,ICHSTR,IOPT,I5)
C...
C...
CHECK OPTIONS
C...
IF(IOPT(1).EQ.1) THEN
C...
ZERO PORTION OF CEPSTRUM
IMODE=1
ELSE IF (IOPT(2).EQ.1) THEN
C...
INVERSE CEPSTRUM
IMODE=2
ELSE IF (IOPT(4).EQ.1) THEN
C...
CREATE TEST FILE
IMODE=4
ELSE
C...
FORWARD CEPSTRUM OR PREPROCESS
IMODE=0
END IF
C...

```

```

C...  INVERSE CEPSTRUM WITH PHASE SHIFT
      IF(IMODE.EQ.2.AND.IOPT(3).EQ.1)IMODE=3
C...
C...  SET FRAMES TO DEFAULT TO WHOLE FILE
C...
      DO 10 I=1,4
      IF(IOPT(I).NE.0)GOTO 11
10    CONTINUE
C...  FIND CEPSTRUM.....SET UP NUMBER OF POINTS
C...
      IF(N1.EQ.0.AND.N2.EQ.0)THEN
          N1A=0
          N2A=1
      ELSE
          N1A=N1
          N2A=N2
      END IF
      GOTO 12
C...
C...  OPTIONS [I] AND [Z] USE THE ENTIRE FILE
11    IF(IOPT(4).EQ.1)THEN
          NPTS=2048
          LFILSA=2048/NDPBK
          LFRAM=NDPF
          NFRAM=2048/NDPF
          GOTO 260
      END IF
      N1A=0
      N2A=1
C...
12    CALL CHKFL(NFLPA,IDKPA,IFLPA,LENPA,LFILPA,IHR,ICHS,IERR)
      IF(IERR.NE.0) GO TO 310
C...
C...  GET SAMPLING FREQUENCY
C...
      ISF=IHR(62)
      IPWR=IHR(61)
      CALL SFCNV(FS,ISF,IPWR,I1)
      CALL ANCHK(NSC,N1A,N2A,N1SAV,N2SAV,NSCA,ISTAN,
1        NAN,NDPF,NSDBK,LFILPA)
      ISTFR=N1SAV
      NFR=N2SAV
C...
C...  GET NUMBER OF POINTS.....TRUNCATE TO A POWER OF 2
C...
      CALL FTOP(N1SAV,N2SAV,NSC,NDPF,STPT,PTS,ENDPT)
      NPTS=IFIX(PTS+0.5)
      INDX=16384
20    IF(NPTS.GE.INDX)THEN
          NPTS=INDX
          GOTO 30
      ELSE
          INDX=INDX/2

```



```

      END IF
      GOTO 20

C...
C...   CONVERT BACK TO FRAMES
C...
30     PTS=FLOAT(NPTS)
      CALL PTOF(STPT,PTS,NDPF,NSC,ISTFR,NFR)
      NPTX=NPTS
      NPTY=NPTS
      IST=ISTFR
      LST=ISTFR+NFR-1
C...   SET SCALE
      IF(IMODE.EQ.1) THEN
          SCALE=1.0
      ELSE IF(IMODE.EQ.0) THEN
          SCALE=1.0
      ELSE IF(IMODE.EQ.2.OR.IMODE.EQ.3) THEN
          SCALE=SCALR
      END IF
      CALL GETD(NSC,ISTFR,NPTS,NPTY,IR,LIR,IY,IIBR,
1         IFLPA,LENPA,LFILPA)
      IF(IIBR.LT.0) GOTO 260

C...
      CALL MI2R(IY,Y,NPTS,SCALE)
200    CONTINUE
C...   ALL POINTS READ
      LFILSA=LFILPA
260    IFLGO=2
      ITYPE=-1
      CALL CHKFL(NFLSA,IDKSA,IFLSA,LENSA,LFILSA,IHW,ITYPE,IFLGO)
      ITYPE=ICHNP
      CALL SETUP(IFLSA,LENSA,LFILSA,NFRAM,LFRAM,NCWFH,IHW,ITYPE)
      IF(NFRAM.EQ.-1) GOTO 310

C...
C...
      IHW(62) = IHR(62)
      IHW(61) = IHR(61)
      IHW(63) = -32000
      IF(IOPT(5).EQ.1) THEN
          IF(IMODE.EQ.0) THEN
              F1=FLOAT(N4)
              F2=FLOAT(N5)
              IHW(75)=N4
              IHW(76)=N5
          ELSE
              F1=FLOAT(IHR(75))
              F2=FLOAT(IHR(76))
              IHW(75)=IHR(75)
              IHW(76)=IHR(76)
          END IF
      ELSE
          IHW(75)=0
          IHW(76)=0
      END IF

```

```

      END IF
      IF (IMODE.EQ.1) SCALW=SCALR
C...
      NCEP=NPTS
      AMULT=ACEP
C...
C...
C...
      MAP FORWARD TRANSFORM
      IF (IOPT(5).EQ.1.AND.IMODE.EQ.0) THEN
          CALL CMAP(NCEP,Y,X,FS,F1,F2,IMODE)
C      STOP
          CALL MR2R(X,Y,NCEP,I1)
      END IF
          CALL CCPC(NCEP,Y,X,IMODE)
C...
C...
C...
      MAP INVERSE???
      IF (IOPT(5).EQ.1.AND.(IMODE.EQ.2.OR.IMODE.EQ.3)) THEN
          CALL CMAP(NCEP,X,Y,FS,F1,F2,IMODE)
          CALL MR2R(Y,X,NCEP,I1)
      END IF
C...
C...
C...
C...
C...
      X IS THE OUTPUT VARIABLE
      SCALE OUTPUT VARIABLE
      IF (IMODE.EQ.1) GOTO 275
      CALL FPPIC(X,I1,NPTS,JLOC)
      SCALE=ABS(32767./X(JLOC))
      SCALW=1./SCALE
C...
C...
C...
C...
C...
      NOW THAT OUTPUT SCALE IS KNOWN..WRITE HEADER
      SCALE IS EQUIVALENCED TO HEADER LOCATION 70
      WHEN THE ZERO OPTION IS USED, NO SCALING IS DONE
      ALSO SET ALPHA=AMULT TO STORE IN HEADER
C...
275    ALPHA=AMULT
      CALL WHEAD(IHW,IFLSA,LENSA)
C...
C...
C...
      CONVERT TO INTEGER..SCALE FOR MAXIMUM ACCURACY
      CALL MR2I(X,IX,NPTS,SCALE)
      IFRAM=1
      CALL WRITD(NSC,IFRAM,ISDB,NPTS,IS,LIS,IX,IIBW,
1        IFLSA,LENSA,LFILSA)
280    CONTINUE
C...
      DUMP BUFFER
      IIBW=-IIBW
      CALL WRITD(NSC,IFRAM,ISDB,NPTS,IS,LIS,IX,IIBW,
1        IFLSA,LENSA,LFILSA)
100    CONTINUE
999    CONTINUE
310    CALL WCOMM
      CALL EXILS
      END

```

\*\*\*\*\*  
 List of Fortran Modules needed to compile XCP.FOR  
 \*\*\*\*\*

AMODSQ

FFA

FFS

FFT

FFT842

MR1DF

ORD1

ORD2

PHADVT

PHAUNW

PHCHCK

PPVPHA

R2TR

R2TX

R4SYN

R4TR

R4TX

R8SYN

R8TR

R8TX

RP

SPCVAL

----- contained in CCEPS.FOR and CCXTRA.FOR

SHIFTF

\*\*\*\*\*  
 CMAP.FOR  
 \*\*\*\*\*

SUBROUTINE CMAP(NSAMP,Y,B,FS,F1,F2,MODE)

C...

C...

C...

C...

C...

C...

C...

C...

C...

C...

C...

C...

C...

C...

C...

C...

C...

C...

C...

C...

ARGUMENTS:

NSAMP - NUMBER OF SAMPLES

Y - INPUT VARIABLE TIME SERIES

B - OUTPUT VARIABLE TIME SERIES

FS - SAMPLING FREQUENCY

F1 - LOWER FREQUENCY BOUND

F2 - UPPER FREQUENCY BOUND

MODE - IMODE FROM CEPSTRUM

\*\*\*\*\*

C...

C...

C...

C...

C...

VARIABLE LIST

DELF FREQUENCY INCREMENT

DELT TIME INCREMENT

DINT INTERPOLATING FRACTION

FINT REAL LOCATION OF LOWER FREQUENCY BOUND

```

C...  FMAX      MAXIMUM FREQUENCY
C...  ILOC      INTEGER LOCATION OF INTERPOLATED FREQUENCY
C...  INTF1     INTEGER LOCATION OF LOWER FREQUENCY BOUND
C...  INTF2     INTEGER LOCATION OF UPPER FREQUENCY BOUND
C...  IOFSET    ADDER TO LOCATE FREQUENCY RANGE
C...  IPOS      LOCATION OF REAL PART OF LOWER INTERPOLATION PAIR
C...  NINT      NUMBER OF FREQUENCIES TO MAP(F2-F1), DOES NOT
C          INCLUDE F1
C...  R1        REAL PART OF LOWER INTERPOLATION PAIR
C...  R2        REAL PART OF UPPER INTERPOLATION PAIR
C...  RATIO     RATIO OF MAPPED FREQUENCIES TO TOTAL FREQUENCIES
C...  VLOC      LOCATION OF INTERPOLATED FREQUENCY
C...  X1        IMAGINARY PART OF LOWER INTERPOLATION PAIR
C...  X2        IMAGINARY PART OF UPPER INTERPOLATION PAIR
C*****
      DIMENSION Y(1),B(NSAMP+2)
      DELT=1./FS
      DELF=FS/NSAMP
      NFREQ=NSAMP/2.
      FMAX=FS/2.

C...
C...  WHERE IN THE FREQUENCY ARRAY IS THE LOWER FREQUENCY BOUND
C...
      FINT=F1/DELF
      INTF1=IFIX(FINT)
      IF(FLOAT(INTF1).LT.FINT) INTF1=INTF1+1
      INTF2=IFIX(F2/DELF)
      NINT=INTF2-INTF1
      IOFSET=2*INTF1+1

C...
C...  CONVERT TO FREQUENCY DOMAIN
C...
C          SUBROUTINE FFA REPLACES THE REAL VECTOR B(K),
C          (K=1,2,...,N),
C...  WITH ITS FINITE DISCRETE FOURIER TRANSFORM.  THE DC TERM IS
C...  RETURNED IN LOCATION B(1) WITH B(2) SET TO 0.  THEREAFTER,
C...  THE
C...  JTH HARMONIC IS RETURNED AS A COMPLEX NUMBER STORED AS
C...  B(2*J+1) + I B(2*J+2).  NOTE THAT THE N/2 HARMONIC IS
C...  RETURNED
C...  IN B(N+1) WITH B(N+2) SET TO 0.  HENCE, B MUST BE
C...  DIMENSIONED
C...  TO SIZE N+2.
C...  SUBROUTINE IS CALLED AS FFA (B,N) WHERE N=2**M AND B IS AN
C...  N TERM REAL ARRAY.
C...
C...  CALL FFA(Y,NSAMP)

C...
C...  Y NOW CONTAINS FREQUENCY VALUES
C...  MAP SELECTED RANGE INTO ARRAY B
C...
      IF(MODE.EQ.0) THEN
          B(1)=Y(IOFSET)
          B(2)=Y(IOFSET+1)

```

```

C...      RATIO=FLOAT(NINT)/FLOAT(NFREQ)
C      TYPE *,DELT,DELF,NFREQ,FMAX,RATIO
C      TYPE *,INTF1,INTF2,NINT,IOFSET
          DO 100 KJ=1,NFREQ
              VLOC=FLOAT(KJ)*RATIO
              ILOC=IFIX(VLOC)
              DINT=VLOC-ILOC
              IPOS=2*ILOC+IOFSET
              R1=Y(IPOS)
              R2=Y(IPOS+2)
              X1=Y(IPOS+1)
              X2=Y(IPOS+3)
              B(2*KJ+1)=R1+DINT*(R2-R1)
              B(2*KJ+2)=X1+DINT*(X2-X1)

100      CONTINUE
9999     FORMAT((I8,4(4F7.0,2X)))/
C...
C...     SHIFT PHASE
C...
          CALL SHIFTF(B,NSAMP)
C      WRITE(3,9997)
9997     FORMAT(8X,4(5X,'INPUT'9X,'SHIFTED',4X))
C      WRITE(3,9998)
9998     FORMAT(' FREQ ',4(2(' REAL IMAG ')))
C      WRITE(3,9999)(J,(Y(2*J+1),Y(2*J+2),B(2*J+1),B(2*J+2),
C      1          Y(2*J+3),Y(2*J+4),B(2*J+3),B(2*J+4),
C      2          Y(2*J+5),Y(2*J+6),B(2*J+5),B(2*J+6),
C      3          Y(2*J+7),Y(2*J+8),B(2*J+7),B(2*J+8)),
C      4          J=0,NFREQ-3,4)
      ELSE
          RATIO=FLOAT(NFREQ)/FLOAT(NINT)
          B(1)=0.
          B(2)=0.
          DO 200 KJ=1,NFREQ
              IF(KJ.LT.INTF1.OR.KJ.GT.INTF2) THEN
                  B(2*KJ+1)=0.
                  B(2*KJ+2)=0.
              ELSE
                  INDX=KJ-INTF1
                  VLOC=RATIO*FLOAT(INDX)
                  ILOC=IFIX(VLOC)
                  DINT=VLOC-ILOC
                  IPOS=2*ILOC+1
                  R1=Y(IPOS)
                  R2=Y(IPOS+2)
                  X1=Y(IPOS+1)
                  X2=Y(IPOS+3)
                  B(2*KJ+1)=R1+DINT*(R2-R1)
                  B(2*KJ+2)=X1+DINT*(X2-X1)
              END IF
          END IF
200      CONTINUE
C...     END IF

```

```

C*****
C...   SUBROUTINE FFS SYNTHESIZES THE REAL VECTOR B(K), WHERE
C...   K=1,2,...,N. THE INITIAL FOURIER COEFFICIENTS ARE PLACED IN
C...   THE B ARRAY OF SIZE N+2. THE DC TERM IS IN B(1) WITH
C...   B(2) EQUAL TO 0.
C...   THE JTH HARMONIC IS STORED AS B(2*J+1) + I B(2*J+2).
C...   THE N/2 HARMONIC IS IN B(N+1) WITH B(N+2) EQUAL TO 0.
C...   THE SUBROUTINE IS CALLED AS FFS(B,N) WHERE N=2**M AND
C...   B IS THE N TERM REAL ARRAY DISCUSSED ABOVE.
C...
      CALL FFS(B,NSAMP)
C..
      RETURN
      END

```

```

*****
CCEPS.FOR
*****

```

```

C
C-----
C-----
C SUBROUTINE: CCEPS
C SUBROUTINE TO COMPUTE THE COMPLEX CEPSTRUM OF A SEQUENCE X(N)
C-----
C-----
C
      SUBROUTINE CCEPS(NX,X,ISNX,ISFX,ISSUC,CX,AUX)
C
      DIMENSION X(1),CX(1),AUX(1)
      COMMON PI,TWOPI,THLINC,THLCON,NFFT,NPTS,N,L,H,H1,DVTMN2
      LOGICAL ISSUC
      NPTS=NFFT/2
      N=12
      L=2**N
      H=FLOAT(L)*FLOAT(NFFT)
      H1=PI/H
      ISSUC=.TRUE.
      ISNX=1
C
      DO 10 I=1,NX
        CX(I)=X(I)
        AUX(I)=FLOAT(I-1)*X(I)
10  CONTINUE
      INITL=NX+1
      IEND=NFFT+2
      DO 20 I=INITL,IEND
        CX(I)=0.0
        AUX(I)=0.0
20  CONTINUE

```

```

C      CALL PFA(CX,NFFT)
      CALL PFA(AUX,NFFT)
C
      IF(CX(1).LT.0.0) ISNX=-1
C
      IO=-1
      DVTMN2=0.0
      IEND=NPTS+1
      DO 30 I=1,IEND
        IO=IO+2
        IE=IO+1
        AMAGSQ=AMODSQ(CX(IO),CX(IE))
        PDVT=PHADVT(CX(IO),CX(IE),AUX(IO),AUX(IE),AMAGSQ)
        AUX(IO)=AMAGSQ
        AUX(IE)=PDVT
        DVTMN2=DVTMN2+PDVT
30    CONTINUE
      DVTMN2=(2.*DVTMN2-AUX(2)-PDVT)/FLOAT(NPTS)
C
      PPDVT=AUX(2)
      PPHASE=0.0
      PPV=PPVPHA(CX(1),CX(2),ISNX)
      CX(1)=.5*ALOG(AUX(1))
      CX(2)=0.0
      IO=1
      DO 50 I=2,IEND
        IO=IO+2
        IE=IO+1
        PDVT=AUX(IE)
        PPV=PPVPHA(CX(IO),CX(IE),ISNX)
        PHASE=PHAUNW(X,NX,ISNX,I,PPHASE,PPDVT,PPV,PDVT,ISSUC)
C
        IF(ISSUC)GO TO 40
        ISSUC=.FALSE.
        RETURN
40    PPDVT=PDVT
        PPHASE=PHASE
        CX(IO)=.5*ALOG(AUX(IO))
        CX(IE)=PHASE
50    CONTINUE
C
      ISFX=(ABS(PHASE/PI)+.1)
      IF(PHASE.LT.0.0) ISFX=-ISFX
      H=PHASE/FLOAT(NPTS)
      IE=0
      DO 60 I=1,IEND
        IE=IE+2
        CX(IE)=CX(IE)-H*FLOAT(I-1)
60    CONTINUE

```

```

C      CALL FFS(CX,NFFT)
C      RETURN
C      END

C-----
C SUBROUTINE: SPCVAL
C SUBROUTINE TO COMPUTRE A SPECTRAL VALUE AT A FREQUENCY
C FREQ(RADIANS) FOR SEQUENCE X(N) AND N*X(N)
C-----
C
C      SUBROUTINE SPCVAL(NX,X,FREQ,XR,XI,YR,YI)
C      DIMENSION X(1)
C      DOUBLE PRECISION U0,U1,U2,W0,W1,W2,A,B,C,D,A1,A2,SA0,CA0
C
C      CA0=DBLE(COS(FREQ))
C      SA0=DBLE(SIN(FREQ))
C      A1=2.D+0*CA0
C      U1=0.D+0
C      U2=U1
C      W1=U1
C      W2=U1
C
C      DO 10 J=1,NX
C          XJ=DBLE(X(J))
C          U0=XJ+A1*U1-U2
C          W0=DBLE(FLOAT(J-1))*XJ+A1*W1-W2
C          U2=U1
C          U1=U0
C          W2=W1
C          W1=W0
C      10 CONTINUE
C
C      A=U1-U2*CA0
C      B=U2*SA0
C      C=W1-W2*CA0
C      D=W2*SA0
C      A2=DBLE(FREQ*FLOAT(NX-1))
C      U1=DCOS(A2)
C      U2=-DSIN(A2)
C      XR=SNGL(U1*A-U2*B)
C      XI=SNGL(U2*A+U1*B)
C      YR=SNGL(U1*C-U2*D)
C      YI=SNGL(U2*C+U1*D)
C      RETURN
C      END

C-----
C FUNCTION: PHAUNW
C PHASE UNWRAPPING BASED ON TRIBOLET'S ADAPTIVE INTEGRATION SCHEME.

```



C THE UNWRAPPED PHASE ESTIMATE IS RETURNED IN PHAUNW.

```
C-----
C
C      FUNCTION PHAUNW(X,NX,ISNX,I,PPHASE,PPDVT,PPV,PDVT,ISCONS)
C
C      DIMENSION SDVT(17),SPPV(17),X(1)
C      INTEGER SINDEXT(17),PINDEXT,SP
C      LOGICAL ISCONS,FIRST
C      COMMON PI,TWOPI,THLINC,THLCON,NFFT,NPTS,N,L,H,H1,DVTMN2
C
C      FIRST=.TRUE.
C      PINDEXT=1
C      SP=1
C      SPPV(SP)=PPV
C      SDVT(SP)=PDVT
C      SINDEXT(SP)=L+1
C
C      GO TO 40
C
C 10  PINDEXT=SINDEXT(SP)
C      PPHASE=PPHASE
C      PPDVT=SDVT(SP)
C      SP=SP+1
C      GO TO 40
C
C 20  IF((SINDEXT(SP)-PINDEXT).GT.1)GO TO 30
C      ISCONS=.FALSE.
C      PHAUNW=0.
C      RETURN
C
C 30  K=(SINDEXT(SP)+PINDEXT)/2
C
C      FREQ=TWOPI*(FLOAT(I-2)*FLOAT(L)+FLOAT(K-1))/H
C      CALL SPCVAL(NX,X,FREQ,XR,XI,YR,YI)
C
C      SP=SP+1
C      SINDEXT(SP)=K
C      SPPV(SP)=PPVPHA(XR,XI,ISNX)
C      XMAG=AMODSQ(XR,XI)
C      SDVT(SP)=PHADVT(XR,XI,YR,YI,XMAG)
C
C 40  DELTA=H1*FLOAT(SINDEXT(SP)-PINDEXT)
C      PHAINC=DELTA*(PPDVT+SDVT(SP))
C
C      IF(ABS(PHAINC-DELTA*DVTMN2).GT.THLINC)GO TO 20
C
C      PHASE=PPHASE+PHAINC
C      CALL PHCHCK(PHASE,SPPV(SP),ISCONS)
C      IF(.NOT.ISCONS)GO TO 20
C
C      IF(ABS(PHASE-PPHASE).GT.PI)GO TO 20
```

```

C      IF(SP.NE.1)GO TO 10
C      PHAUNW=PHASE
C      RETURN
C      END
C
C-----
C
C FUNCTION: PPVPHA
C COMPUTE THE PRINCIPLE VALUE OF THE PHASE OF A SPECTRAL VALUE
C-----
C
C      FUNCTION PPVPHA(XR,XI,ISNX)
C
C      IF (ISNX.EQ.1) PPVPHA=(ATAN2((XI),(XR)))
C      IF (ISNX.EQ.(-1)) PPVPHA=(ATAN2(-(XI),-(XR)))
C      RETURN
C      END
C
C-----
C
C FUNCTION: PHADVT
C COMPUTE THE PHASE DERIVATIVE OF A SPECTRAL VALUE OF A SEQUENCE
C      X(N)
C-----
C
C      FUNCTION PHADVT(XR,XI,YR,YI,XMAG)
C
C      PHADVT=-SNGL((DBLE(XR)*DBLE(YR)+DBLE(XI)*DBLE(YI))/DBLE(XMAG))
C      RETURN
C      END
C
C-----
C
C FUNCTION: AMODSQ
C COMPUTE THE SQUARE OF THE MODULUS OF A COMPLEX NUMBER
C-----
C
C      FUNCTION AMODSQ(ZR,ZI)
C
C      AMODSQ=SNGL(DBLE(ZR)*DBLE(ZR)+DBLE(ZI)*DBLE(ZI))
C      RETURN
C      END
C
C-----
C
C SUBROUTINE: PHCHCK
C SUBROUTINE TO CHECK CONSISTENCY OF A PHASE ESTIMATE
C-----

```

```

C
SUBROUTINE PHCHCK(PH,PV,ISCONS)
C
COMMON PI,TWOPI,THLINC,THLCON,NFFT,NPTS,N,L,H,H1,DVTMN2
LOGICAL ISCONS
C
A0=(PH-PV)/TWOPI
A1=FLOAT(IFIX(A0))*TWOPI+PV
A2=A1+SIGN(TWOPI,A0)
A3=ABS(A1-PH)
A4=ABS(A2-PH)
C
ISCONS=.FALSE.
IF(A3.GT.THLCN.AND.A4.GT.THLCN) RETURN
ISCONS=.TRUE.
C
PH=A1
IF(A3.GT.A4) PH=A2
RETURN
END
C
-----
C SUBROUTINE: FFA
C FAST FOURIER ANALYSIS SUBROUTINE
C -----
C
SUBROUTINE FFA(B, NFFT)
C
C THIS SUBROUTINE REPLACES THE REAL VECTOR B(K), (K=1,2,...,N),
C WITH ITS FINITE DISCRETE FOURIER TRANSFORM. THE DC TERM IS
C RETURNED IN LOCATION B(1) WITH B(2) SET TO 0. THEREAFTER, THE
C JTH HARMONIC IS RETURNED AS A COMPLEX NUMBER STORED AS
C B(2*J+1) + I B(2*J+2). NOTE THAT THE N/2 HARMONIC IS RETURNED
C IN B(N+1) WITH B(N+2) SET TO 0. HENCE, B MUST BE DIMENSIONED
C TO SIZE N+2.
C SUBROUTINE IS CALLED AS FFA (B,N) WHERE N=2**M AND B IS AN
C N TERM REAL ARRAY. A REAL-VALUED, RADIX 8 ALGORITHM IS USED
C WITH IN-PLACE REORDERING AND THE TRIG FUNCTIONS ARE COMPUTED AS
C NEEDED.
C
DIMENSION B(2)
COMMON /CON/ PII, P7, P7TWO, C22, S22, PI2
C
C IW IS A MACHINE DEPENDENT WRITE DEVICE NUMBER
C
C IW = ILMACH(2)
C IW=6
C
C
PII = 4.*ATAN(1.)
PI8 = PII/8.
P7 = 1./SQRT(2.)

```

```

P7TWO = 2.*P7
C22 = COS(PI8)
S22 = SIN(PI8)
PI2 = 2.*PII
N = 1
DO 10 I=1,15
    M = I
    N = N*2
    IF (N.EQ.NFFT) GO TO 20
10  CONTINUE
    WRITE (IW,9999)
9999 FORMAT (30H NFFT NOT A POWER OF 2 FOR FFA)
    STOP
20  CONTINUE
    N8POW = M/3
C
C DO A RADIX 2 OR RADIX 4 ITERATION FIRST IF ONE IS REQUIRED
C
    IF (M-N8POW*3-1) 50, 40, 30
30  NN = 4
    INT = N/NN
    CALL R4TR(INT, B(1), B(INT+1), B(2*INT+1), B(3*INT+1))
    GO TO 60
40  NN = 2
    INT = N/NN
    CALL R2TR(INT, B(1), B(INT+1))
    GO TO 60
50  NN = 1
C
C PERFORM RADIX 8 ITERATIONS
C
60  IF (N8POW) 90, 90, 70
70  DO 80 IT=1,N8POW
    NN = NN*8
    INT = N/NN
    CALL R8TR(INT, NN, B(1), B(INT+1), B(2*INT+1), B(3*INT+1),
    *      B(4*INT+1), B(5*INT+1), B(6*INT+1), B(7*INT+1), B(1),
    *      B(INT+1), B(2*INT+1), B(3*INT+1), B(4*INT+1),
B(5*INT+1),
    *      B(6*INT+1), B(7*INT+1))
80  CONTINUE
C
C PERFORM IN-PLACE REORDERING
C
90  CALL ORD1(M, B)
    CALL ORD2(M, B)
    T = B(2)
    B(2) = 0.
    B(NFFT+1) = T
    B(NFFT+2) = 0.
    DO 100 I=4,NFFT,2
        B(I) = -B(I)

```

```

100  CONTINUE
      RETURN
      END

C
C-----
C SUBROUTINE:  FFS
C FAST FOURIER SYNTHESIS SUBROUTINE
C RADIX 8-4-2
C-----
C
C      SUBROUTINE FFS(B, NFFT)
C
C THIS SUBROUTINE SYNTHESIZES THE REAL VECTOR B(K), WHERE
C K=1,2,...,N. THE INITIAL FOURIER COEFFICIENTS ARE PLACED IN
C THE B ARRAY OF SIZE N+2. THE DC TERM IS IN B(1) WITH
C B(2) EQUAL TO 0.
C THE JTH HARMONIC IS STORED AS B(2*J+1) + I B(2*J+2).
C THE N/2 HARMONIC IS IN B(N+1) WITH B(N+2) EQUAL TO 0.
C THE SUBROUTINE IS CALLED AS FFS(B,N) WHERE N=2**M AND
C B IS THE N TERM REAL ARRAY DISCUSSED ABOVE.
C
C      DIMENSION B(2)
C      COMMON /CON1/ PII, P7, P7TWO, C22, S22, PI2
C
C IW IS A MACHINE DEPENDENT WRITE DEVICE NUMBER
C
C      IW = ILMACH(2)
C      IW=6
C
C      PII = 4.*ATAN(1.)
C      PI8 = PII/8.
C      P7 = 1./SQRT(2.)
C      P7TWO = 2.*P7
C      C22 = COS(PI8)
C      S22 = SIN(PI8)
C      PI2 = 2.*PII
C      N = 1
C      DO 10 I=1,15
C          M = I
C          N = N*2
C          IF (N.EQ.NFFT) GO TO 20
10  CONTINUE
      WRITE (IW,9999)
9999 FORMAT (30H NFFT NOT A POWER OF 2 FOR FFS)
      STOP
20  CONTINUE
      B(2) = B(NFFT+1)
      DO 30 I=1,NFFT
          B(I) = B(I)/FLOAT(NFFT)
30  CONTINUE
      DO 40 I=4,NFFT,2
          B(I) = -B(I)

```

```

40  CONTINUE
    N8POW = M/3
C
C  REORDER THE INPUT FOURIER COEFFICIENTS
C
    CALL ORD2(M, B)
    CALL ORD1(M, B)
C
    IF (N8POW.EQ.0) GO TO 60
C
C  PERFORM THE RADIX 8 ITERATIONS
C
    NN = N
    DO 50 IT=1,N8POW
        INT = N/NN
        CALL R8SYN(INT, NN, B, B(INT+1), B(2*INT+1), B(3*INT+1),
*           B(4*INT+1), B(5*INT+1), B(6*INT+1), B(7*INT+1), B(1),
*           B(INT+1), B(2*INT+1), B(3*INT+1), B(4*INT+1),
B(5*INT+1),
*           B(6*INT+1), B(7*INT+1))
        NN = NN/8
    50  CONTINUE
C
C  DO A RADIX 2 OR RADIX 4 ITERATION IF ONE IS REQUIRED
C
    60  IF (M-N8POW*3-1) 90, 80, 70
    70  INT = N/4
        CALL R4SYN(INT, B(1), B(INT+1), B(2*INT+1), B(3*INT+1))
        GO TO 90
    80  INT = N/2
        CALL R2TR(INT, B(1), B(INT+1))
    90  RETURN
        END
C
C-----
C
C  SUBROUTINE:  R2TR
C  RADIX 2 ITERATION SUBROUTINE
C-----
C
C
C  SUBROUTINE R2TR(INT, B0, B1)
C  DIMENSION B0(2), B1(2)
C  DO 10 K=1,INT
C      T = B0(K) + B1(K)
C      B1(K) = B0(K) - B1(K)
C      B0(K) = T
    10  CONTINUE
        RETURN
        END
C
C-----
C

```

```

C SUBROUTINE: R4TR
C RADIX 4 ITERATION SUBROUTINE

```

```

C-----

```

```

C
SUBROUTINE R4TR(INT, B0, B1, B2, B3)
DIMENSION B0(2), B1(2), B2(2), B3(2)
DO 10 K=1,INT
  R0 = B0(K) + B2(K)
  R1 = B1(K) + B3(K)
  B2(K) = B0(K) - B2(K)
  B3(K) = B1(K) - B3(K)
  B0(K) = R0 + R1
  B1(K) = R0 - R1
10 CONTINUE
RETURN
END

```

```

C-----

```

```

C SUBROUTINE: R8TR
C RADIX 8 ITERATION SUBROUTINE

```

```

C-----

```

```

C
SUBROUTINE R8TR(INT, NN, BR0, BR1, BR2, BR3, BR4, BR5, BR6,
BR7,
*   BI0, BI1, BI2, BI3, BI4, BI5, BI6, BI7)
DIMENSION L(15), BR0(2), BR1(2), BR2(2), BR3(2), BR4(2),
BR5(2),
*   BR6(2), BR7(2), BI0(2), BI1(2), BI2(2), BI3(2), BI4(2),
*   BI5(2), BI6(2), BI7(2)
COMMON /CON/ PII, P7, P7TWO, C22, S22, PI2
EQUIVALENCE (L15,L(1)), (L14,L(2)), (L13,L(3)), (L12,L(4)),
*   (L11,L(5)), (L10,L(6)), (L9,L(7)), (L8,L(8)), (L7,L(9)),
*   (L6,L(10)), (L5,L(11)), (L4,L(12)), (L3,L(13)),
(L2,L(14)),
*   (L1,L(15))

```

```

C
C SET UP COUNTERS SUCH THAT JTHET STEPS THROUGH THE ARGUMENTS
C OF W, JR STEPS THROUGH STARTING LOCATIONS FOR THE REAL PART OF
THE
C INTERMEDIATE RESULTS AND JI STEPS THROUGH STARTING LOCATIONS
C OF THE IMAGINARY PART OF THE INTERMEDIATE RESULTS.

```

```

C
L(1) = NN/8
DO 40 K=2,15
  IF (L(K-1)-2) 10, 20, 30
10  L(K-1) = 2
20  L(K) = 2
  GO TO 40
30  L(K) = L(K-1)/2
40  CONTINUE
PIOVN = PII/FLOAT(NN)

```

```

JI = 3
JL = 2
JR = 2
DO 120 J1=2,L1,2
DO 120 J2=J1,L2,L1
DO 120 J3=J2,L3,L2
DO 120 J4=J3,L4,L3
DO 120 J5=J4,L5,L4
DO 120 J6=J5,L6,L5
DO 120 J7=J6,L7,L6
DO 120 J8=J7,L8,L7
DO 120 J9=J8,L9,L8
DO 120 J10=J9,L10,L9
DO 120 J11=J10,L11,L10
DO 120 J12=J11,L12,L11
DO 120 J13=J12,L13,L12
DO 120 J14=J13,L14,L13
DO 120 JTHET=J14,L15,L14
TH2 = JTHET - 2
IF (TH2) 50, 50, 90
50 DO 60 K=1,INT
    T0 = BR0(K) + BR4(K)
    T1 = BR1(K) + BR5(K)
    T2 = BR2(K) + BR6(K)
    T3 = BR3(K) + BR7(K)
    T4 = BR0(K) - BR4(K)
    T5 = BR1(K) - BR5(K)
    T6 = BR2(K) - BR6(K)
    T7 = BR3(K) - BR7(K)
    BR2(K) = T0 - T2
    BR3(K) = T1 - T3
    T0 = T0 + T2
    T1 = T1 + T3
    BR0(K) = T0 + T1
    BR1(K) = T0 - T1
    PR = P7*(T5-T7)
    PI = P7*(T5+T7)
    BR4(K) = T4 + PR
    BR7(K) = T6 + PI
    BR6(K) = T4 - PR
    BR5(K) = PI - T6
60 CONTINUE
IF (NN-8) 120, 120, 70
70 K0 = INT*8 + 1
KL = K0 + INT - 1
DO 80 K=K0,KL
    PR = P7*(BI2(K)-BI6(K))
    PI = P7*(BI2(K)+BI6(K))
    TR0 = BI0(K) + PR
    TI0 = BI4(K) + PI
    TR2 = BI0(K) - PR
    TI2 = BI4(K) - PI
    PR = P7*(BI3(K)-BI7(K))
    PI = P7*(BI3(K)+BI7(K))

```



```

      TR1 = BI1(K) + PR
      TI1 = BI5(K) + PI
      TR3 = BI1(K) - PR
      TI3 = BI5(K) - PI
      PR = TR1*C22 - TI1*S22
      PI = TI1*C22 + TR1*S22
      BI0(K) = TR0 + PR
      BI6(K) = TR0 - PR
      BI7(K) = TI0 + PI
      BI1(K) = PI - TI0
      PR = -TR3*S22 - TI3*C22
      PI = TR3*C22 - TI3*S22
      BI2(K) = TR2 + PR
      BI4(K) = TR2 - PR
      BI5(K) = TI2 + PI
      BI3(K) = PI - TI2
80    CONTINUE
      GO TO 120
90    ARG = TH2*PIOVN
      C1 = COS(ARG)
      S1 = SIN(ARG)
      C2 = C1**2 - S1**2
      S2 = C1*S1 + C1*S1
      C3 = C1*C2 - S1*S2
      S3 = C2*S1 + S2*C1
      C4 = C2**2 - S2**2
      S4 = C2*S2 + C2*S2
      C5 = C2*C3 - S2*S3
      S5 = C3*S2 + S3*C2
      C6 = C3**2 - S3**2
      S6 = C3*S3 + C3*S3
      C7 = C3*C4 - S3*S4
      S7 = C4*S3 + S4*C3
      INT8 = INT*8
      J0 = JR*INT8 + 1
      K0 = JI*INT8 + 1
      JLAST = J0 + INT - 1
      DO 100 J=J0,JLAST
        K = K0 + J - J0
        TR1 = BR1(J)*C1 - BI1(K)*S1
        TI1 = BR1(J)*S1 + BI1(K)*C1
        TR2 = BR2(J)*C2 - BI2(K)*S2
        TI2 = BR2(J)*S2 + BI2(K)*C2
        TR3 = BR3(J)*C3 - BI3(K)*S3
        TI3 = BR3(J)*S3 + BI3(K)*C3
        TR4 = BR4(J)*C4 - BI4(K)*S4
        TI4 = BR4(J)*S4 + BI4(K)*C4
        TR5 = BR5(J)*C5 - BI5(K)*S5
        TI5 = BR5(J)*S5 + BI5(K)*C5
        TR6 = BR6(J)*C6 - BI6(K)*S6
        TI6 = BR6(J)*S6 + BI6(K)*C6
        TR7 = BR7(J)*C7 - BI7(K)*S7
        TI7 = BR7(J)*S7 + BI7(K)*C7

```

C

$T0 = BR0(J) + TR4$   
 $T1 = BI0(K) + TI4$   
 $TR4 = BR0(J) - TR4$   
 $TI4 = BI0(K) - TI4$   
 $T2 = TR1 + TR5$   
 $T3 = TI1 + TI5$   
 $TR5 = TR1 - TR5$   
 $TI5 = TI1 - TI5$   
 $T4 = TR2 + TR6$   
 $T5 = TI2 + TI6$   
 $TR6 = TR2 - TR6$   
 $TI6 = TI2 - TI6$   
 $T6 = TR3 + TR7$   
 $T7 = TI3 + TI7$   
 $TR7 = TR3 - TR7$   
 $TI7 = TI3 - TI7$

C

$TR0 = T0 + T4$   
 $TI0 = T1 + T5$   
 $TR2 = T0 - T4$   
 $TI2 = T1 - T5$   
 $TR1 = T2 + T6$   
 $TI1 = T3 + T7$   
 $TR3 = T2 - T6$   
 $TI3 = T3 - T7$   
 $T0 = TR4 - TI6$   
 $T1 = TI4 + TR6$   
 $T4 = TR4 + TI6$   
 $T5 = TI4 - TR6$   
 $T2 = TR5 - TI7$   
 $T3 = TI5 + TR7$   
 $T6 = TR5 + TI7$   
 $T7 = TI5 - TR7$   
 $BR0(J) = TR0 + TR1$   
 $BI7(K) = TI0 + TI1$   
 $BI6(K) = TR0 - TR1$   
 $BR1(J) = TI1 - TI0$   
 $BR2(J) = TR2 - TI3$   
 $BI5(K) = TI2 + TR3$   
 $BI4(K) = TR2 + TI3$   
 $BR3(J) = TR3 - TI2$   
 $PR = P7*(T2-T3)$   
 $PI = P7*(T2+T3)$   
 $BR4(J) = T0 + PR$   
 $BI3(K) = T1 + PI$   
 $BI2(K) = T0 - PR$   
 $BR5(J) = PI - T1$   
 $PR = -P7*(T6+T7)$   
 $PI = P7*(T6-T7)$   
 $BR6(J) = T4 + PR$   
 $BI1(K) = T5 + PI$   
 $BI0(K) = T4 - PR$   
 $BR7(J) = PI - T5$

```

100    CONTINUE
      JR = JR + 2
      JI = JI - 2
      IF (JI-JL) 110, 110, 120
110    JI = 2*JR - 1
      JL = JR
120    CONTINUE
      RETURN
      END

```

C

C-----

C SUBROUTINE: R4SYN  
C RADIX 4 SYNTHESIS

C-----

C

```

SUBROUTINE R4SYN(INT, B0, B1, B2, B3)
DIMENSION B0(2), B1(2), B2(2), B3(2)
DO 10 K=1,INT
  T0 = B0(K) + B1(K)
  T1 = B0(K) - B1(K)
  T2 = B2(K) + B2(K)
  T3 = B3(K) + B3(K)
  B0(K) = T0 + T2
  B2(K) = T0 - T2
  B1(K) = T1 + T3
  B3(K) = T1 - T3

```

```

10    CONTINUE
      RETURN
      END

```

C

C-----

C SUBROUTINE: R8SYN  
C RADIX 8 SYNTHESIS SUBROUTINE

C-----

C

```

SUBROUTINE R8SYN(INT, NN, BR0, BR1, BR2, BR3, BR4, BR5, BR6,
BR7,
*    BI0, BI1, BI2, BI3, BI4, BI5, BI6, BI7)
DIMENSION L(15), BR0(2), BR1(2), BR2(2), BR3(2), BR4(2),
BR5(2),
*    BR6(2), BR7(2), BI0(2), BI1(2), BI2(2), BI3(2), BI4(2),
*    BI5(2), BI6(2), BI7(2)
COMMON /CON1/ PII, P7, P7TWO, C22, S22, PI2
EQUIVALENCE (L15,L(1)), (L14,L(2)), (L13,L(3)), (L12,L(4)),
*    (L11,L(5)), (L10,L(6)), (L9,L(7)), (L8,L(8)), (L7,L(9)),
*    (L6,L(10)), (L5,L(11)), (L4,L(12)), (L3,L(13)),
(L2,L(14)),
*    (L1,L(15))
L(1) = NN/8
DO 40 K=2,15

```

```

      IF (L(K-1)-2) 10, 20, 30
10    L(K-1) = 2
20    L(K) = 2
      GO TO 40
30    L(K) = L(K-1)/2
40    CONTINUE
      PIOVN = PII/FLOAT(NN)
      JI = 3
      JL = 2
      JR = 2

```

C

```

      DO 120 J1=2,L1,2
      DO 120 J2=J1,L2,L1
      DO 120 J3=J2,L3,L2
      DO 120 J4=J3,L4,L3
      DO 120 J5=J4,L5,L4
      DO 120 J6=J5,L6,L5
      DO 120 J7=J6,L7,L6
      DO 120 J8=J7,L8,L7
      DO 120 J9=J8,L9,L8
      DO 120 J10=J9,L10,L9
      DO 120 J11=J10,L11,L10
      DO 120 J12=J11,L12,L11
      DO 120 J13=J12,L13,L12
      DO 120 J14=J13,L14,L13
      DO 120 JTHET=J14,L15,L14
      TH2 = JTHET - 2
      IF (TH2) 50, 50, 90
50    DO 60 K=1,INT
      T0 = BR0(K) + BR1(K)
      T1 = BR0(K) - BR1(K)
      T2 = BR2(K) + BR2(K)
      T3 = BR3(K) + BR3(K)
      T4 = BR4(K) + BR6(K)
      T6 = BR7(K) - BR5(K)
      T5 = BR4(K) - BR6(K)
      T7 = BR7(K) + BR5(K)
      PR = P7*(T7+T5)
      PI = P7*(T7-T5)
      TT0 = T0 + T2
      TT1 = T1 + T3
      T2 = T0 - T2
      T3 = T1 - T3
      T4 = T4 + T4
      T5 = PR + PR
      T6 = T6 + T6
      T7 = PI + PI
      BR0(K) = TT0 + T4
      BR1(K) = TT1 + T5
      BR2(K) = T2 + T6
      BR3(K) = T3 + T7
      BR4(K) = TT0 - T4
      BR5(K) = TT1 - T5

```

```

        BR6(K) = T2 - T6
        BR7(K) = T3 - T7
60      CONTINUE
        IF (NN-8) 120, 120, 70
70      K0 = INT*8 + 1
        KL = K0 + INT - 1
        DO 80 K=K0, KL
            T1 = BI0(K) + BI6(K)
            T2 = BI7(K) - BI1(K)
            T3 = BI0(K) - BI6(K)
            T4 = BI7(K) + BI1(K)
            PR = T3*C22 + T4*S22
            PI = T4*C22 - T3*S22
            T5 = BI2(K) + BI4(K)
            T6 = BI5(K) - BI3(K)
            T7 = BI2(K) - BI4(K)
            T8 = BI5(K) + BI3(K)
            RR = T8*C22 - T7*S22
            RI = -T8*S22 - T7*C22
            BI0(K) = (T1+T5) + (T1+T5)
            BI4(K) = (T2+T6) + (T2+T6)
            BI1(K) = (PR+RR) + (PR+RR)
            BI5(K) = (PI+RI) + (PI+RI)
            T5 = T1 - T5
            T6 = T2 - T6
            BI2(K) = P7TWO*(T6+T5)
            BI6(K) = P7TWO*(T6-T5)
            RR = PR - RR
            RI = PI - RI
            BI3(K) = P7TWO*(RI+RR)
            BI7(K) = P7TWO*(RI-RR)
80      CONTINUE
        GO TO 120
90      ARG = TH2*PIOVN
        C1 = COS(ARG)
        S1 = -SIN(ARG)
        C2 = C1**2 - S1**2
        S2 = C1*S1 + C1*S1
        C3 = C1*C2 - S1*S2
        S3 = C2*S1 + S2*C1
        C4 = C2**2 - S2**2
        S4 = C2*S2 + C2*S2
        C5 = C2*C3 - S2*S3
        S5 = C3*S2 + S3*C2
        C6 = C3**2 - S3**2
        S6 = C3*S3 + C3*S3
        C7 = C3*C4 - S3*S4
        S7 = C4*S3 + S4*C3
        INT8 = INT*8
        J0 = JR*INT8 + 1
        K0 = JI*INT8 + 1
        JLAST = J0 + INT - 1
        DO 100 J=J0, JLAST

```

$K = K0 + J - J0$   
 $TR0 = BR0(J) + BI6(K)$   
 $TI0 = BI7(K) - BR1(J)$   
 $TR1 = BR0(J) - BI6(K)$   
 $TI1 = BI7(K) + BR1(J)$   
 $TR2 = BR2(J) + BI4(K)$   
 $TI2 = BI5(K) - BR3(J)$   
 $TR3 = BI5(K) + BR3(J)$   
 $TI3 = BI4(K) - BR2(J)$   
 $TR4 = BR4(J) + BI2(K)$   
 $TI4 = BI3(K) - BR5(J)$   
 $T0 = BR4(J) - BI2(K)$   
 $T1 = BI3(K) + BR5(J)$   
 $TR5 = P7*(T0+T1)$   
 $TI5 = P7*(T1-T0)$   
 $TR6 = BR6(J) + BI0(K)$   
 $TI6 = BI1(K) - BR7(J)$   
 $T0 = BR6(J) - BI0(K)$   
 $T1 = BI1(K) + BR7(J)$   
 $TR7 = -P7*(T0-T1)$   
 $TI7 = -P7*(T1+T0)$   
 $T0 = TR0 + TR2$   
 $T1 = TI0 + TI2$   
 $T2 = TR1 + TR3$   
 $T3 = TI1 + TI3$   
 $TR2 = TR0 - TR2$   
 $TI2 = TI0 - TI2$   
 $TR3 = TR1 - TR3$   
 $TI3 = TI1 - TI3$   
 $T4 = TR4 + TR6$   
 $T5 = TI4 + TI6$   
 $T6 = TR5 + TR7$   
 $T7 = TI5 + TI7$   
 $TTR6 = TI4 - TI6$   
 $TI6 = TR6 - TR4$   
 $TTR7 = TI5 - TI7$   
 $TI7 = TR7 - TR5$   
 $BR0(J) = T0 + T4$   
 $BI0(K) = T1 + T5$   
 $BR1(J) = C1*(T2+T6) - S1*(T3+T7)$   
 $BI1(K) = C1*(T3+T7) + S1*(T2+T6)$   
 $BR2(J) = C2*(TR2+TTR6) - S2*(TI2+TI6)$   
 $BI2(K) = C2*(TI2+TI6) + S2*(TR2+TTR6)$   
 $BR3(J) = C3*(TR3+TTR7) - S3*(TI3+TI7)$   
 $BI3(K) = C3*(TI3+TI7) + S3*(TR3+TTR7)$   
 $BR4(J) = C4*(T0-T4) - S4*(T1-T5)$   
 $BI4(K) = C4*(T1-T5) + S4*(T0-T4)$   
 $BR5(J) = C5*(T2-T6) - S5*(T3-T7)$   
 $BI5(K) = C5*(T3-T7) + S5*(T2-T6)$   
 $BR6(J) = C6*(TR2-TTR6) - S6*(TI2-TI6)$   
 $BI6(K) = C6*(TI2-TI6) + S6*(TR2-TTR6)$   
 $BR7(J) = C7*(TR3-TTR7) - S7*(TI3-TI7)$   
 $BI7(K) = C7*(TI3-TI7) + S7*(TR3-TTR7)$

```

100  CONTINUE
      JR = JR + 2
      JI = JI - 2
      IF (JI-JL) 110, 110, 120
110  JI = 2*JR - 1
      JL = JR
120  CONTINUE
      RETURN
      END

```

C

C-----

----

```

C SUBROUTINE:  ORD1
C IN-PLACE REORDERING SUBROUTINE

```

C-----

----

C

```

      SUBROUTINE ORD1(M, B)
      DIMENSION B(2)

```

C

```

      K = 4
      KL = 2
      N = 2**M
      DO 40 J=4,N,2
        IF (K-J) 20, 20, 10
10     T = B(J)
        B(J) = B(K)
        B(K) = T
20     K = K - 2
        IF (K-KL) 30, 30, 40
30     K = 2*J
        KL = J
40  CONTINUE
      RETURN
      END

```

C

C-----

----

```

C SUBROUTINE:  ORD2
C IN-PLACE REORDERING SUBROUTINE

```

C-----

----

C

```

      SUBROUTINE ORD2(M, B)
      DIMENSION L(15), B(2)
      EQUIVALENCE (L15,L(1)), (L14,L(2)), (L13,L(3)), (L12,L(4)),
*      (L11,L(5)), (L10,L(6)), (L9,L(7)), (L8,L(8)), (L7,L(9)),
*      (L6,L(10)), (L5,L(11)), (L4,L(12)), (L3,L(13)),
(L2,L(14)),
*      (L1,L(15))
      N = 2**M
      L(1) = N
      DO 10 K=2,M
        L(K) = L(K-1)/2

```

```

10  CONTINUE
    DO 20 K=M,14
        L(K+1) = 2
20  CONTINUE
    IJ = 2
    DO 40 J1=2,L1,2
    DO 40 J2=J1,L2,L1
    DO 40 J3=J2,L3,L2
    DO 40 J4=J3,L4,L3
    DO 40 J5=J4,L5,L4
    DO 40 J6=J5,L6,L5
    DO 40 J7=J6,L7,L6
    DO 40 J8=J7,L8,L7
    DO 40 J9=J8,L9,L8
    DO 40 J10=J9,L10,L9
    DO 40 J11=J10,L11,L10
    DO 40 J12=J11,L12,L11
    DO 40 J13=J12,L13,L12
    DO 40 J14=J13,L14,L13
    DO 40 JI=J14,L15,L14
        IF (IJ-JI) 30, 40, 40
30  T = B(IJ-1)
    B(IJ-1) = B(JI-1)
    B(JI-1) = T
    T = B(IJ)
    B(IJ) = B(JI)
    B(JI) = T
40  IJ = IJ + 2
    RETURN
    END

```

C

C-----

```

C SUBROUTINE:  FFT842
C FAST FOURIER TRANSFORM FOR N=2**M
C COMPLEX INPUT
C-----

```

C

```

    SUBROUTINE FFT842(IN, N, X, Y)

```

C

```

C THIS PROGRAM REPLACES THE VECTOR Z=X+IY BY ITS FINITE
C DISCRETE, COMPLEX FOURIER TRANSFORM IF IN=0.  THE INVERSE
C TRANSFORM

```

```

C IS CALCULATED FOR IN=1.  IT PERFORMS AS MANY BASE
C 8 ITERATIONS AS POSSIBLE AND THEN FINISHES WITH A BASE 4
C ITERATION

```

```

C OR A BASE 2 ITERATION IF NEEDED.

```

C

```

C THE SUBROUTINE IS CALLED AS SUBROUTINE FFT842 (IN,N,X,Y).
C THE INTEGER N (A POWER OF 2), THE N REAL LOCATION ARRAY X, AND
C THE N REAL LOCATION ARRAY Y MUST BE SUPPLIED TO THE SUBROUTINE.

```



```

C      DIMENSION X(2), Y(2), L(15)
      COMMON /CON2/ PI2, P7
      EQUIVALENCE (L15,L(1)), (L14,L(2)), (L13,L(3)), (L12,L(4)),
*      (L11,L(5)), (L10,L(6)), (L9,L(7)), (L8,L(8)), (L7,L(9)),
*      (L6,L(10)), (L5,L(11)), (L4,L(12)), (L3,L(13)),
(L2,L(14)),
*      (L1,L(15))
C
C
C IW IS A MACHINE DEPENDENT WRITE DEVICE NUMBER
C
C      IW = ILMACH(2)
      IW=6
C
      PI2 = 8.*ATAN(1.)
      P7 = 1./SQRT(2.)
      DO 10 I=1,15
        M = I
        NT = 2**I
        IF (N.EQ.NT) GO TO 20
10     CONTINUE
      WRITE (IW,9999)
9999  FORMAT (35H N IS NOT A POWER OF TWO FOR FFT842)
      STOP
20     N2POW = M
      NTHPO = N
      FN = NTHPO
      IF (IN.EQ.1) GO TO 40
      DO 30 I=1,NTHPO
        Y(I) = -Y(I)
30     CONTINUE
40     N8POW = N2POW/3
      IF (N8POW.EQ.0) GO TO 60
C
C RADIX 8 PASSES, IF ANY.
C
      DO 50 IPASS=1,N8POW
        NXTLT = 2** (N2POW-3*IPASS)
        LENGT = 8*NXTLT
        CALL R8TX(NXTLT, NTHPO, LENGT, X(1), X(NXTLT+1),
X(2*NXTLT+1),
*      X(3*NXTLT+1), X(4*NXTLT+1), X(5*NXTLT+1), X(6*NXTLT+1),
*      X(7*NXTLT+1), Y(1), Y(NXTLT+1), Y(2*NXTLT+1),
Y(3*NXTLT+1),
*      Y(4*NXTLT+1), Y(5*NXTLT+1), Y(6*NXTLT+1), Y(7*NXTLT+1))
50     CONTINUE
C
C IS THERE A FOUR FACTOR LEFT
C
C      60 IF (N2POW-3*N8POW-1) 90, 70, 80
C
C GO THROUGH THE BASE 2 ITERATION

```

```

C
C
70 CALL R2TX(NTHPO, X(1), X(2), Y(1), Y(2))
   GO TO 90
C
C GO THROUGH THE BASE 4 ITERATION
C
80 CALL R4TX(NTHPO, X(1), X(2), X(3), X(4), Y(1), Y(2), Y(3),
Y(4))
C
90 DO 110 J=1,15
    L(J) = 1
    IF (J-N2POW) 100, 100, 110
100 L(J) = 2**(N2POW+1-J)
110 CONTINUE
    IJ = 1
    DO 130 J1=1,L1
    DO 130 J2=J1,L2,L1
    DO 130 J3=J2,L3,L2
    DO 130 J4=J3,L4,L3
    DO 130 J5=J4,L5,L4
    DO 130 J6=J5,L6,L5
    DO 130 J7=J6,L7,L6
    DO 130 J8=J7,L8,L7
    DO 130 J9=J8,L9,L8
    DO 130 J10=J9,L10,L9
    DO 130 J11=J10,L11,L10
    DO 130 J12=J11,L12,L11
    DO 130 J13=J12,L13,L12
    DO 130 J14=J13,L14,L13
    DO 130 JI=J14,L15,L14
    IF (IJ-JI) 120, 130, 130
120 R = X(IJ)
    X(IJ) = X(JI)
    X(JI) = R
    FI = Y(IJ)
    Y(IJ) = Y(JI)
    Y(JI) = FI
130 IJ = IJ + 1
    IF (IN.EQ.1) GO TO 150
    DO 140 I=1,NTHPO
        Y(I) = -Y(I)
140 CONTINUE
    GO TO 170
150 DO 160 I=1,NTHPO
        X(I) = X(I)/FN
        Y(I) = Y(I)/FN
160 CONTINUE
170 RETURN
    END

```

C

C-----

----

C SUBROUTINE: R2TX  
 C RADIX 2 ITERATION SUBROUTINE

C-----  
 ----

C  
 SUBROUTINE R2TX(NTHPO, CR0, CR1, CI0, CI1)  
 DIMENSION CR0(2), CR1(2), CI0(2), CI1(2)  
 DO 10 K=1,NTHPO,2  
   R1 = CR0(K) + CR1(K)  
   CR1(K) = CR0(K) - CR1(K)  
   CR0(K) = R1  
   FI1 = CI0(K) + CI1(K)  
   CI1(K) = CI0(K) - CI1(K)  
   CI0(K) = FI1  
 10 CONTINUE  
 RETURN  
 END

C  
 C-----  
 ----

C SUBROUTINE: R4TX  
 C RADIX 4 ITERATION SUBROUTINE

C-----  
 ----

C  
 SUBROUTINE R4TX(NTHPO, CR0, CR1, CR2, CR3, CI0, CI1, CI2,  
 CI3)  
 DIMENSION CR0(2), CR1(2), CR2(2), CR3(2), CI0(2), CI1(2),  
 CI2(2),  
 \*    CI3(2)  
 DO 10 K=1,NTHPO,4  
   R1 = CR0(K) + CR2(K)  
   R2 = CR0(K) - CR2(K)  
   R3 = CR1(K) + CR3(K)  
   R4 = CR1(K) - CR3(K)  
   FI1 = CI0(K) + CI2(K)  
   FI2 = CI0(K) - CI2(K)  
   FI3 = CI1(K) + CI3(K)  
   FI4 = CI1(K) - CI3(K)  
   CR0(K) = R1 + R3  
   CI0(K) = FI1 + FI3  
   CR1(K) = R1 - R3  
   CI1(K) = FI1 - FI3  
   CR2(K) = R2 - FI4  
   CI2(K) = FI2 + R4  
   CR3(K) = R2 + FI4  
   CI3(K) = FI2 - R4  
 10 CONTINUE  
 RETURN  
 END

C  
 C-----  
 ----

C SUBROUTINE: R8TX  
 C RADIX 8 ITERATION SUBROUTINE

C-----  
 C  
 SUBROUTINE R8TX(NXTLT, NTHPO, LENGT, CR0, CR1, CR2, CR3, CR4,  
 \* CR5, CR6, CR7, CI0, CI1, CI2, CI3, CI4, CI5, CI6, CI7)  
 DIMENSION CR0(2), CR1(2), CR2(2), CR3(2), CR4(2), CR5(2),  
 CR6(2),  
 \* CR7(2), CI1(2), CI2(2), CI3(2), CI4(2), CI5(2), CI6(2),  
 \* CI7(2), CI0(2)  
 COMMON /CON2/ PI2, P7

C  
 SCALE = PI2/FLOAT(LENGT)  
 DO 30 J=1,NXTLT  
 ARG = FLOAT(J-1)\*SCALE  
 C1 = COS(ARG)  
 S1 = SIN(ARG)  
 C2 = C1\*\*2 - S1\*\*2  
 S2 = C1\*S1 + C1\*S1  
 C3 = C1\*C2 - S1\*S2  
 S3 = C2\*S1 + S2\*C1  
 C4 = C2\*\*2 - S2\*\*2  
 S4 = C2\*S2 + C2\*S2  
 C5 = C2\*C3 - S2\*S3  
 S5 = C3\*S2 + S3\*C2  
 C6 = C3\*\*2 - S3\*\*2  
 S6 = C3\*S3 + C3\*S3  
 C7 = C3\*C4 - S3\*S4  
 S7 = C4\*S3 + S4\*C3  
 DO 20 K=J,NTHPO,LENGT  
 AR0 = CR0(K) + CR4(K)  
 AR1 = CR1(K) + CR5(K)  
 AR2 = CR2(K) + CR6(K)  
 AR3 = CR3(K) + CR7(K)  
 AR4 = CR0(K) - CR4(K)  
 AR5 = CR1(K) - CR5(K)  
 AR6 = CR2(K) - CR6(K)  
 AR7 = CR3(K) - CR7(K)  
 AI0 = CI0(K) + CI4(K)  
 AI1 = CI1(K) + CI5(K)  
 AI2 = CI2(K) + CI6(K)  
 AI3 = CI3(K) + CI7(K)  
 AI4 = CI0(K) - CI4(K)  
 AI5 = CI1(K) - CI5(K)  
 AI6 = CI2(K) - CI6(K)  
 AI7 = CI3(K) - CI7(K)  
 BR0 = AR0 + AR2  
 BR1 = AR1 + AR3  
 BR2 = AR0 - AR2  
 BR3 = AR1 - AR3  
 BR4 = AR4 - AI6  
 BR5 = AR5 - AI7  
 BR6 = AR4 + AI6

```

BR7 = AR5 + AI7
BI0 = AI0 + AI2
BI1 = AI1 + AI3
BI2 = AI0 - AI2
BI3 = AI1 - AI3
BI4 = AI4 + AR6
BI5 = AI5 + AR7
BI6 = AI4 - AR6
BI7 = AI5 - AR7
CR0(K) = BR0 + BR1
CI0(K) = BI0 + BI1
IF (J.LE.1) GO TO 10
CR1(K) = C4*(BR0-BR1) - S4*(BI0-BI1)
CI1(K) = C4*(BI0-BI1) + S4*(BR0-BR1)
CR2(K) = C2*(BR2-BI3) - S2*(BI2+BR3)
CI2(K) = C2*(BI2+BR3) + S2*(BR2-BI3)
CR3(K) = C6*(BR2+BI3) - S6*(BI2-BR3)
CI3(K) = C6*(BI2-BR3) + S6*(BR2+BI3)
TR = P7*(BR5-BI5)
TI = P7*(BR5+BI5)
CR4(K) = C1*(BR4+TR) - S1*(BI4+TI)
CI4(K) = C1*(BI4+TI) + S1*(BR4+TR)
CR5(K) = C5*(BR4-TR) - S5*(BI4-TI)
CI5(K) = C5*(BI4-TI) + S5*(BR4-TR)
TR = -P7*(BR7+BI7)
TI = P7*(BR7-BI7)
CR6(K) = C3*(BR6+TR) - S3*(BI6+TI)
CI6(K) = C3*(BI6+TI) + S3*(BR6+TR)
CR7(K) = C7*(BR6-TR) - S7*(BI6-TI)
CI7(K) = C7*(BI6-TI) + S7*(BR6-TR)
GO TO 20
10 CR1(K) = BR0 - BR1
   CI1(K) = BI0 - BI1
   CR2(K) = BR2 - BI3
   CI2(K) = BI2 + BR3
   CR3(K) = BR2 + BI3
   CI3(K) = BI2 - BR3
   TR = P7*(BR5-BI5)
   TI = P7*(BR5+BI5)
   CR4(K) = BR4 + TR
   CI4(K) = BI4 + TI
   CR5(K) = BR4 - TR
   CI5(K) = BI4 - TI
   TR = -P7*(BR7+BI7)
   TI = P7*(BR7-BI7)
   CR6(K) = BR6 + TR
   CI6(K) = BI6 + TI
   CR7(K) = BR6 - TR
   CI7(K) = BI6 - TI
20 CONTINUE
30 CONTINUE
RETURN
END

```

\*\*\*\*\*  
 CCXTRA.FOR  
 \*\*\*\*\*

```

      SUBROUTINE RP(X,Y,R,TH,N,ID)
C   REC - POLAR CONVERSION
C   ID NOT =1 ==> R->P
C   ID =1 ==> P->R
      REAL*8 X(16384),Y(16384),R(16384),TH(16384)
      REAL*8 DSQRT,DATAN2,DCOS,DSIN
      IF(ID.EQ.1)GO TO 10
      DO 5 I=1,N
      R(I)=DSQRT(X(I)*X(I)+Y(I)*Y(I))
      TH(I)=0.D0
      IF(Y(I).EQ.0.D0.AND.X(I).EQ.0.D0)GO TO 5
      TH(I)=DATAN2(Y(I),X(I))
      5 CONTINUE
      RETURN
      10 CONTINUE
      DO 15 I=1,N
      X(I)=R(I)*DCOS(TH(I))
      15 Y(I)=R(I)*DSIN(TH(I))
      RETURN
      END

C
C
C
      SUBROUTINE MR1DF(LOG2N,X,Y,SIGN)
C   FORTRAN VERSION
C   MIXED RADIX FOURIER TRANSFORM
      INTEGER LOG2N
      REAL*8 X(16384),Y(16384)
      DIMENSION S(13),U(13)

C
      INTEGER J1,J2,J3,J4,JT,N,M4
      REAL*8 ARG,C1,C2,C3,S1,S2,S3,R1,R2,R3,R4,R5,R6,R7,R8,T
      REAL*8 DCOS,DSIN

C
C
C
      INTEGER A,B,C,D,E,F,G,H,I,J,K,L,M,
      1      BS,CS,DS,ES,FS,GS,HS,IS,JS,KS,LS,MS,
      2      AL,BL,CL,DL,EL,FL,GL,HL,IL,JL,KL,ML,S,U
      EQUIVALENCE
      (BS,S(2)),(CS,S(3)),(DS,S(4)),(ES,S(5)),(FS,S(6)),
      1
      (GS,S(7)),(HS,S(8)),(IS,S(9)),(JS,S(10)),(KS,S(11)),(LS,S(12)),
      2
      (MS,S(13)),(AL,U(1)),(BL,U(2)),(CL,U(3)),(DL,U(4)),(EL,U(5)),
      3
      (FL,U(6)),(GL,U(7)),(HL,U(8)),(IL,U(9)),(JL,U(10)),(KL,U(11)),
      4 (LL,U(12)),(ML,U(13))

```

C

```

      N=2**LOG2N
      IF (SIGN) 8000,8000,8002
8000  DO 8001 I=1,N
8001  Y(I)=-Y(I)
8002  CONTINUE
      IF (LOG2N-1) 500,500,501
501   CONTINUE
      DO 400 K=2,LOG2N,2
      M=2**(LOG2N-K)
      M4=4*M
      DO 300 J=1,M
        ARG=6.28315D0*DBLE(FLOAT(J-1)/FLOAT(M4))
        C1=DCOS(ARG)
        S1=DSIN(ARG)
        C2=C1*C1-S1*S1
        S2=C1*S1+C1*S1
        C3=C2*C1-S2*S1
        S3=C2*S1+S2*C1
        DO 200 I=M4,N,M4
          J1=I+J-M4
          J2=J1+M
          J3=J2+M
          J4=J3+M
          R1=X(J1)+X(J3)
          R2=X(J1)-X(J3)
          R3=Y(J1)+Y(J3)
          R4=Y(J1)-Y(J3)
          R5=X(J2)+X(J4)
          R6=X(J2)-X(J4)
          R7=Y(J2)+Y(J4)
          R8=Y(J2)-Y(J4)
          X(J1)=R1+R5
          Y(J1)=R3+R7
          IF(ARG) 101,100,101
101   CONTINUE
          X(J3)=(R2+R8)*C1+(R4-R6)*S1
          Y(J3)=(R4-R6)*C1-(R2+R8)*S1
          X(J2)=(R1-R5)*C2+(R3-R7)*S2
          Y(J2)=(R3-R7)*C2-(R1-R5)*S2
          X(J4)=(R2-R8)*C3+(R4+R6)*S3
          Y(J4)=(R4+R6)*C3-(R2-R8)*S3
          GO TO 200
100   CONTINUE
          X(J3)=R2+R8
          Y(J3)=R4-R6
          X(J2)=R1-R5
          Y(J2)=R3-R7
          X(J4)=R2-R8
          Y(J4)=R4+R6
200   CONTINUE
300   CONTINUE
400   CONTINUE

```

```

500    CONTINUE
      ITEST=LOG2N-(LOG2N/2*2)
      IF(ITEST) 701,700,701
701    CONTINUE
      DO 600 I=1,N,2
      R1=X(I)+X(I+1)
      R2=X(I)-X(I+1)
      R3=Y(I)+Y(I+1)
      R4=Y(I)-Y(I+1)
      X(I)=R1
      Y(I)=R3
      X(I+1)=R2
      Y(I+1)=R4
600    CONTINUE
700    CONTINUE
      MS=N/2
      ML=N
      DO 800 K=2,12
      J=14-K
      S(J)=1
      U(J)=S(J+1)
      IF(S(J+1)-1) 7701,7701,7700
7700    S(J)=S(J+1)/2
7701    CONTINUE
800    CONTINUE
      AL=BS
      JT=0
      DO 900 A=1,AL
      DO 900 B=A,BL,BS
      DO 900 C=B,CL,CS
      DO 900 D=C,DL,DS
      DO 900 E=D,EL,ES
      DO 900 F=E,FL,FS
      DO 900 G=F,GL,GS
      DO 900 H=G,HL,HS
      DO 900 I=H,IL,IS
      DO 900 J=I,JL,JS
      DO 900 K=J,KL,KS
      DO 900 L=K,LL,LS
      DO 900 M=L,ML,MS
      JT=JT+1
      IF(JT-M) 900,900,901
901    CONTINUE
      T=X(JT)
      X(JT)=X(M)
      X(M)=T
      T=Y(JT)
      Y(JT)=Y(M)
      Y(M)=T
900    CONTINUE
      RETURN
      END

```





```

      AMULT=A/SQRT(A**2+B**2)
      BMULT=B/SQRT(A**2+B**2)
C... DO 100 I=1,N+2,2
      X=F(I)
      Y=F(I+1)
      F(I)=AMULT*X+BMULT*Y
      F(I+1)=-BMULT*X+AMULT*Y
100  CONTINUE
      RETURN
      END

```

**Appendix C**

**PATTERN RECOGNITION  
SCATTER PLOT PRODUCTION**

## Appendix C

### PATTERN RECOGNITION SCATTER PLOT PRODUCTION

Start with several sampled data files that represent the data from which pattern similarities are to be extracted. Each segment of data to be analyzed is labeled. Label files must be created with the ILS command \$ LBF 0, which is similar to the manner in which sampled data files are created with the \$ FIL 0 command as in this example. Use of a command file is suggested when many label files are to be created.

To label the segments, use ILS command \$ LBA. Data segments must have been previously selected by use of one of the ILS commands \$ DSP or \$ CUR so that this information is in the user's common file. A label file contains a series of labels, each one of which is a two-line record of ASCII characters that describes a segment of interest in a sampled data file. Next, use ILS command \$ QUR to extract signal features. Because ILS command \$ QUR restricts the user to 32 frequency spectra, a local version was used that selects 32 evenly spaced spectra from the output of a 2048-point FFT. To use QUR (or XUR), first designate the label file that contains pointers to the area of interest in the signal. \$ QUR output will be in the form of feature record files.

Using ILS command \$ SME, create output feature record files which contain mean vector and eigen records. Since \$ SME is now restricted to 20 elements in analysis, first extract the 20 most significant elements in the output records using ILS command \$ MRE. Use the same elements for all records analyzed. If you have used QUR on single data samples, collect the samples in two or more record files using \$ TRE for analysis with \$ SME. In this study, ILS command \$ SME 6 was used. One output record file is produced by \$ SME containing the mean and eigen vector records. Next, using the files input to \$ SME and the output file of \$ SME, perform a principal component analysis of those input files. \$ PCO outputs a record file for each input file. Plot the results with \$ PLR.

Examples are given on the following pages.

Example:

Ten sampled data files exist which are numbered 1405 through 1495 in steps of 10.

Create the label file(s).

\$ LBF O <cr>

OPTIONS TO MANIPULATE LABEL FILES:

ENTER FILENAME [,DK] TO SELECT A FILE

ENTER C <RETURN> TO CREATE A FILE

ENTER I <RETURN> TO SET THE INITIALS

ENTER <RETURN> TO EXIT

->C

ENTER FILENAME [,DK]

->TAM1

DRB0:[ILSMGR.DGN]TAM1.LAB LABEL DATA

0 LABELS, INITIALS:

PRIMARY LABEL FILE

OPTIONS TO MANIPULATE LABEL FILES:

ENTER FILENAME [,DK] TO SELECT A FILE

ENTER C <RETURN> TO CREATE A FILE

ENTER I <RETURN> TO SET THE INITIALS

ENTER <RETURN> TO EXIT

->I

ENTER THE INITIALS

->TAM

DRB0:[ILSMGR.DGN]TEST2.LAB LABEL DATA

0 LABELS, INITIALS: TAM

PRIMARY LABEL FILE

OPTIONS TO MANIPULATE LABEL FILES:

ENTER FILENAME [,DK] TO SELECT A FILE

ENTER C <RETURN> TO CREATE A FILE

ENTER I <RETURN> TO SET THE INITIALS

ENTER <RETURN> TO EXIT

-><cr>

In like manner create the other nine label files.

Use of a command file is suggested when many label files are to be created.

Label the segments, having previously indicated the range by the cursor command or by the \$ DSP range.

Point to the label file by issuing the command \$ LBF TAM1.

DRB0:[ILSMGR.ILS]TAM1.LAB LABEL DATA

1 LABELS, INITIALS:

PRIMARY LABEL FILE

\$ LBA

FIELD-1 IS OBTAINED FROM THE FILE HEADER, @=ABORT

ENTER FIELD-2;FIELD-3;FIELD-4;FIELD-5;FIELD-6

->TAMTEST

List the label file

\$ LLA

DRB0:[ILSMGR.ILS]TAM1.LAB

22-MAY-84

PAGE 1

```

***** LABEL      1 *****
TAMTEST          ; ; ; ;
1; 1; 1;2048; 2000;DRA0:[ILSMGR.ILS]WD1405.          ;27-JAN-84;TAM
Label the rest of the files.
Extract the features
$ XUR
ENTER MODE, EXTRACTION CODE AND NUMBER OF FEATURES
MODE:      M = MEAN FRAMES
           C = CONSECUTIVE FRAMES
FEATURE: 1 = AUTOREGRESSIVE COEFFICIENTS
          2 = REFLECTION COEFFICIENTS
          3 = AUTOCORRELATION COEFFICIENTS
          4 = FREQUENCY SPECTRA
          5 = LP CEPSTRAL COEFFICIENTS
          6 = MEL CEPSTRAL COEFFICIENTS
          7 = RESONANCE FREQ AND BAND (PEAK PICK)
          8 = RESONANCE FREQ AND BAND (ROOT SOLVE)
          9 = NORMALIZED RESONANCE FREQ AND BAND (ROOT SOLVE)

->4
      EXTRACTION CODE =      4
      USING LABEL FILE....DRB0:[ILSMGR.ILS]TXM1.LAB
      USING RECORD FILE....DRB0:[ILSMGR.ILS]WD1000.
      LEVEL 1
      FIELD-1 IS OBTAINED FROM THE FILE HEADER, @=ABORT
      ENTER FIELD-2;FIELD-3;FIELD-4;FIELD-5;FIELD-6 FOR SEARCH
      ENTER <RETURN> TO START SEARCH

->
      DRB0:[ILSMGR.ILS]WD1000.          RECORD      1 STORED
      And so on with the rest of the files.
      Using $ MRE move the first twenty elements in each file
      to another file
      $ MRE
      MOVE FEATURE RECORDS FROM A FILE AND
      EXTRACT SELECTED ITEMS AND ELEMENTS FROM RECORDS
      =====

      ENTER...ELEMENT NUMBER(S) TO EXTRACT FROM EACH ITEM
      USE <ALL> <RETURN> FOR ALL ELEMENTS
      USE <F>   <RETURN> TO FINISH
      USE <A>   <RETURN> TO ABORT .
      MAXIMUM OF 12 ENTRIES/LINE, NEGATIVE IMPLIES INCLUSIVE

->1,-20
->F
      ENTER...ITEM NUMBER(S) TO EXTRACT FROM EACH RECORD
      USE <ALL> <RETURN> FOR ALL ITEMS
      USE <F>   <RETURN> TO FINISH
      USE <A>   <RETURN> TO ABORT
      MAXIMUM OF 12 ENTRIES/LINE, NEGATIVE IMPLIES INCLUSIVE

->ALL
      ENTER...RECORD NUMBER(S) OF EACH RECORD TO TRANSFER
      USE <ALL> <RETURN> FOR ALL RECORDS
      USE <F>   <RETURN> TO FINISH
      USE <T>   <RETURN> TO TEST ON ELEMENTS (MAX=10)
      USE <SK>  NO. REC <RETURN> TO SKIP RECORDS

```

USE <AV> NO. REC <RETURN> TO AVERAGE RECORDS  
 USE <AVI> NO. REC <RETURN> TO AVERAGE ITEMS IN RECORDS  
 USE <A> <RETURN> TO ABORT  
 MAXIMUM OF 12 ENTRIES/LINE, NEGATIVE IMPLIES INCLUSIVE

```
->ALL
  ENTER...FILE NO., DISK NO. OF INPUT FILE
->1010
  ENTER...FILE NO., DISK NO. OF OUTPUT FILE
->1021
  DRB0:[ILSMGR.ILS]WD1021.          RECORD      1 STORED
  And so on with the rest of the files.
  Collect the records in two files for analysis.
$ FIL 1021
  DRB0:[ILSMGR.ILS]WD1021.          RECORD DATA
      12 DK BLKS,      1 RECORDS
  PRIMARY FILE
$ FIL S1031
  DRB0:[ILSMGR.ILS]WD1031.          DOES NOT EXIST
  SECONDARY FILE
$ OPN S2
$ TRE 1,1,5,1,2,5
  USING DRB0:[ILSMGR.ILS]WD1021.
  DRB0:[ILSMGR.ILS]WD1031.          RECORD      1 STORED
  USING DRB0:[ILSMGR.ILS]WD1022.
  DRB0:[ILSMGR.ILS]WD1031.          RECORD      2 STORED
  USING DRB0:[ILSMGR.ILS]WD1023.
  DRB0:[ILSMGR.ILS]WD1031.          RECORD      3 STORED
  USING DRB0:[ILSMGR.ILS]WD1024.
  DRB0:[ILSMGR.ILS]WD1031.          RECORD      4 STORED
  USING DRB0:[ILSMGR.ILS]WD1025.
  DRB0:[ILSMGR.ILS]WD1031.          RECORD      5 STORED
  USING DRB0:[ILSMGR.ILS]WD1026.
  DRB0:[ILSMGR.ILS]WD1032.          RECORD      1 STORED
  USING DRB0:[ILSMGR.ILS]WD1027.
  DRB0:[ILSMGR.ILS]WD1032.          RECORD      2 STORED
  USING DRB0:[ILSMGR.ILS]WD1028.
  DRB0:[ILSMGR.ILS]WD1032.          RECORD      3 STORED
  USING DRB0:[ILSMGR.ILS]WD1029.
  DRB0:[ILSMGR.ILS]WD1032.          RECORD      4 STORED
  USING DRB0:[ILSMGR.ILS]WD1030.
  DRB0:[ILSMGR.ILS]WD1032.          RECORD      5 STORED
$ FIL 1031
  DRB0:[ILSMGR.ILS]WD1031.          RECORD DATA
      12 DK BLKS,      5 RECORDS
  PRIMARY FILE
$ FIL S1033
  DRB0:[ILSMGR.ILS]WD1033.          DOES NOT EXIST
  SECONDARY FILE
$ OPN S3
$ SME 6,2
  DO YOU WANT VARIANCE RATIOS PRINTED? (Y OR N)
->N
  DRB0:[ILSMGR.ILS]WD1033.          RECORD      4 STORED
$ FIL S1034
```

```

DRB0:[ILSMGR.ILS]WD1034.      RECORD DATA
    12 DK BLKS,      0 RECORDS
SECONDARY FILE
$ PCO 2,1033
USER$$DISK:[ILSMGR.ILS]WD1034.  RECORD      5 STORED

USER$$DISK:[ILSMGR.ILS]WD1035.  RECORD      5 STORED
Plot the results using $ PLR
$ PLR
  ILS PLOTTING ROUTINE
    UP TO 20 INPUT FILES
    UP TO 1024 POINTS PLOTTED
  ENTER X-COMPONENT
->1
  ENTER Y-COMPONENT
->2
  ENTER FILE NO., DISK NO.
  AND NO. CONSECUTIVE FILES
  DEFAULTS TO DRB0:[ILSMGR.ILS]WD1031.
->1034,,2
  SYMBOLS USED ARE A-B
  READ IN ANOTHER FILE? (Y,N)
->N
  FILES BEING READ
  ENTER SYMBOL TO PLOT, OR ENTER <ALL>
->ALL
  CC = CARTESIAN
  SL = SEMI-LOG
  LL = LOG-LOG
->CC
  AUTOMATIC SCALING? (Y,N)
->Y
  GRID? (Y,N)
->N
  ENTER OPTION
  E=EXIT
  N[S,A]=NEW FILES
    [START,APPEND]
  C=COMPONENTS
  M[I]=MARK [IDENTIFY]
  R=REASSIGN
  G=GRID
  S=SCALING
  T=TYPE OF PLOT
  D=DATA TO PLOT
  P[N]=PLOT [NO ERASE]
  F[A,E,L]N1,N2=
    FACTOR ANAL
  H=HARD COPY
->E

```



**END**

**FILMED**

**3-85**

**DTIC**